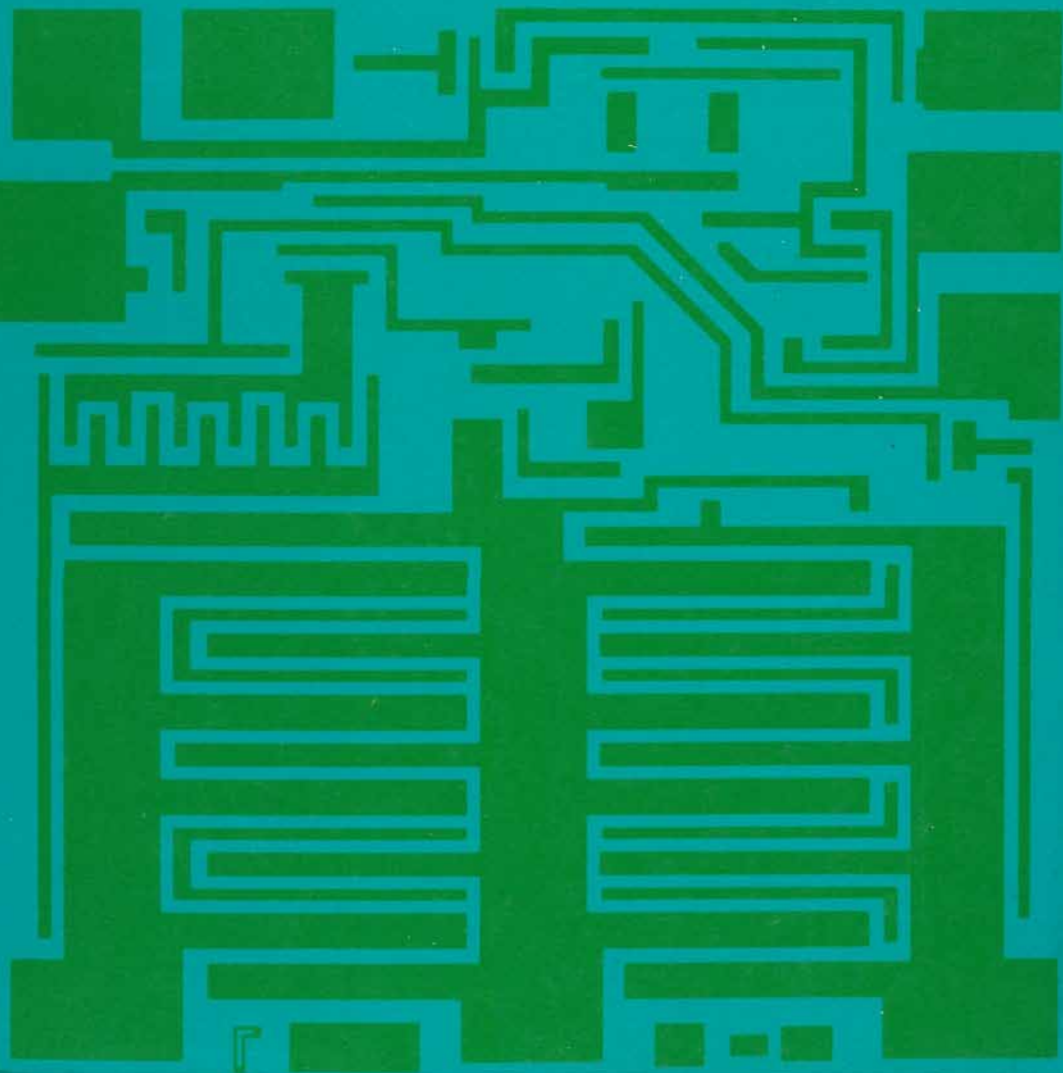


CONSUMER SEMICONDUCTOR

DATA BOOK



INTEGRATED CIRCUITS
SMALL SIGNAL TRANSISTORS
1973/74



CONSUMER SEMICONDUCTOR DATABOOK



INTEGRATED CIRCUITS
SMALL SIGNAL TRANSISTORS
1977-1978

INTRODUCTION

This databook contains data sheets on the SGS-ATES range of small signal transistors and integrated circuits intended for consumer applications.

To permit ease of consultation, this book has been divided into four main sections:

General Information, Germanium Transistors, Silicon Transistors, and Integrated Circuits.

The General Information section contains definitions of symbols and terms used in order to facilitate correct technical interpretation of the data sheets, as well as an alphanumerical list of types.

The information on each product has been specially presented in order that the performance of the product can be readily evaluated within any required equipment design.

An arrow (→) at left hand side of table indicates parameter which has been modified since previous data sheet issue.

OTHER SGS-ATES DATABOOKS

Data sheets on the SGS-ATES range of discrete devices and integrated circuits for professional applications, and high power devices for professional and consumer applications can be found in the following databooks:

SGS-ATES Professional Semiconductor Databook 1 (discrete devices)

SGS-ATES Professional Semiconductor Databook 2 (integrated circuits)

SGS-ATES Power Semiconductor Databook

SGS-ATES GROUP OF COMPANIES

SGS-ATES Componenti Elettronici S.p.A.

International Head Quarter
Via C. Olivetti 1 - 20041 Agrate Brianza - Milano - Italy
Phone: (039) 65341-5 - Telex: 31436

ITALY

SGS-ATES Componenti Elettronici S.p.A.
Via Tempesta, 2
20149 Milano
Tel.: (02) 46 95 651
Telex: 31481

FRANCE

SGS-ATES France S.A.
58, Rue du Dessous des Berges
Paris 13e
Tel.: 589-52-23
Telex: 25938

GERMANY

SGS-ATES Deutschland Halbleiter-Bauelemente GmbH
809 Wasserburg (Inn)
Postfach 1269
Tel.: (08071) 721
Telex: 05-25743

SINGAPORE

SGS-ATES Singapore (Pte) Ltd.
Lorong 4 & 6 Toa Payoh
Singapore 12
Tel.: 53 14 11
Telex: RS 21412

SWEDEN

SGS-ATES Scandinavia AB
Postbox
19501 Märsta
Tel.: 0760/40 120
Telex: 10932

UNITED KINGDOM

SGS-ATES (United Kingdom) Ltd.
Planar House,
Walton Street,
Aylesbury, Bucks
Tel.: (0296) 5977
Telex: 83245

U.S.A.

SGS-ATES Semiconductor Corporation
435 Newtonville Avenue
Newtonville, Mass. 02160
Tel.: (617) 969-1610
Telex: 922482

GENERAL INFORMATION

GERMANIUM TRANSISTORS

SILICON TRANSISTORS

INTEGRATED CIRCUITS

GENERAL INFORMATION

1. LETTER SYMBOLS FOR SEMICONDUCTOR DEVICES	Page	VI
1.1. QUANTITY SYMBOLS	»	VI
1.2. SUBSCRIPTS FOR QUANTITY SYMBOLS	»	VI
1.3. CONVENTIONS FOR SUBSCRIPT SEQUENCE	»	VIII
1.4. ELECTRICAL PARAMETER SYMBOLS	»	IX
1.5. SUBSCRIPTS FOR PARAMETER SYMBOLS	»	X
2. ALPHABETICAL LIST OF SYMBOLS	Page	XI
3. RATING SYSTEMS FOR ELECTRONIC DEVICES	Page	XIX
3.1. DEFINITIONS OF TERMS USED	»	XIX
3.2. ABSOLUTE MAXIMUM RATING SYSTEM	»	XIX
3.3. DESIGN - MAXIMUM RATING SYSTEM	»	XX
3.4. DESIGN - CENTRE RATING SYSTEM	»	XX
4. TYPE DESIGNATION CODE	Page	XXI
4.1. FOR DISCRETE DEVICES	»	XXI
4.2. FOR INTEGRATED CIRCUITS	»	XXII
4.2.1. Types designated by three letters and three figures	»	XXII
4.2.2. Types designated by three letters and four figures	»	XXIV
5. ALPHANUMERICAL LIST OF TYPES	Page	XXV

1. LETTER SYMBOLS FOR SEMICONDUCTOR DEVICES

(referred to diodes, transistors and linear integrated circuits)

1.1. QUANTITY SYMBOLS

- a. Instantaneous values of current, voltage and power, which vary with time are represented by the appropriate lower case letter.

Examples: i , v , p

- b. Maximum (peak), average, d.c. and root-mean-square values are represented by appropriate upper case letter.

Examples: I , V , P

1.2. SUBSCRIPTS FOR QUANTITY SYMBOLS

- a. Total values are indicated by upper case subscripts.

Examples: I_C , i_C , V_{EB} , P_C , p_C

- b. Values of varying components are indicated by lower case subscripts.

Examples: i_c , I_c , v_{eb} , p_c , P_c

- c. To distinguish between maximum (peak), average, d.c. and root-mean-square values, it is possible to represent maximum and average values adding the subscripts m or M and respectively av or AV .

Examples: I_{cm} , I_{CM} , I_{cav} , I_{CAV}

It is possible to represent R.M.S. values by adding the subscripts (rms) and (RMS)

Examples: I_c (rms), I_C (RMS)

- d. List of subscripts (for examples see figure 1 and the fundamental symbols schedule e.)

A, a = Anode terminal

K, k = Cathode terminal

E, e	= Emitter terminal
B, b	= Base terminal
C, c	= Collector terminal
J, j	= Generic terminal
(BR)	= Primary break-down
X, x	= Specified circuit
M, m	= Maximum (peak) value
Min, min	= Minimum value
AV, av	= Average value
(RMS), (rms)	= R.M.S. value
F, f	= Forward
R, r	= As first subscript: Reverse. As second subscript: Repetitive
O, o	= As third subscript: The terminal not mentioned is open circuited
S, s	= As second subscript: Non repetitive. As third subscript: Short circuit between the terminal not mentioned and the reference terminal
Z	= Zener. (Replaces R to indicate the actual zener voltage, current or power of voltage reference or voltage regulator diodes)

e. Fundamental symbols schedule (meaning of symbol with subscript)

	i	v	p	i	V	P
e	instantaneous value of the			R.M.S. value of the variable component,		
b	variable component			or (with appropriate supplementary		
c				subscripts) the maximum or average		
				value (direct current) of the variable		
				component		
E	instantaneous			average value (direct current and		
B	total value			without signal) or (with appropriate		
C				supplementary subscripts) the total		
				average value (with signal), or the total		
				maximum value		

f. Examples of the application of the rules:

Figure 1 represents a transistor collector current, consisting of a direct current and a variable component as a function of time.

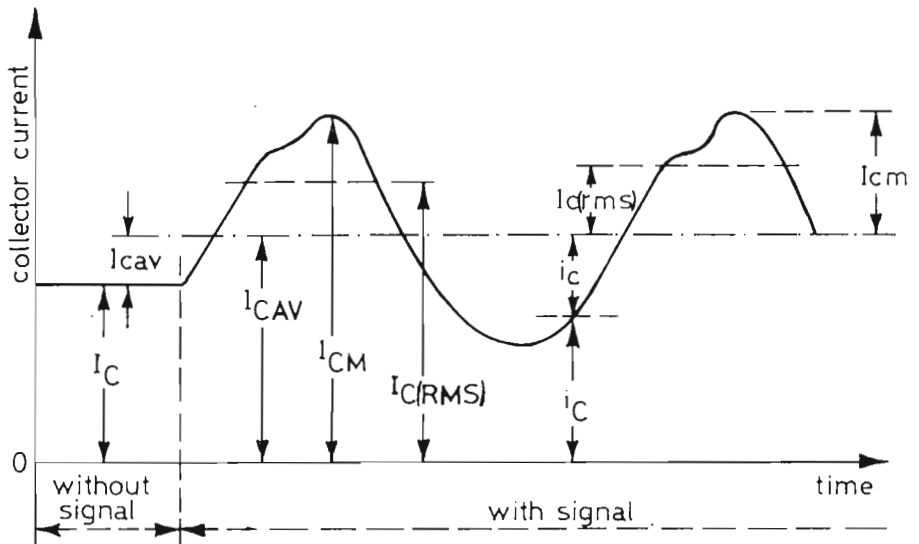


fig. 1

I_C	- DC value, no signal
I_{CAV}	- Average total value
I_{CM}	- Maximum total value
$I_{C(RMS)}$	- R.M.S. total value
I_{cav}	- Average value of the variable component
$I_{c(rms)}$	- R.M.S. value of the variable component
I_{cm}	- Maximum value of the variable component
i_C	- Instantaneous total value
i_c	- Instantaneous value of the variable component

1.3. CONVENTIONS FOR SUBSCRIPT SEQUENCE

a. Currents

For transistor the first subscript indicates the terminal carrying the current (conventional current flow from the external circuit into the terminal is positive).

Instead for diodes a forward current (conventional current flow into the

anode terminal) is represented by the subscript F or f; a reverse current (conventional current flow out of the anode terminal) is represented by the subscript R or r.

b. Voltages

For transistors normally, two subscripts are used to indicate the points between which the voltage is measured. The first subscript indicates one terminal point and the second the reference terminal.

Where there is no possibility of confusion, the second subscript may be omitted.

Instead for diodes a forward voltage (anode positive with respect to cathode) is represented by the subscript F or f and a reverse voltage (anode negative with respect to cathode) by the subscript R or r.

c. Supply voltages

Supply voltages may be indicated by repeating the terminal subscript.

Examples: V_{EE} , V_{CC} , V_{BB}

The reference terminal may then be indicated by a third subscript.

Examples: V_{EEB} , V_{CCB} , V_{BBC}

d. In devices having more than one terminal of the same type, the terminal subscripts are modified by adding a number following the subscript and on the same line.

Example: B_{B2-E} voltage between second base and emitter

In multiple unit devices, the terminal subscripts are modified by a number preceding the terminal subscripts:

Example: V_{1B-2B} voltage between the base of the first unit and that of the second one.

1.4. ELECTRICAL PARAMETER SYMBOLS

a. The values of four pole matrix parameters or other resistances, impedances admittances, etc., inherent in the device, are represented by the lower case symbol with the appropriate subscripts.

Examples: h_{ib} , Z_{fb} , Y_{oc} , h_{FE}

Note: The symbol of the capacitances that is represented by the upper case (C) is an exception to this rule.

b. The four pole matrix parameters of external circuits and of circuits in which the device forms only a part are represented by the upper case symbols with the appropriate subscripts.

Examples: H_i , Z_o , H_F , Y_R

1.5. SUBSCRIPTS FOR PARAMETER SYMBOLS

- a. The static values of parameters are indicated by upper case subscripts.

Examples: h_{IB} , h_{FE}

Note: The static value is the slope of the line from the origin to the operating point on the appropriate characteristic curve, i.e. the quotient of the appropriate electrical quantities at the operating point.

- b. The small-signal values of parameters are indicated by lower case subscripts.

Examples: h_{ib} , Z_{ob}

- c. The first subscript, in matrix notation identifies the element of the four pole matrix.

i (for 11) = input

o (for 22) = output

f (for 21) = forward transfer

r (for 12) = reverse transfer

Examples: $V_1 = h_i I_1 + h_r V_2$
 $I_2 = h_f I_1 + h_o V_2$

Notes

- 1 - The voltage and current symbols in matrix notation are indicated by a single digit subscript.

The subscript 1 = input; the subscript 2 = output.

- 2 - The voltages and currents in these equations may be complex quantities.

- d. The second subscript identifies the circuit configuration.

e = common emitter

b = common base

c = common collector

j = common terminal, general

Examples: (common base)

$$I_1 = y_{ib} V_{1b} + y_{rb} V_{2b}$$
$$I_2 = y_{fb} V_{1b} + y_{ob} V_{2b}$$

When the common terminal is understood, the second subscript may be omitted.

- e. If it is necessary to distinguish between real and imaginary parts of the four pole parameters, the following notations may be used.

$\text{Re}(h_{ib})$ etc... for the real part

$\text{Im}(h_{ib})$ etc... for the imaginary part

2. ALPHABETICAL LIST OF SYMBOLS

AMR	Amplitude modulation rejection
B	Bandwidth
b_{fb}	Common-base, forward transfer susceptance (output short-circuited, y matrix)
b_{fe}	Common-emitter, forward transfer susceptance (output short-circuited, y matrix)
b_{ib}	Common-base, input susceptance (output short-circuited, y matrix)
b_{ie}	Common-emitter, input susceptance (output short-circuited, y matrix)
b_{ob}	Common-base, output susceptance (input short-circuited, y matrix)
b_{oe}	Common-emitter, output susceptance (input short-circuited, y matrix)
b_{rb}	Common-base, reverse transfer susceptance (input short-circuited, y matrix)
b_{re}	Common-emitter, reverse transfer susceptance (input short-circuited, y matrix)
$C_{b'c}$	Intrinsic base-collector capacitance
$C_{b'e}$	Intrinsic base-emitter capacitance
C_{CBO}	Collector-base capacitance (emitter open to a.c. and d.c.)
C_{CSS}	Collector-substrate capacitance (emitter and base open to a.c. and d.c.)
C_{EBO}	Emitter-base capacitance (collector open to a.c. and d.c.)
C_i	Input capacitance
C_{ib}	Common-base, input capacitance (output a.c. short-circuited, h and y matrix)
C_{ibo}	Common-base, input capacitance (output a.c. open-circuited)
C_{ie}	Common-emitter, input capacitance (output a.c. short-circuited, h and y matrix)
C_L	Load capacitance
CMRR	Common mode rejection ratio
C_o	Output capacitance

C_{ob}	Common-base, output capacitance (input a.c. short-circuited, y matrix)
C_{obo}	Common-base, output capacitance (input a.c. open-circuited, h matrix)
C_{oe}	Common-emitter, output capacitance (input a.c. short-circuited, y matrix)
C_{oeo}	Common-emitter, output capacitance (input a.c. open-circuited, h matrix)
C_{rb}	Common-base, reverse capacitance (input a.c. short-circuited, y matrix)
C_{re}	Common-emitter, reverse capacitance (input a.c. short-circuited, y matrix)
d	Distortion
e_N	Noise voltage
$E_{s/b}$	Second breakdown energy (with base-emitter junction reverse biased)
f	Frequency
δf	Frequency change or drift
Δf	Frequency deviation
$\frac{\delta f}{\Delta T} \left(\frac{\Delta f}{\Delta T} \right)$	Frequency drift with temperature variation
$\frac{\delta f}{\Delta V} \left(\frac{\Delta f}{\Delta V} \right)$	Frequency drift with voltage variation
f_{hfb}	Common-base, cut-off frequency
f_{hfe}	Common-emitter, cut-off frequency
f_m	Modulation frequency
f_{max}	Maximum oscillator frequency
f_T	Transition frequency
$f_{,fe}$	Common-emitter cut-off frequency
G_A	Available power gain
G_{AM}	Maximum available power gain
g_{fb}	Common-base, forward transconductance (input short-circuited, y matrix)

g_{fe}	Common-emitter, forward transconductance (input short-circuited, y matrix)
g_{ib}	Common-base, input conductance (output short-circuited, y matrix)
g_{ie}	Common-emitter, input conductance (output short-circuited, y matrix)
g_{ob}	Common-base, output conductance (input short-circuited, y matrix)
g_{oe}	Common-emitter, output conductance (input short-circuited, y matrix)
G_p	Power gain
G_{pb}	Common-base, power gain
G_{pe}	Common-emitter, power gain
G_{pM}	Maximum power gain
g_{rb}	Common-base, reverse transconductance (input short-circuited, y matrix)
g_{re}	Common-emitter, reverse transconductance (input short-circuited, y matrix)
G_{SM}	Maximum stable power gain
G_{tr}	Transducer power gain
G_U	Unilateralized power gain
G_{UM}	Maximum unilateralized power gain
G_v	Voltage gain
h_{fb}	Common-base, small-signal value of the short-circuit forward current transfer ratio
h_{fe}	Common-emitter, small-signal value of the short-circuit forward current transfer ratio
h_{FE}	Common-emitter, static value of the forward current transfer ratio
h_{FE1}/h_{FE2}	Common-emitter, static value of the forward current transfer matched pair ratio
h_{ib}	Common-base, small-signal value of the short-circuit input impedance
h_{ie}	Common-emitter, small-signal value of the short-circuit input impedance
h_{ob}	Common-base, small-signal value of the open-circuit output admittance
h_{oe}	Common-emitter, small-signal value of the open-circuit output admittance

h_{rb}	Common-base, small-signal value of the open-circuit reverse voltage transfer ratio
h_{re}	Common-emitter, small-signal value of the open-circuit reverse voltage transfer ratio
I_b	Bias current
I_B	Base current
I_{B1}	Turn-on current
I_{B2}	Turn-off current
$ I_{B1} - I_{B2} $	Input offset current
I_{BF}	Base forward current
I_{BFM}	Base forward peak current
I_{BM}	Base peak current
I_{BR}	Base reverse current
I_{BRM}	Base reverse peak current
I_C	Collector current
I_{CBO}	Collector cut-off current with emitter open
I_{CBV}	Collector cut-off current with specified reverse voltage between emitter and base
I_{CEO}	Collector cut-off current with base open
I_{CER}	Collector cut-off current with specified resistance between emitter and base
I_{CES}	Collector cut-off current with emitter short-circuited to base
I_{CEV}	Collector cut-off current with specified reverse voltage between emitter and base
I_{CEX}	Collector cut-off current with specified circuit between emitter and base
I_{CM}	Collector peak current
I_d	Drain current
I_E	Emitter current
I_{EBO}	Emitter cut-off current with collector open
i_N	Noise current
I_o	Output current
I_s	Supply current

I_{sc}	Output current during output short-circuit
$I_{s/b}$	Second breakdown collector current (with base-emitter junction forward biased)
I_z	Zener current
m	Modulation factor
NF	Noise figure
NF _c	Conversion noise figure
P_o	Output power of a specified circuit
PRT	Power ratio test
P_{tot}	Total power dissipation
$r_{bb'}$	Base spreading resistance
$r_{bb'}C_{b'c}$	Feedback time constant
R_{BB}	Base dropping resistance
R_{BE}	Resistance between base and emitter
R_{CC}	Collector dropping resistance
R_{EE}	Emitter dropping resistance
R_g	Internal resistance of generator
R_i	Input resistance
R_L	Load resistance
R_o	Output resistance
R_{th}	Thermal resistance
$R_{th\ j-amb}$ ($R_{th\ j-a}$)	Thermal resistance junction-to-ambient
$R_{th\ j-case}$ ($R_{th\ j-c}$)	Thermal resistance junction-to-case
r_z	Dynamic zener resistance
$\frac{S+N}{N}$	Signal and noise to noise ratio
SR	Slew rate
SVR	Supply voltage rejection
t	Time
T_{amb} (T_a)	Ambient temperature
T_{case} (T_c)	Case temperature
t_d	Delay time

t_f	Fall time
T_j	Junction temperature
T_l	Lead temperature
t_{off}	Turn-off-time
t_{on}	Turn-on-time
T_{op}	Operating temperature
t_p	Pulse time
t_r	Rise time
t_s	Storage time
$T_{stg} (T_s)$	Storage temperature
$\frac{\Delta V}{\Delta T}$	Voltage drift with temperature variation
$\frac{\Delta V}{V}$	Relative voltage variation
V_{BE}	Base-emitter voltage
$V_{BE (sat)}$	Base-emitter saturation voltage
$V_{BE1} - V_{BE2}$	Base-emitter voltage difference
$ V_{BE1} - V_{BE2} $	Input offset voltage
$\frac{ V_{BE1} - V_{BE2} }{\Delta T}$	Input-offset voltage temperature coefficient
$V_{(BR) CBO}$	Collector-base breakdown voltage with emitter open
$V_{(BR) CEO}$	Collector-emitter breakdown voltage with base open
$V_{(BR) CER}$	Collector-emitter breakdown voltage with specified resistance
$V_{(BR) CES}$	Collector-emitter breakdown voltage with emitter short-circuited to base
$V_{(BR) CEV}$	Collector-emitter breakdown voltage with specified reverse voltage between emitter and base
$V_{(BR) EBO}$	Emitter-base breakdown voltage with collector open
V_{CB}	Collector-base voltage
V_{CBO}	Collector-base voltage with emitter open
V_{CBV}	Collector-base voltage with specified reverse voltage between emitter and base
V_{CE}	Collector-emitter voltage

V_{CEK}	Knee voltage at specified condition
$V_{CEK (HF)}$	High frequency knee voltage at specified condition
V_{CEO}	Collector-emitter voltage with base open
$V_{CEO (sus)}$	Collector-emitter sustaining voltage with base open
V_{CER}	Collector-emitter voltage with specified resistance between emitter and base
$V_{CER (sus)}$	Collector-emitter sustaining voltage with specified resistance between emitter and base
$V_{CE (sat)}$	Collector-emitter saturation voltage
V_{CES}	Collector-emitter voltage with emitter short-circuited to base
$V_{CES (sus)}$	Collector-emitter sustaining voltage with emitter short-circuited to base
V_{CEV}	Collector-emitter voltage with specified reverse voltage between emitter and base
$V_{CEV (sus)}$	Collector-emitter sustaining voltage with specified reverse voltage between emitter and base
V_{CEX}	Collector-emitter voltage with specified circuit between emitter and base
$V_{CEX (sus)}$	Collector-emitter sustaining voltage with specified circuit between emitter and base
V_{CSS}	Collector-substrate voltage
V_{EB}	Emitter-base voltage
V_{EBO}	Emitter-base voltage with collector open
V_i	Input voltage of a specified circuit
$V_{i(threshold)}$	Input limiting voltage
V_{int}	Interfering voltage
V_o	Output voltage of a specified circuit
V_{pp}	Peak-to-peak voltage
V_{pt}	Punch-through voltage
V_{ref}	Reference voltage
V_s	Supply voltage
V_z	Zener voltage
Y_{fb}	Common-base, small-signal value of the short-circuit forward transfer admittance

y_{fe}	Common-emitter, small-signal value of the short-circuit forward transfer admittance
y_{ib}	Common-base, small-signal value of the short-circuit input admittance
y_{ie}	Common-emitter, small-signal value of the short-circuit input admittance
y_{ob}	Common-base, small-signal value of the short-circuit output admittance
y_{oe}	Common-emitter, small-signal value of the short-circuit output admittance
y_{rb}	Common-base, small-signal value of the short-circuit reverse transfer admittance
y_{re}	Common-emitter, small-signal value of the short-circuit reverse transfer admittance
Z_{BE}	Impedance between base and emitter
Z_i	Input impedance
Z_o	Output impedance
η	Efficiency
η_C	Collector efficiency
τ_s	Storage time constant
φ_{ib}	Common-base, phase angle of the forward transadmittance (output short-circuited, y matrix)
φ_{fe}	Common-emitter, phase angle of the forward transadmittance (output short-circuited, y matrix)
φ_{ib}	Common-base, phase angle of the input admittance (output short-circuited, y matrix)
φ_{ie}	Common-emitter, phase angle of the input admittance (output short-circuited, y matrix)
φ_{ob}	Common-base, phase angle of the output admittance (input short-circuited, y matrix)
φ_{oe}	Common-emitter, phase angle of the output admittance (input short-circuited, y matrix)
φ_{rb}	Common-base, phase angle of the reverse transadmittance (input short-circuited, y matrix)
φ_{re}	Common-emitter, phase angle of the reverse transadmittance (input short-circuited, y matrix)

3. RATING SYSTEMS FOR ELECTRONIC DEVICES

3.1. DEFINITIONS OF TERMS USED

- a. **Electronic device.** An electronic tube or valve, transistor or other semiconductor device.
Note: This definition excludes inductors, capacitors, resistors and similar components.
- b. **Characteristic.** A characteristic is an inherent and measurable property of a device. Such a property may be electrical, mechanical, thermal, hydraulic, electro-magnetic, or nuclear, and can be expressed as a value for stated or recognized conditions. A characteristic may also be a set of related values, usually shown in graphical form.
- c. **Bogey electronic device.** An electronic device whose characteristics have the published nominal values for the type. A bogey electronic device for any particular application can be obtained by considering only those characteristics which are directly related to the application.
- d. **Rating.** A value which establishes either a limiting capability or a limiting condition for an electronic device. It is determined for specified values of environment and operation, and may be stated in any suitable terms.
Note: Limiting conditions may be either maxima or minima.
- e. **Rating system.** The set of principles upon which ratings are established and which determines their interpretation.
Note: The rating system indicates the division of responsibility between the device manufacturer and the circuit designer, with the object of ensuring that the working conditions do not exceed the ratings.

3.2. ABSOLUTE MAXIMUM RATING SYSTEM

Absolute maximum ratings are limiting values of operating and environmental conditions applicable to any electronic device of a specified type as defined by its published data, which should not be exceeded under the worst probable conditions.

These values are chosen by the device manufacturer to provide acceptable serviceability of the device, taking no responsibility for equipment variations, environmental variations, and the effects of changes in operating conditions due to variations in the characteristics of the device under consideration and of all other electronic devices in the equipment.

The equipment manufacturer should design so that, initially and throughout life, no absolute maximum value for the intended service is exceeded with any device under the worst probable operating conditions with respect to supply voltage variation, equipment component variation, equipment control adjustment, load variations, signal variation, environmental conditions, and variations in characteristics of the device under consideration and of all other electronic devices in the equipment.

3.3. DESIGN - MAXIMUM RATING SYSTEM

Design-maximum ratings are limiting values of operating and environmental conditions applicable to a bogey electronic device of a specified type as defined by its published data, and should not be exceeded under the worst probable conditions.

These values are chosen by the device manufacturer to provide acceptable serviceability of the device, taking responsibility for the effects of changes in operating conditions due to variations in the characteristics of the electronic device under consideration.

The equipment manufacturer should design so that, initially and throughout life, no design-maximum value for the intended service is exceeded with a bogey device under the worst probable operating conditions with respect to supply-voltage variation, equipment component variation, variation in characteristics of all other devices in the equipment, equipment control adjustment, load variation, signal variation and environmental conditions.

3.4. DESIGN - CENTRE RATING SYSTEM

Design-centre ratings are limiting values of operating and environmental conditions applicable to a bogey electronic device of a specified type as defined type as defined by its published data, and should not be exceeded under normal conditions.

These values are chosen by the device manufacturer to provide acceptable serviceability of the device in average applications, taking responsibility for normal changes in operating conditions due to rated supply-voltage variation, equipment component variation, equipment control adjustment, load variation, signal variation, environmental conditions, and variations in the characteristics of all electronic devices.

The equipment manufacturer should design so that, initially, no design-centre value for the intended service is exceeded with a bogey electronic device in equipment operating at the stated normal supply-voltage.

The Absolute Maximum Rating System is commonly used for semiconductor devices.

4. TYPE DESIGNATION CODE

4.1. FOR DISCRETE DEVICES

The type number for "discrete" semiconductor devices consists of:
TWO LETTERS FOLLOWED BY A SERIAL NUMBER

The first letter gives information about the **material** used for the active part of the devices:

- A Material with a band gap of 0.6 to 1.0eV, such as germanium
- B Material with a band gap of 1.0 to 1.3eV, such as silicon
- C Material with a band gap of 1.3eV and more, such as gallium arsenide
- D Material with a band gap of less than 0.6eV, such as indium antimonide
- R Compound material as employed in Hall generators and photoconductive cells, such as cadmium-sulphide, lead-selenide

The second letter indicates the **function** according with the applications and the construction:

- A Detection diode, switching diode, mixer diode
- B Variable capacitance diode
- C Transistor for a.f. applications (Rth j-case $> 15^{\circ}\text{C/W}$)
- D Power transistor for a.f. applications (Rth j-case $\leq 15^{\circ}\text{C/W}$)
- E Tunnel diode
- F Transistor for h.f. applications (Rth j-case $> 15^{\circ}\text{C/W}$)
- G Multiple of dissimilar devices (1); Miscellaneous
- H Magnetic sensitive diode; Field probe
- K Hall generator in an open magnetic circuit, e.g. magnetogram or signal probe
- L Power transistor for h.f. applications (Rth j-case $\leq 15^{\circ}\text{C/W}$)
- M Hall generator in a closed electrically energised magnetic circuit, e.g. Hall modulator or multiplier
- P Radiation sensitive device
- Q Radiation generating device
- R Electrically triggered controlling and switching device having a breakdown characteristic (Rth j-case $> 15^{\circ}\text{C/W}$)
- S Transistor for switching applications (Rth j-case $> 15^{\circ}\text{C/W}$)
- T Electrically, or by means of light, triggered controlling and switching power device having a breakdown characteristic (Rth j-case $\leq 15^{\circ}\text{C/W}$)
- U Power transistor for switching applications (Rth j-case $\leq 15^{\circ}\text{C/W}$)
- X Multiplier diode, e.g. varactor, step recovery diode
- Y Rectifying diode, booster diode, efficiency diode
- Z Voltage reference or voltage regulator diode

- 1) A multiple device is defined as a combination of similar or dissimilar active devices, contained in a common encapsulation that cannot be dismantled, and of which all electrodes of the individual devices are accessible from the outside.

Multiples of similar devices as well as multiples consisting of a main device and an auxiliary device are designated according to the code for the discrete devices described above.

Multiples of dissimilar devices of other nature are designated by the second letter G.

The serial number is formed by:

Three figures for semiconductor devices which are primarily intended for use in domestic equipment.

Two figures and a letter (this letter starts back from z through y, x, etc. bears no signification).

Version letter

A version letter can be used, for instance, for a diode with up-rated voltage, for a sub-division of a transistor type in different gain ranges, a low noise version of an existing transistor and for a diode, transistor, or thyristor with minor mechanical differences, such as finish of the leads, length of the leads etc. The letters never have a fixed meaning, the only exception being the letter R which indicates reverse polarity.

Examples

BC 107 Silicon low power audio frequency transistor primarily intended for domestic equipment

BUY 46 Silicon power transistor for switching applications in professional equipment

4.2. FOR INTEGRATED CIRCUITS

4.2.1. Types designated by three letters and three figures

The integrated circuits are divided in four groups:

- digital types belonging to a family of circuits;
- digital solitary circuits;
- analogue circuits including linear circuits;
- mixed analogue/digital circuits.

Digital Family Types

First two letters:	family
Third letter:	circuit function
First two figures:	serial number
Third figure:	operating ambient temperature

Digital Solitary Types

First letter:	"S"
Second letter:	extension of serial number
Third letter:	circuit function
First two figures:	serial number
Third figure:	operating ambient temperature range

Analogue (Linear) Types

First letter:	"T"
Second and third letter:	extension of serial number
First two figures:	serial number
Third figure:	operating ambient temperature range

Mixed Digital/Analogue Types

First letter:	"U"
Second and third letter:	extension of serial number
First two figures:	serial number
Third figure:	operating ambient temperature range

Function

- H Combinatorial circuit
- J Bistable or multistable sequential circuit
- K Monostable sequential circuit
- L Level converter
- N Bi-metastable or multi-metastable sequential circuit
- Q Read-write memory circuit
- R Read only memory circuit
- S Sense amplifier with digital output
- Y Miscellaneous

Operating ambient temperature range

- 1 0 to + 70 °C
- 2 -55 to + 125 °C
- 3 -10 to + 85 °C
- 4 +15 to + 55 °C
- 5 -25 to + 70 °C
- 6 -40 to + 85 °C

0 It means no temperature range indicated in the type number

If a circuit is published for a wider temperature range, but does not qualify for a higher classification, the figure indicating the narrower temperature range is used.

Version letter

A version letter can be added to a type number of an existing type to indicate a different version of the same type, for instance, encapsulated

in another package, with other interconnections or showing minor differences in ratings or electrical characteristics. The letter Z is used to indicate a type with discretionary wiring.

4.2.2. Types designated by three letters and four figures

The serial number can be a four figure number assigned by Pro Electron or the serial number of an existing company number.

The first two letters:

A. FAMILY CIRCUITS

The FIRST TWO LETTERS give information about the family of circuits. These letters can be FA...FZ, GA...GZ, HA... etc.

B. SOLITARY CIRCUITS

The FIRST LETTER divides the solitary circuits into:

S Solitary digital circuits

T Analogue circuits

U Mixed analogue/digital circuits

The SECOND LETTER is a serial letter without any further significance.

The third letter indicates the operational temperature range or another significant characteristic.

The letters B thru F give information about the temperature range (note 1):

B 0 °C to + 70 °C

C -55 °C to + 125 °C

D -25 °C to + 70 °C

E -25 °C to + 85 °C

F -40 °C to + 85 °C

Other "third" letters refer to electrical or mechanical versions of a family and have no fixed meaning. If no temperature range or another characteristic is indicated, the letter A is used as a third letter.

The serial number can be either a 4 figure number assigned by Pro Electron or the serial number (also numbers comprising letters) of an existing company type designation. Company serial numbers of less than 4 figures are completed to a 4 figure number by "0" 's in front of the number.

A version letter can be used to indicate a deviation of a single characteristic of a type, either electrically or mechanically. The letter never has a fixed meaning, the only exception being the letter Z, indicating "custom-wired" devices.

Note 1: If a circuit is published for a wider temperature range, but does not qualify for a higher classification, the letter indicating the narrower temperature range is used.

5. ALPHANUMERICAL LIST OF TYPES

Type	Page	Type	Page	Type	Page
AF 106	3	BC 301	107	★ BF 500A	227
AF 109R	5	BC 302	107	★ BF 516	231
AF 139	7	BC 303	113	TAA 550	239
AF 239	13	BC 304	113	TAA 611A	245
AF 239S	19	BC 323	117	TAA 611B	257
BC 107	25	BC 377	119	TAA 611C	267
BC 108	25	BC 378	119	TAA 621	281
BC 109	25	BC 440	123	★ TAA 630S	293
BC 113	33	BC 441	123	TAA 661	299
BC 114	33	BC 460	127	TAA 691	307
BC 115	39	BC 461	127	TBA 231	315
BC 116A	43	BC 477	131	TBA 261	321
BC 119	45	BC 478	131	TBA 271	239
BC 125	47	BC 479	131	TBA 311	327
BC 125B	47	BF 155	139	TBA 331	333
BC 126	51	BF 158	141	TBA 435	341
BC 132	55	BF 160	143	TBA 625A	349
BC 139	57	BF 161	145	TBA 625B	357
BC 140	61	BF 166	147	TBA 625C	365
BC 141	61	BF 167	149	TBA 631	373
BC 153	65	BF 173	155	TBA 641A	383
BC 154	65	BF 222	161	TBA 641B	393
BC 160	73	BF 233	163	TBA 651	403
BC 161	73	BF 234	163	TBA 780	407
BC 177	77	BF 257	167	★ TBA 800	415
BC 178	77	BF 258	167	★ TBA 810S	427
BC 179	77	BF 259	167	★ TBA 810AS	427
BC 204	85	BF 260	171	★ TBA 820	439
BC 205	85	BF 271	181	★ TCA 511	447
BC 206	85	★ BF 272A	185	★ TCA 600	455
BC 207	91	BF 273	191	★ TCA 610	455
BC 208	91	BF 274	195	★ TCA 900	463
BC 209	91	BF 287	197	★ TCA 910	463
BC 225	97	BF 288	201	★ TDA 1200	471
BC 288	99	★ BF 316A	205	★ SAJ 210	479
BC 297	103	★ BF 454	211		
BC 298	103	★ BF 455	217		
BC 300	107	★ BF 479	223		
		★ BF 500	227		

★ new type

GERMANIUM TRANSISTORS

AF 106

GERMANIUM MESA PNP

VHF MIXER/OSCILLATOR

The AF 106 is a germanium mesa PNP transistor in a Jedec TO-72 metal case. It is particularly designed for use as preamplifier mixer and oscillator up to 260 MHz.

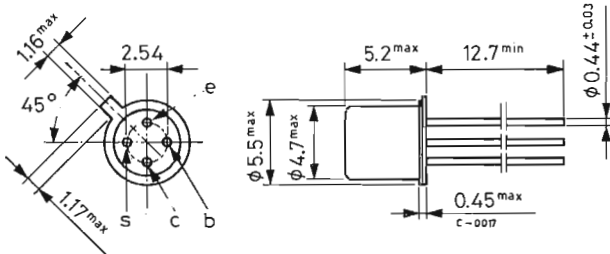
ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-25 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-18 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-0.3 V
I_C	Collector current	-10 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 45^\circ\text{C}$	60 mW
	at $T_{case} \leq 66^\circ\text{C}$	60 mW
T_{stg}	Storage temperature	-30 to 75 °C
T_j	Junction temperature	90 °C

MECHANICAL DATA

Dimensions in mm

Shield lead connected to case



TO-72

AF 106

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	400	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	750	°C/W

ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -12\text{ V}$			-10	μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -100\ \mu\text{A}$	-25			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -500\ \mu\text{A}$	-18			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -100\ \mu\text{A}$	-0.3			V
V_{BE} Base-emitter voltage	$I_C = -1\text{ mA}$ $V_{CE} = -12\text{ V}$ $I_C = -2\text{ mA}$ $V_{CE} = -6\text{ V}$	-0.25	-0.325	-0.38	V
h_{FE} DC current gain	$I_C = -1\text{ mA}$ $V_{CE} = -12\text{ V}$ $I_C = -2\text{ mA}$ $V_{CE} = -6\text{ V}$	20	50	70	—
f_T Transition frequency	$I_C = -1\text{ mA}$ $V_{CE} = -12\text{ V}$ $f = 100\text{ MHz}$		220		MHz
$-C_{re}$ Reverse capacitance	$I_C = -1\text{ mA}$ $V_{CE} = -12\text{ V}$ $f = 450\text{ kHz}$		0.45		pF
NF Noise figure	$I_C = -1\text{ mA}$ $V_{CE} = -12\text{ V}$ $R_g = 60\ \Omega$ $f = 200\text{ MHz}$		5.5	7.5	dB
$r_{bb'}$, $C_{b'c}$ Feedback time constant	$I_C = -1\text{ mA}$ $V_{CE} = -12\text{ V}$ $f = 2.5\text{ MHz}$		6		ps
G_{pb} Power gain	$I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_L = 920\ \Omega$ $f = 200\text{ MHz}$		14	17.5	dB

AF 109R

GERMANIUM MESA PNP

VHF PREAMPLIFIER

The AF 109R is a germanium mesa PNP transistor in a Jeduc TO-72 metal case. It is designed for use in AGC prestages up to 260 MHz.

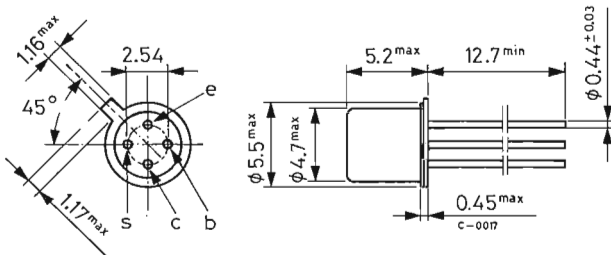
ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-20 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-15 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-0.3 V
I_C	Collector current	-10 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 45^\circ\text{C}$	60 mW
	at $T_{case} \leq 66^\circ\text{C}$	60 mW
T_{stg}	Storage temperature	-30 to 75 °C
T_j	Junction temperature	90 °C

MECHANICAL DATA

Dimensions in mm

Shield lead connected to case



TO-72

AF 109R

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	400	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	750	°C/W

ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -20\text{ V}$		-0.5	-8	μA
I_{CEO} Collector cutoff current ($I_B = 0$)	$V_{CE} = -15\text{ V}$			-500	μA
I_{EBO} Emitter cutoff current ($I_C = 0$)	$V_{EB} = -0.3\text{ V}$			-100	μA
V_{BE} Base-emitter voltage	$I_C = -1.5\text{ mA}$ $V_{CE} = -12\text{ V}$ $I_C = -2\text{ mA}$ $V_{CE} = -6\text{ V}$	-320	-380	-430	mV
h_{FE} DC current gain	$I_C = -1.5\text{ mA}$ $V_{CE} = -12\text{ V}$ $I_C = -2\text{ mA}$ $V_{CE} = -6\text{ V}$	20	50	55	—
$-C_{re}$ Reverse capacitance	$I_C = -1\text{ mA}$ $V_{CE} = -12\text{ V}$ $f = 450\text{ kHz}$		0.25		pF
NF Noise figure	$I_C = -2\text{ mA}$ $V_{CE} = -12\text{ V}$ $R_g = 60\ \Omega$ $f = 200\text{ MHz}$			4.8	dB
G_{db} Power gain	$I_C = -2\text{ mA}$ $V_{CE} = -12\text{ V}$ $R_L = 920\ \Omega$ $R_{EE} = 1\text{ k}\Omega$ $f = 200\text{ MHz}$	13	16.5		dB

AF 139

GERMANIUM MESA PNP

UHF AMPLIFIER

The AF 139 is a germanium mesa PNP transistor in a Jedec TO-72 metal case. It is particularly designed for use in prestages as well as in mixer and oscillator stages up to 860 MHz.

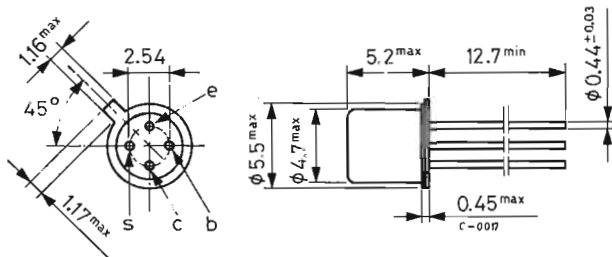
ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-22 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-15 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-0.3 V
I_E	Emitter current	11 mA
I_C	Collector current	-10 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 45^\circ\text{C}$	60 mW
	at $T_{case} \leq 66^\circ\text{C}$	60 mW
T_{stg}	Storage temperature	-30 to 75 °C
T_J	Junction temperature	90 °C

MECHANICAL DATA

Dimensions in mm

Shield lead connected to case



TO-72

AF 139

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	400	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	750	°C/W

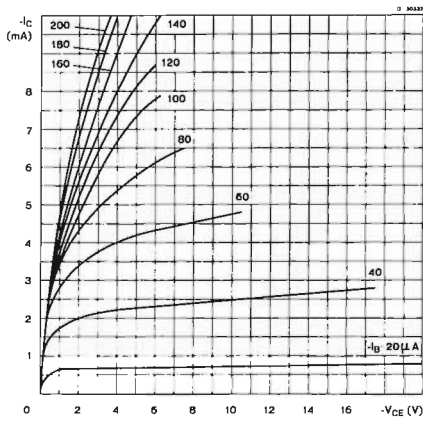
ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
I_{CBO}	Collector cutoff current ($I_E = 0$)	$V_{CE} = -22\text{ V}$			-8	μA
I_{CEO}	Collector cutoff current ($I_B = 0$)	$V_{CE} = -15\text{ V}$			-500	μA
I_{EBO}	Emitter cutoff current ($I_C = 0$)	$V_{EB} = -0.3\text{ V}$			-100	μA
h_{FE}	DC current gain	$I_C = -1.5\text{ mA}$ $V_{CE} = -12\text{ V}$	10	50		—
f_T	Transition frequency	$I_C = -1.5\text{ mA}$ $V_{CE} = -12\text{ V}$ $f = 100\text{ MHz}$		550		MHz
$-C_{re}$	Reverse capacitance	$I_C = -1.5\text{ mA}$ $V_{CE} = -12\text{ V}$ $f = 100\text{ kHz}$		0.25		pF
NF*	Noise figure	$I_C = -1.5\text{ mA}$ $V_{CE} = -12\text{ V}$ $R_q = 60\ \Omega$ $f = 800\text{ MHz}$		7	8.2	dB
$r_{bb}, C_{b'c}$	Feedback time constant	$I_C = -1.5\text{ mA}$ $V_{CE} = -12\text{ V}$ $f = 2.5\text{ MHz}$		3		ps
G_{ob}^*	Power gain	$I_C = -1.5\text{ mA}$ $V_{CE} = -12\text{ V}$ $R_L = 1.4\text{ k}\Omega$ $f = 800\text{ MHz}$	9	11		dB

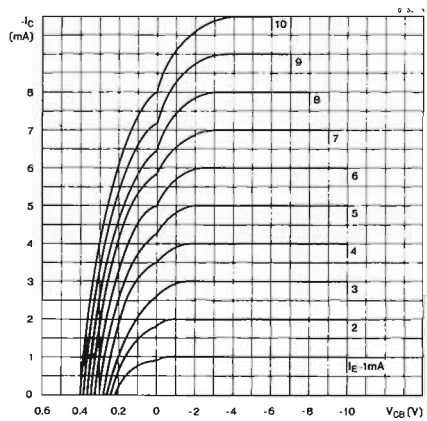
* See test circuit.

AF 139

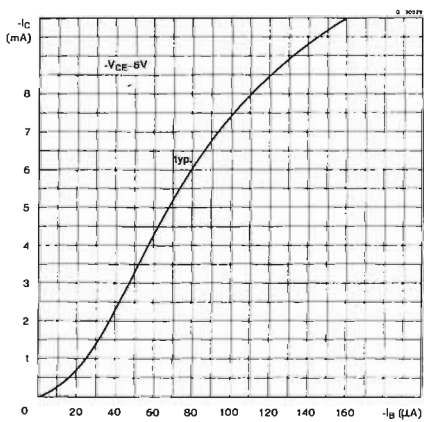
Typical output characteristics



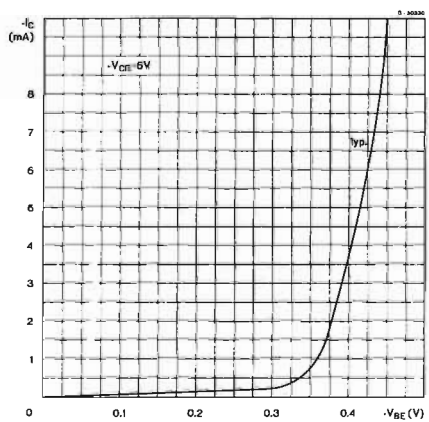
Typical output characteristics



Collector current

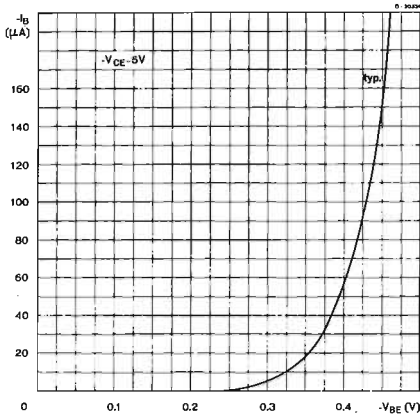


DC transconductance

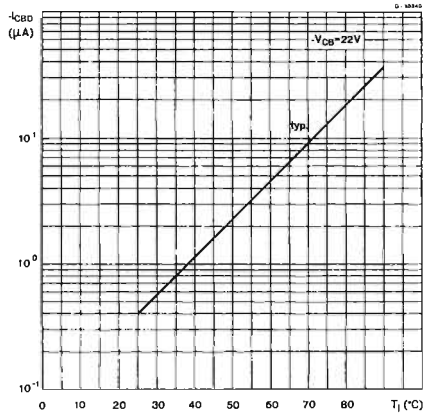


AF 139

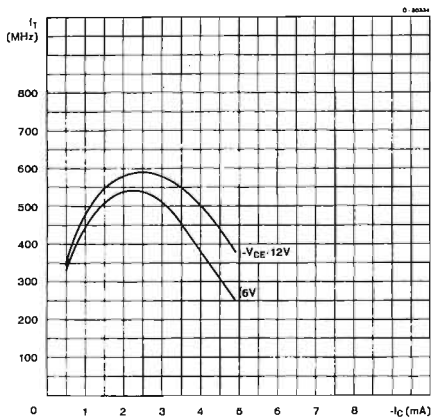
Input characteristics



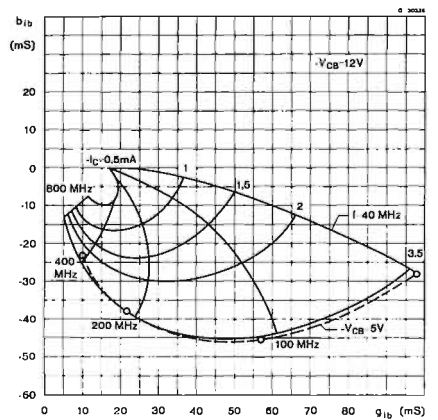
Collector cutoff current



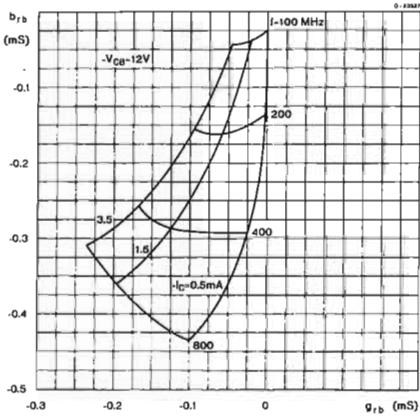
Typical transition frequency



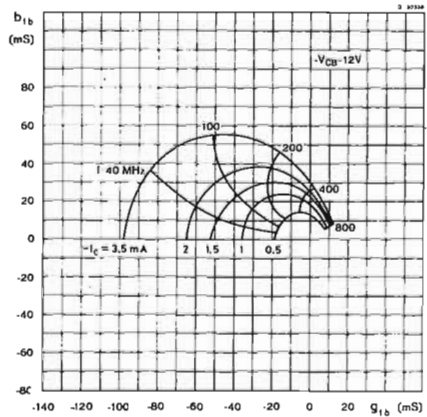
Typical input admittance



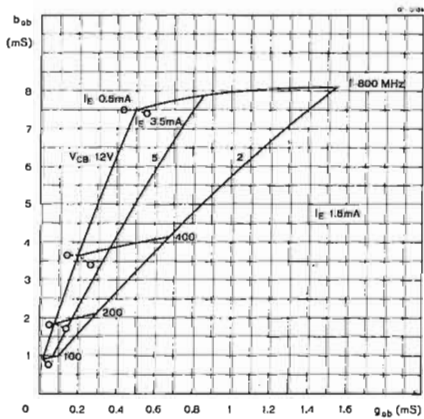
Typical reverse admittance



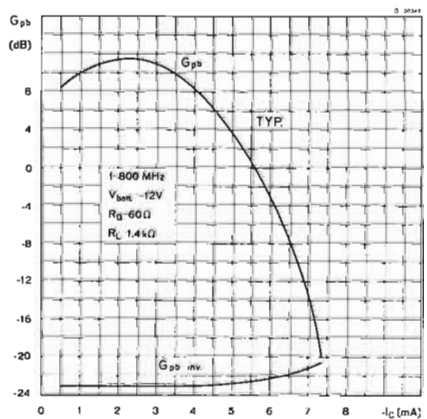
Typical transfer admittance



Typical output admittance

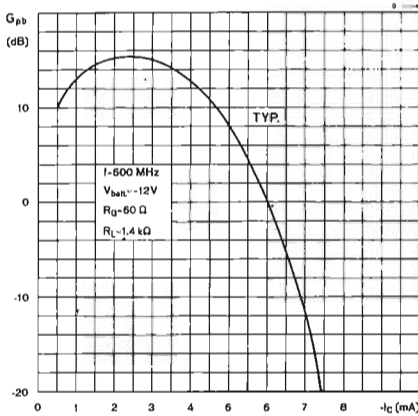


Typical power gain

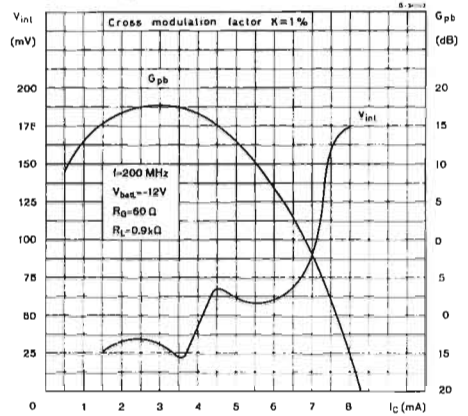


AF 139

Power gain

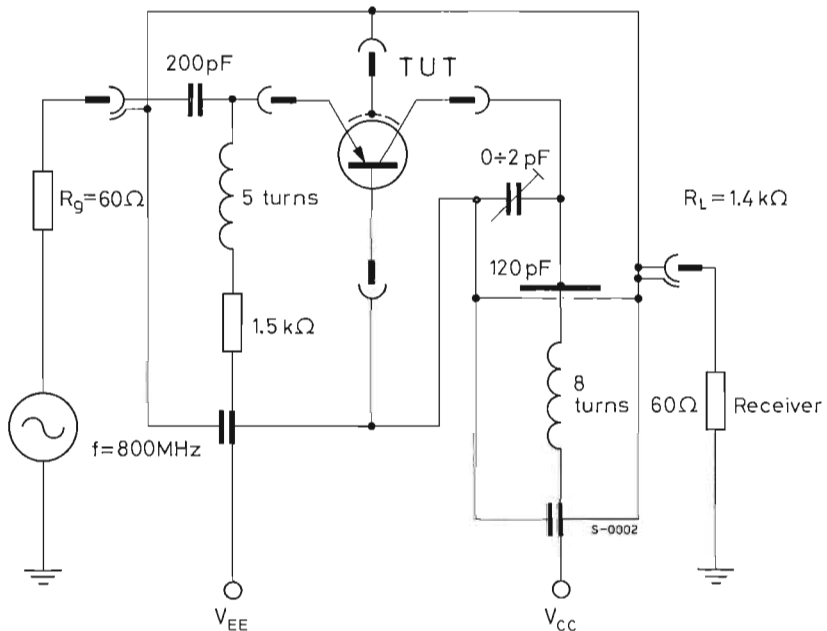


Typical interfering voltage and power gain



TEST CIRCUIT

800 MHz transducer power gain G_{pb} and noise figure test circuit



AF 239

GERMANIUM MESA PNP

UHF PREAMPLIFIER

The AF 239 is a germanium mesa PNP transistor in a Jedec TO-72 metal case. It is particularly designed as preamplifier mixer and oscillator up to 900 MHz.

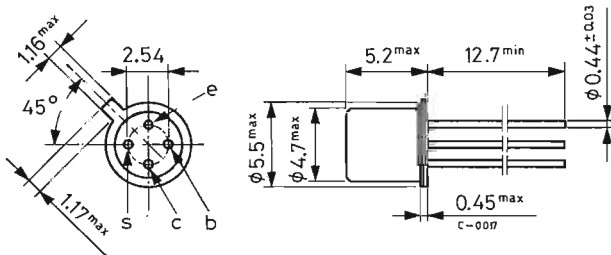
ABSOLUTE MAXIMUM RATINGS

V_{CES}	Collector-emitter voltage ($V_{BE} = 0$)	-20 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-15 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-0.3 V
I_C	Collector current	-10 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 45^\circ\text{C}$	60 mW
	at $T_{case} \leq 66^\circ\text{C}$	60 mW
T_{stg}	Storage temperature	-30 to 75 °C
T_J	Junction temperature	90 °C

MECHANICAL DATA

Dimensions in mm

Shield lead connected to case



TO-72

AF 239

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	400	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	750	°C/W

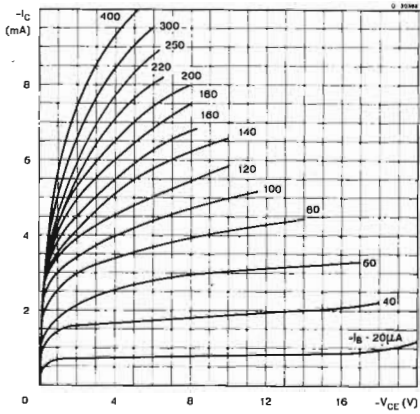
ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	
I_{CES}	Collector cutoff current ($V_{BE} = 0$)			-8	μA	
I_{CEO}	Collector cutoff current ($I_B = 0$)			-500	μA	
I_{EBO}	Emitter cutoff current ($I_C = 0$)			-100	μA	
V_{BE}	Base-emitter voltage	$I_C = -2\text{ mA}$ $I_C = -5\text{ mA}$	$V_{CE} = -10\text{ V}$ $V_{CE} = -5\text{ V}$	-350 -400	mV mV	
h_{FE}	DC current gain	$I_C = -2\text{ mA}$ $I_C = -5\text{ mA}$	$V_{CE} = -10\text{ V}$ $V_{CE} = -5\text{ V}$	10 30	— —	
f_T	Transition frequency	$I_C = -2\text{ mA}$ $f = 100\text{ MHz}$	$V_{CE} = -10\text{ V}$	700	MHz	
$-C_{re}$	Reverse capacitance	$I_C = -2\text{ mA}$ $f = 450\text{ kHz}$	$V_{CE} = -10\text{ V}$	0.23	pF	
NF*	Noise figure	$I_C = -2\text{ mA}$ $R_g = 60\ \Omega$	$V_{CE} = -10\text{ V}$ $f = 800\text{ MHz}$	5	6	dB
G_{pb}^*	Power gain	$I_C = -2\text{ mA}$ $R_L = 2\text{ k}\Omega$	$V_{CE} = -10\text{ V}$ $f = 800\text{ MHz}$	11	14	dB

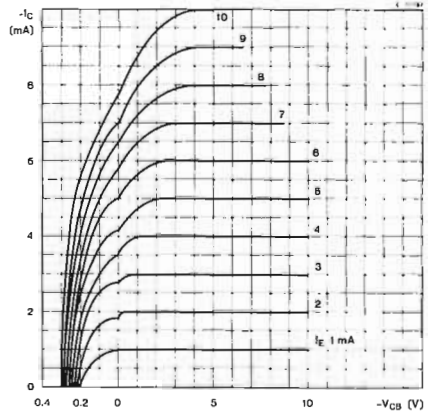
* See test circuit.

AF 239

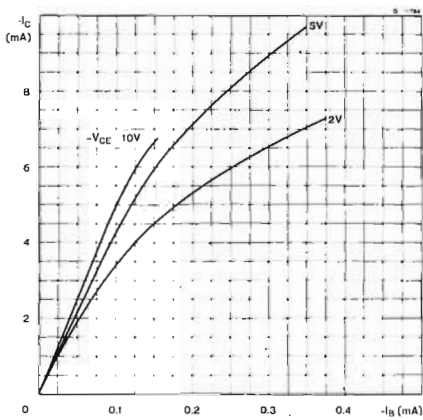
Typical output characteristics



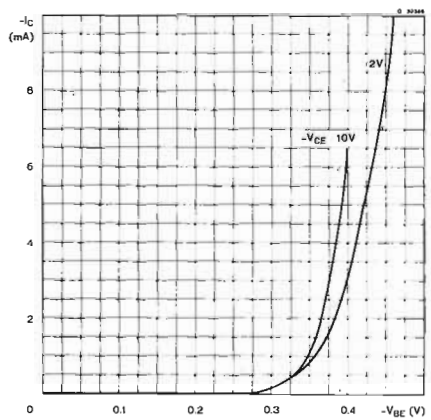
Typical output characteristics



Typical collector current

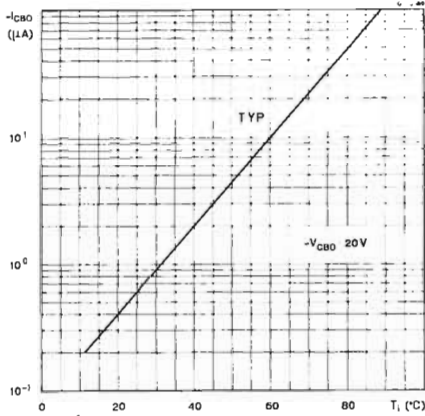


Typical DC transconductance

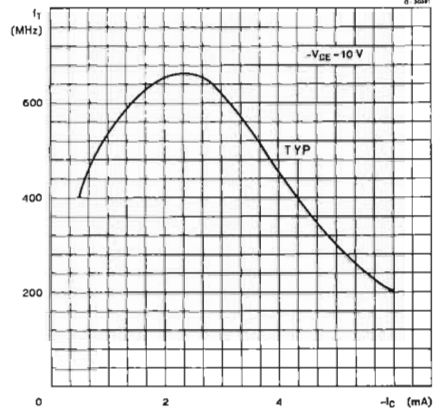


AF 239

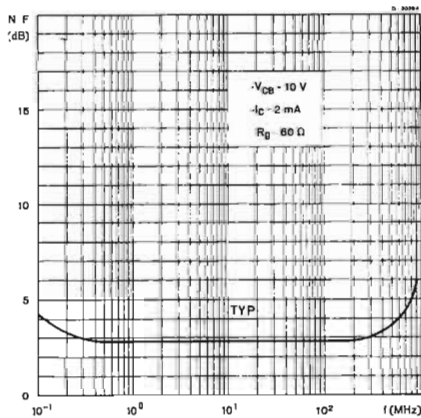
Collector cutoff current



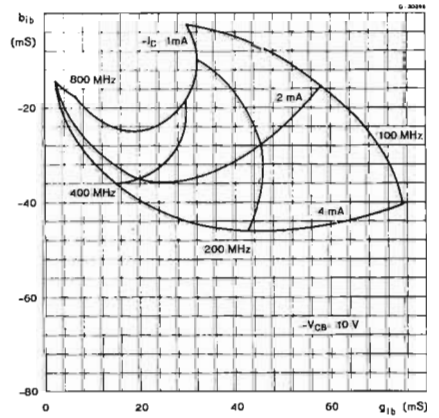
Transition frequency



Noise figure

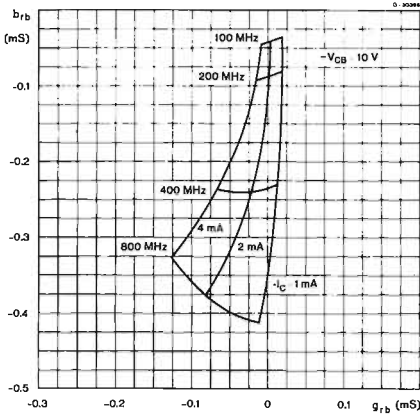


Typical input admittance

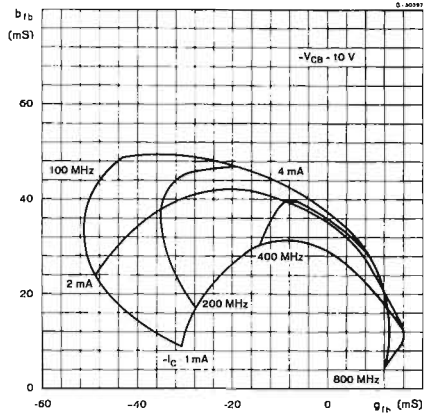


AF 239

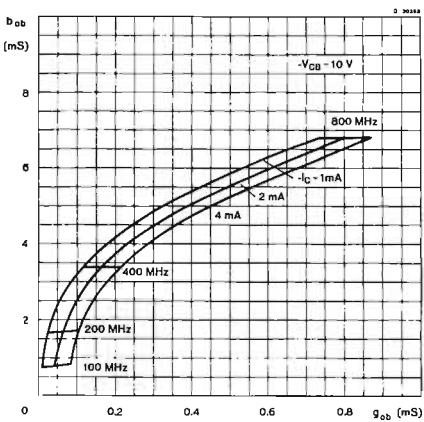
Typical reverse admittance



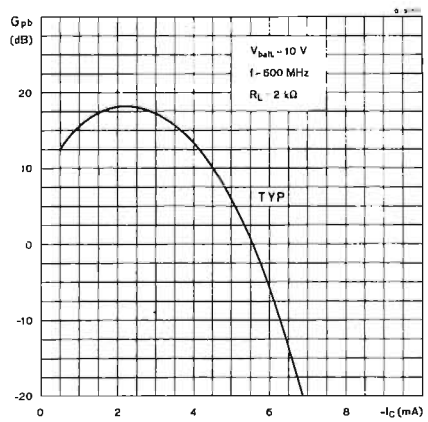
Typical transfer admittance



Typical output admittance

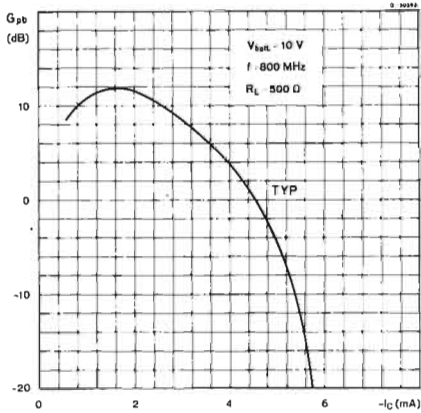


Power gain

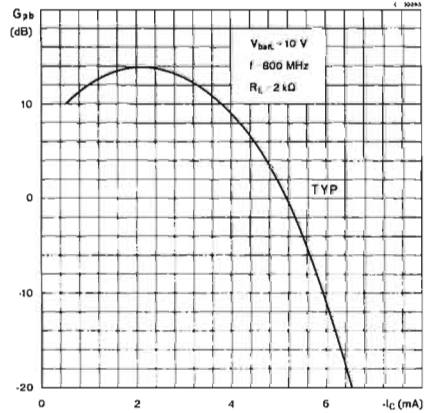


AF 239

Power gain

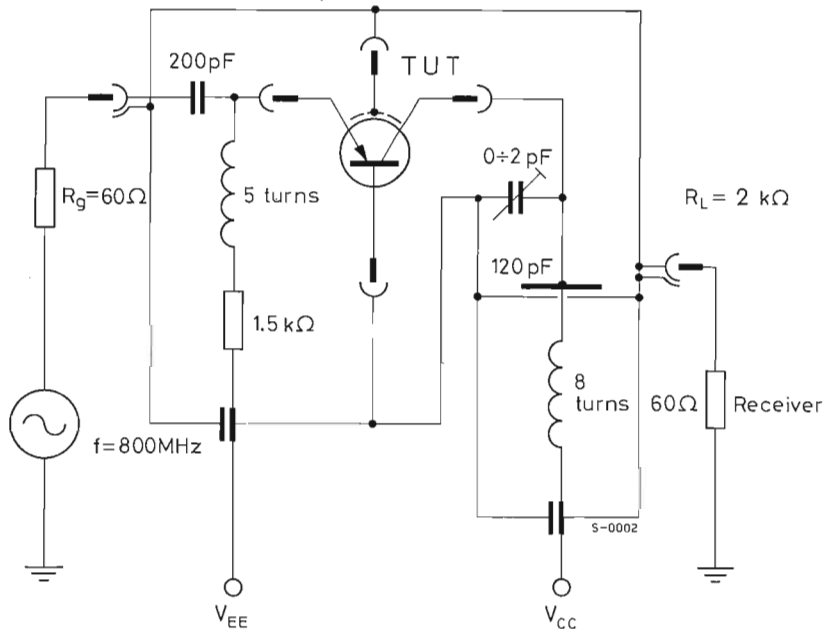


Power gain



TEST CIRCUIT

800 MHz transducer power gain G_{pb} and noise figure test circuit



AF 239S

GERMANIUM MESA PNP

UHF PREAMPLIFIER

The AF 239S is a germanium mesa PNP transistor in a Jedec TO-72 metal case. It is particularly designed as preamplifier, mixer and oscillator up to 900 MHz.

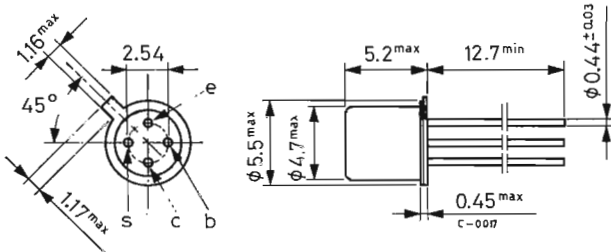
ABSOLUTE MAXIMUM RATINGS

V_{CES}	Collector-emitter voltage ($V_{BE} = 0$)	-20 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-15 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-0.3 V
I_C	Collector current	-10 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 45^\circ\text{C}$	60 mW
	at $T_{case} \leq 66^\circ\text{C}$	60 mW
T_{stg}	Storage temperature	-30 to 75 $^\circ\text{C}$
T_j	Junction temperature	90 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm

Shield lead connected to case



TO-72

AF 239S

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	400	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	750	°C/W

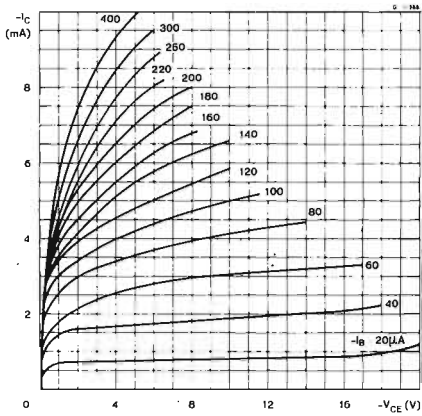
ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES}	Collector cutoff current ($V_{BE} = 0$)			-8	μA
I_{CEO}	Collector cutoff current ($I_B = 0$)			-500	μA
I_{EBO}	Emitter cutoff current ($I_C = 0$)			-100	μA
V_{BE}	Base-emitter voltage	$I_C = -2\text{ mA}$ $I_C = -5\text{ mA}$	$V_{CE} = -10\text{ V}$ $V_{CE} = -5\text{ V}$	-350 -400	mV mV
h_{FE}	DC current gain	$I_C = -2\text{ mA}$ $I_C = -5\text{ mA}$	$V_{CE} = -10\text{ V}$ $V_{CE} = -10\text{ V}$	10 30	— —
f_T	Transition frequency	$I_C = -2\text{ mA}$ $f = 100\text{ MHz}$	$V_{CE} = -10\text{ V}$	780	MHz
$-C_{re}$	Reverse capacitance	$I_C = -2\text{ mA}$ $f = 450\text{ kHz}$	$V_{CE} = -10\text{ V}$	0.2	pF
NF*	Noise figure	$I_C = -1\text{ mA}$ $R_g = 60\ \Omega$ $f = 800\text{ MHz}$	$V_{CE} = -10\text{ V}$	5	dB
G_{pb}^*	Power gain	$I_C = -2\text{ mA}$ $R_L = 2\text{ k}\Omega$ $f = 800\text{ MHz}$	$V_{CE} = -10\text{ V}$ $R_g = 60\ \Omega$	12.5 15	dB

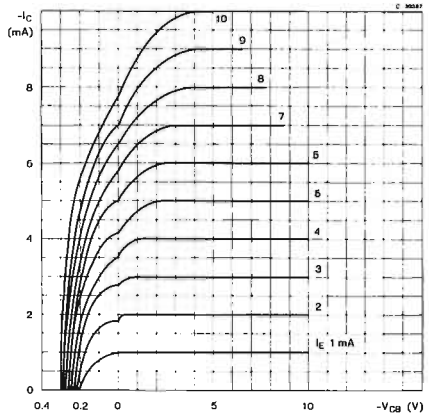
* See test circuit

AF 239S

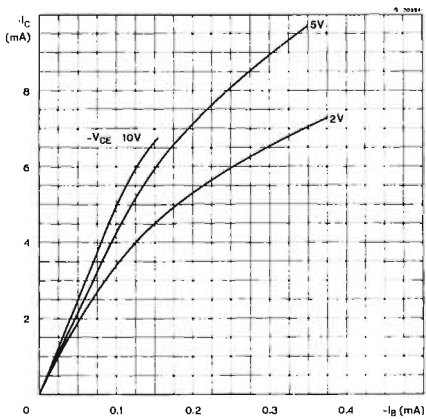
Typical output characteristics



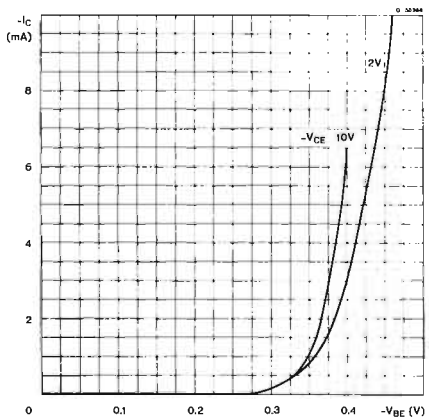
Typical output characteristics



Typical collector current

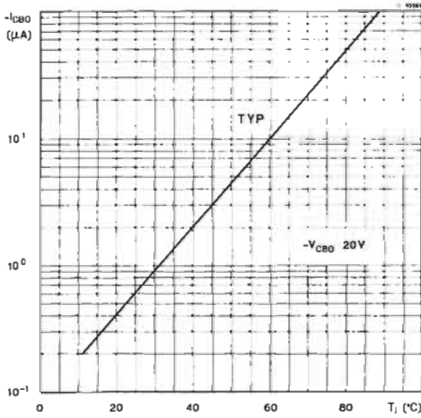


Typical DC transconductance

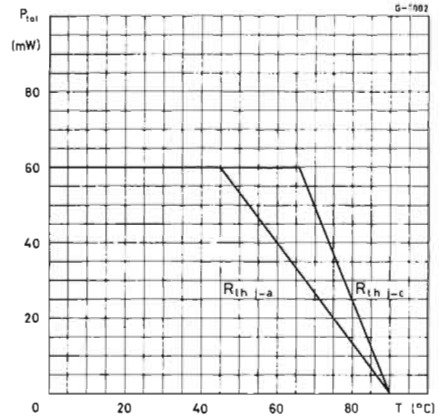


AF 239S

Collector cutoff current

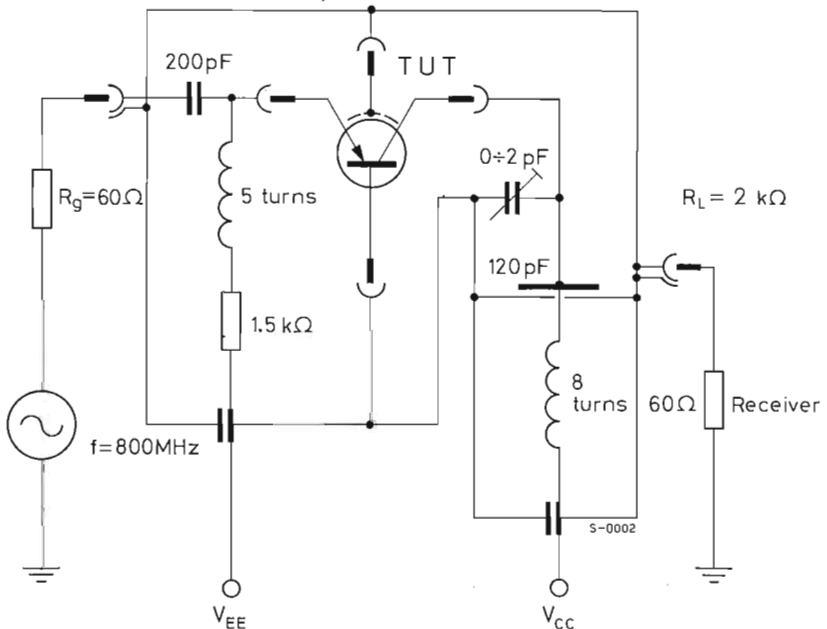


Power rating chart



TEST CIRCUIT

800 MHz transducer power gain G_{pb} and noise figure test circuit



SILICON TRANSISTORS

SILICON PLANAR NPN

LOW NOISE GENERAL PURPOSE AUDIO AMPLIFIERS

The BC 107, BC 108 and BC 109 are silicon planar epitaxial NPN transistors in TO-18 metal case. They are suitable for use in driver stages, low noise input stages and signal processing circuits of television receivers.

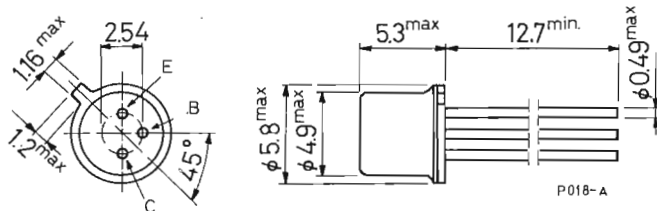
The complementary PNP types are respectively the BC 177, BC 178 and BC 179.

ABSOLUTE MAXIMUM RATINGS

		BC 107	BC 108	BC 109
V_{CB0}	Collector-base voltage ($I_E = 0$)	50 V	30 V	30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	45 V	20 V	20 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	6 V	5 V	5 V
I_C	Collector current	100 mA		
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.3 W		
	at $T_{case} \leq 25^\circ\text{C}$	0.75 W		
T_{stg}	Storage temperature	-55 to 175 °C		
T_j	Junction temperature	175 °C		

MECHANICAL DATA

Dimensions in mm



(sim. to TO-18)

BC 107
BC 108
BC 109

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	for BC 107 $V_{CB} = 40\text{ V}$ $V_{CB} = 40\text{ V}$ $T_{amb} = 150\text{ °C}$ for BC 108 - BC 109 $V_{CB} = 20\text{ V}$ $V_{CB} = 20\text{ V}$ $T_{amb} = 150\text{ °C}$			15 15 15 15	nA μ A nA μ A
$V_{(BR)CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\text{ }\mu\text{A}$ for BC 107 for BC 108 for BC 109			50 30 30	V V V
$V_{(BR)CEO}$ *Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 10\text{ mA}$ for BC 107 for BC 108 for BC 109			45 20 20	V V V
$V_{(BR)EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\text{ }\mu\text{A}$ for BC 107 for BC 108 for BC 109			6 5 5	V V V
$V_{CE(sat)}$ * Collector-emitter saturation voltage	$I_C = 10\text{ mA}$ $I_B = 0.5\text{ mA}$ $I_C = 100\text{ mA}$ $I_B = 5\text{ mA}$			70 250 200 600	mV mV
V_{BE} * Base-emitter voltage	$I_C = 2\text{ mA}$ $V_{CE} = 5\text{ V}$ $I_C = 10\text{ mA}$ $V_{CE} = 5\text{ V}$	550	650	700 770	mV mV
$V_{BE(sat)}$ * Base-emitter saturation voltage	$I_C = 10\text{ mA}$ $I_B = 0.5\text{ mA}$ $I_C = 100\text{ mA}$ $I_B = 5\text{ mA}$			750 900	mV mV

BC 107
BC 108
BC 109

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit		
h_{FE}^* DC current gain	$I_C = 2 \text{ mA}$ $V_{CE} = 5 \text{ V}$	for BC 107	110	230	450	—	
		for BC 107 Gr. A	110	180	220	—	
		for BC 107 Gr. B	200	290	450	—	
		for BC 108	110	350	800	—	
		for BC 108 Gr. A	110	180	220	—	
		for BC 108 Gr. B	200	290	450	—	
		for BC 108 Gr. C	420	520	800	—	
		for BC 109	200	350	800	—	
		for BC 109 Gr. B	200	290	450	—	
		for BC 109 Gr. C	420	520	800	—	
		$I_C = 10 \mu\text{A}$ $V_{CE} = 5 \text{ V}$	for BC 107		120		—
			for BC 107 Gr. A		90		—
			for BC 107 Gr. B	40	150		—
			for BC 108		120		—
			for BC 108 Gr. A		90		—
			for BC 108 Gr. B	40	150		—
for BC 108 Gr. C	100		270		—		
for BC 109			70	210	—		
for BC 109 Gr. B		40	150	—			
for BC 109 Gr. C		100	270	—			
h_{fe} Small signal current gain	$I_C = 2 \text{ mA}$ $V_{CE} = 5 \text{ V}$ $f = 1 \text{ kHz}$	for BC 107		250		—	
		for BC 107 Gr. A		190		—	
		for BC 107 Gr. B		300		—	
		for BC 108		370		—	
		for BC 108 Gr. A		190		—	
		for BC 108 Gr. B		300		—	
		for BC 108 Gr. C		500		—	
		for BC 109		370		—	
		for BC 109 Gr. B		300		—	
		for BC 109 Gr. C		550		—	
		$I_C = 10 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$			2		—
		C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 10 \text{ V}$ $f = 1 \text{ MHz}$		4	6	pF

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 107
BC 108
BC 109

ELECTRICAL CHARACTERISTICS (continued)

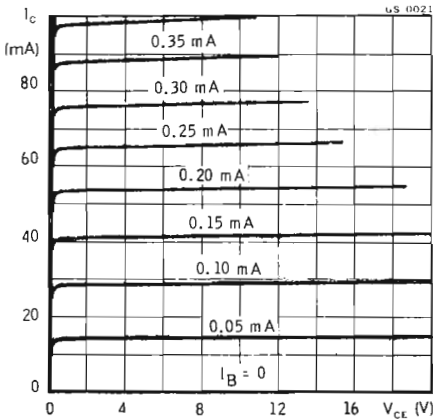
Parameter	Test conditions	Min.	Typ.	Max.	Unit
C_{EBO} Emitter-base capacitance	$I_C = 0$ $V_{EB} = 0.5$ V $f = 1$ MHz		11.5		pF
NF Noise figure	$I_C = 0.2$ mA $V_{CE} = 5$ V $R_g = 2$ k Ω $f = 1$ kHz $B = 200$ Hz for BC 107 for BC 108 for BC 109		2	10	dB
	$I_C = 0.2$ mA $V_{CE} = 5$ V $R_g = 2$ k Ω $f = 10$ Hz to 10 kHz $B = 15.7$ kHz for BC 109		1.5	4	dB
h_{ie} Input impedance	$I_C = 2$ mA $V_{CE} = 5$ V $f = 1$ kHz for BC 107		4		k Ω
	for BC 107 Gr. A		3		k Ω
	for BC 107 Gr. B		4.8		k Ω
	for BC 108		5.5		k Ω
	for BC 108 Gr. A		3		k Ω
	for BC 108 Gr. B		4.8		k Ω
	for BC 108 Gr. C		7		k Ω
	for BC 109		5.5		k Ω
	for BC 109 Gr. B		4.8		k Ω
	for BC 109 Gr. C		7		k Ω
h_{re} Reverse voltage ratio	$I_C = 2$ mA $V_{CE} = 5$ V $f = 1$ kHz for BC 107		2.2×10^{-4}		—
	for BC 107 Gr. A		1.7×10^{-4}		—
	for BC 107 Gr. B		2.7×10^{-4}		—
	for BC 108		3.1×10^{-4}		—
	for BC 108 Gr. A		1.7×10^{-4}		—
	for BC 108 Gr. B		2.7×10^{-4}		—
	for BC 108 Gr. C		3.8×10^{-4}		—
	for BC 109		3.1×10^{-4}		—
	for BC 109 Gr. B		2.7×10^{-4}		—
	for BC 109 Gr. C		3.8×10^{-4}		—

BC 107 BC 108 BC 109

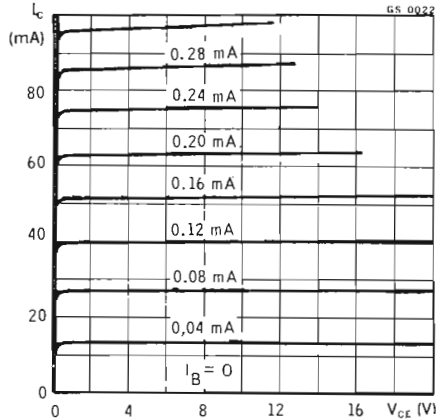
ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
h_{oe} Output admittance	$I_C = 2 \text{ mA}$ $V_{CE} = 5 \text{ V}$ $f = 1 \text{ kHz}$				
	for BC 107		20		μS
	for BC 107 Gr. A		13		μS
	for BC 107 Gr. B		26		μS
	for BC 108		30		μS
	for BC 108 Gr. A		13		μS
	for BC 108 Gr. B		26		μS
	for BC 108 Gr. C		34		μS
	for BC 109		30		μS
	for BC 109 Gr. B		26		μS
for BC 109 Gr. C		34		μS	

Typical output characteristics
(for **BC 107** only)

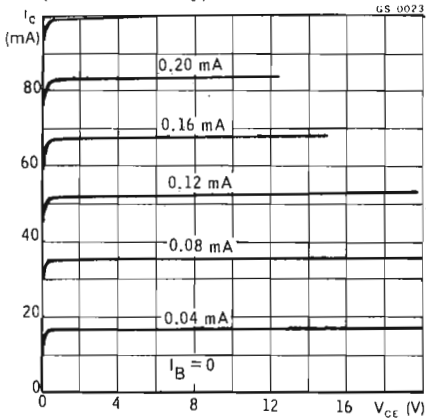


Typical output characteristics
(for **BC 108** only)

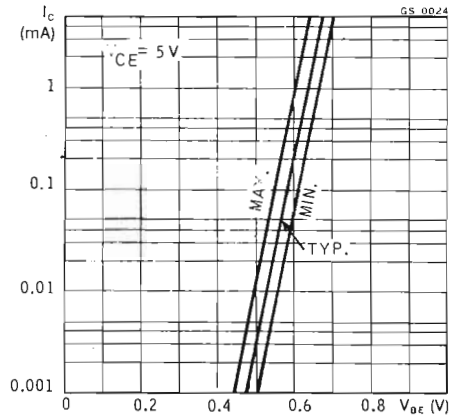


BC 107 BC 108 BC 109

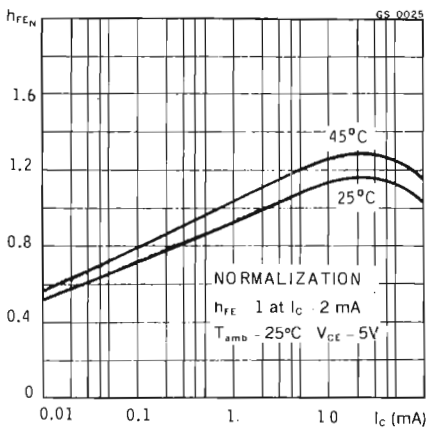
Typical output characteristics
(for BC 109 only)



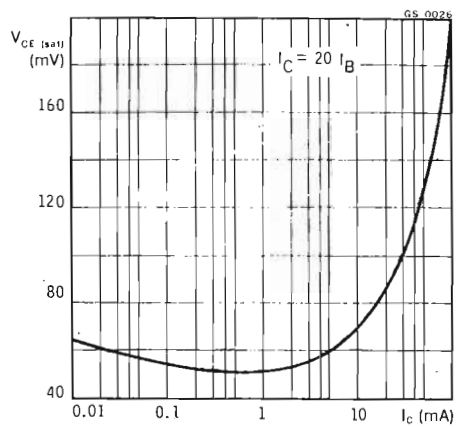
DC transconductance



DC normalized current gain

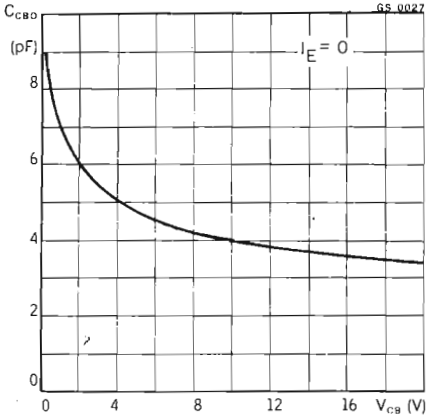


Collector-emitter saturation voltage

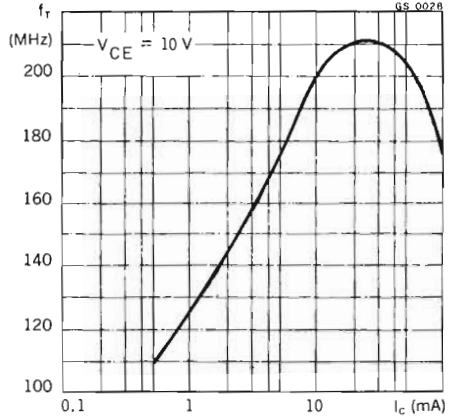


BC 107
BC 108
BC 109

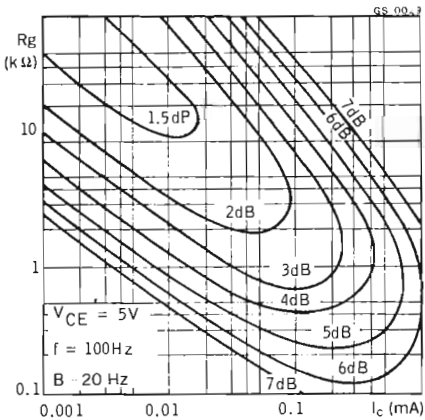
Collector-base capacitance



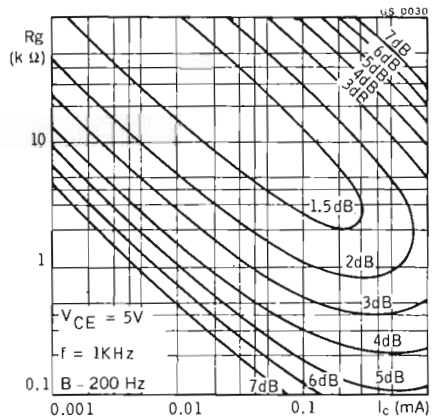
Transition frequency



Noise figure (for BC 109 only)

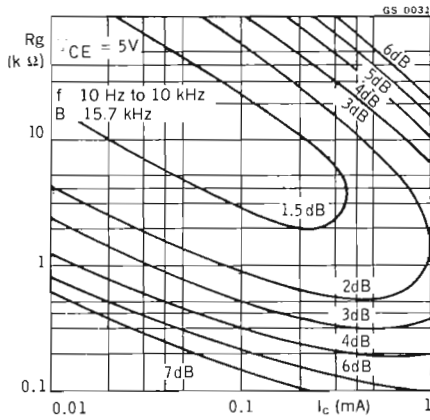


Noise figure (for BC 109 only)

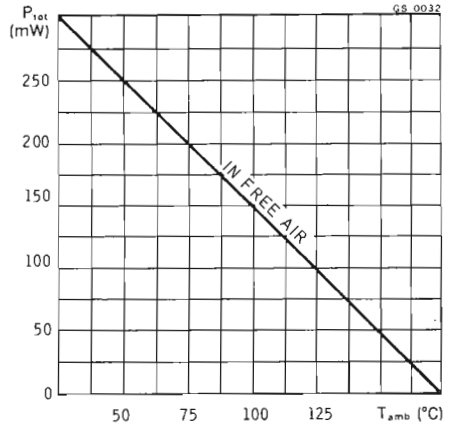


BC 107
BC 108
BC 109

Noise figure (for **BC 109** only)



Power rating chart



SILICON PLANAR NPN

HIGH GAIN, LOW NOISE AUDIO AMPLIFIERS

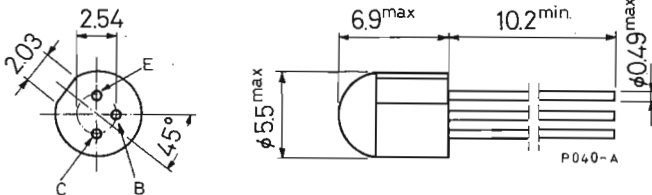
The BC 113 and BC 114 are silicon planar NPN transistors in TO-18 epoxy package. They are specifically designed for use in low-noise audio preamplifiers.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	30	V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	30	V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	6	V
I_C	Collector current	50	mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200	mW
	at $T_{case} \leq 25^\circ\text{C}$	500	mW
T_{stg}	Storage temperature	-55 to 125	$^\circ\text{C}$
T_j	Junction temperature	125	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BC 113 BC 114

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W

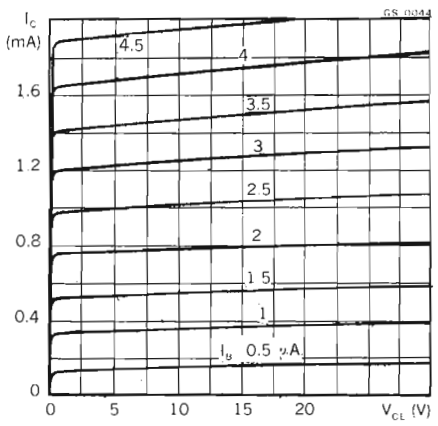
ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 20\text{ V}$ $V_{CE} = 20\text{ V}$ $T_{amb} = 65\text{ °C}$			50 5	nA μA
$V_{(BR)CEO}$ *Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 10\text{ mA}$	30			V
$V_{(BR)CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\text{ }\mu\text{A}$	30			V
$V_{(BR)EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\text{ }\mu\text{A}$	6			V
V_{BE} Base-emitter voltage	$I_C = 1\text{ mA}$ $V_{CE} = 5\text{ V}$	0.64	0.7		V
h_{FE} DC current gain	$I_C = 10\text{ }\mu\text{A}$ $V_{CE} = 5\text{ V}$ $I_C = 100\text{ }\mu\text{A}$ $V_{CE} = 5\text{ V}$ $I_C = 1\text{ mA}$ $V_{CE} = 5\text{ V}$ $I_C = 10\text{ mA}$ $V_{CE} = 5\text{ V}$ for BC 113 for BC 114	120 200 200	170 250 1000 400 400		— — — — —
f_T Transition frequency	$I_C = 1\text{ mA}$ $V_{CE} = 5\text{ V}$ for BC 113 for BC 114	60 70	100 100		MHz MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 5\text{ V}$	2.7	4		pF
NF Noise figure	$I_C = 10\text{ }\mu\text{A}$ $V_{CE} = 5\text{ V}$ $R_g = 10\text{ k}\Omega$ $f = 1\text{ kHz}$ $B = 200\text{ Hz}$ for BC 113 for BC 114		2.5 1.5	3	dB dB

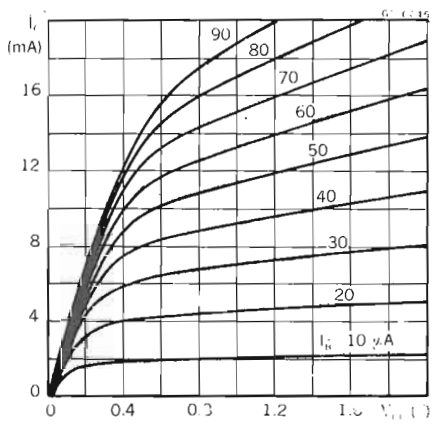
* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 113 BC 114

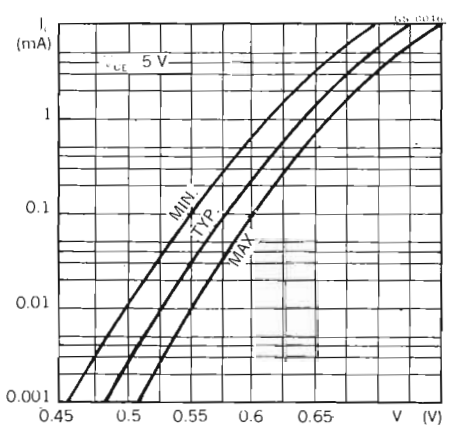
Typical output characteristics



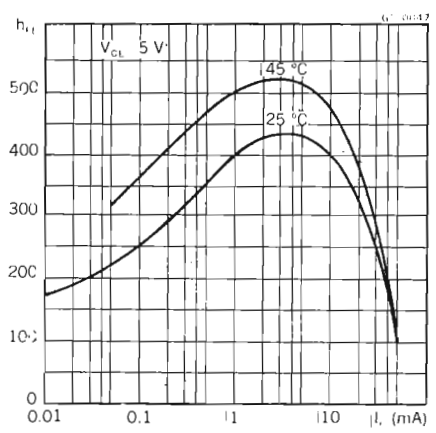
Typical output characteristics



DC transconductance

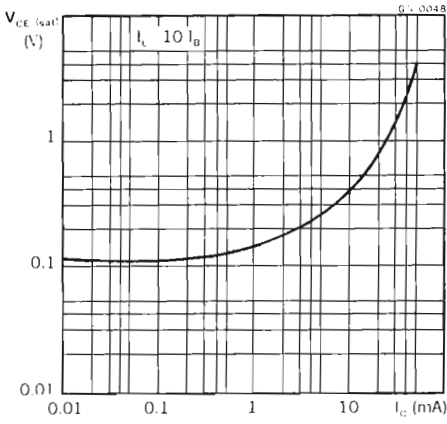


DC current gain

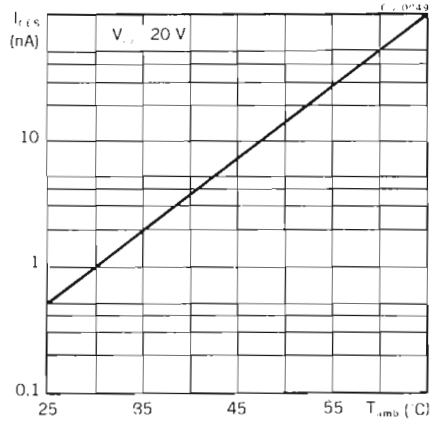


BC 113 BC 114

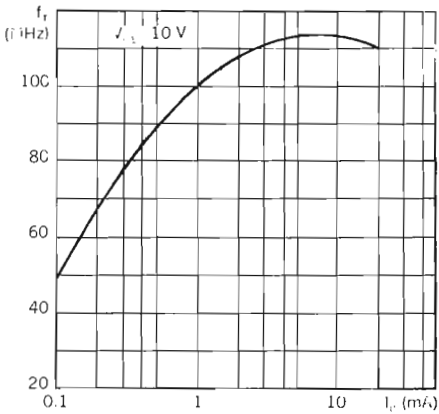
Collector-emitter saturation voltage



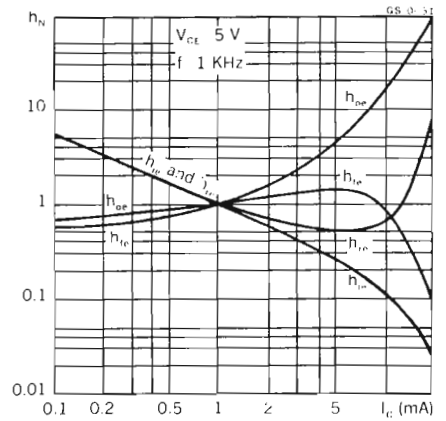
Collector cutoff current



Transition frequency

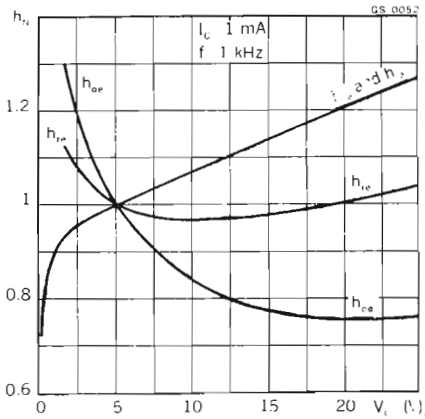


Typical normalized h parameters

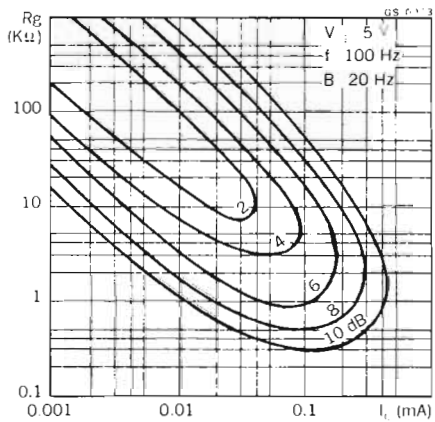


BC 113 BC 114

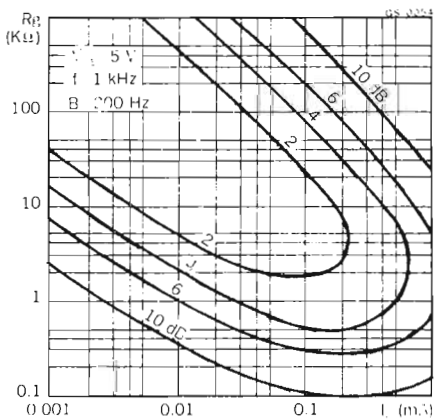
Typical normalized h parameters



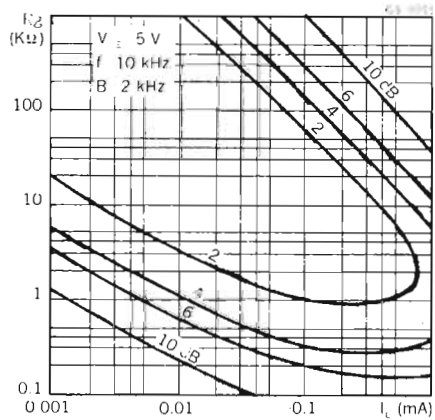
Noise figure (for BC 114 only)



Noise figure (for BC 114 only)

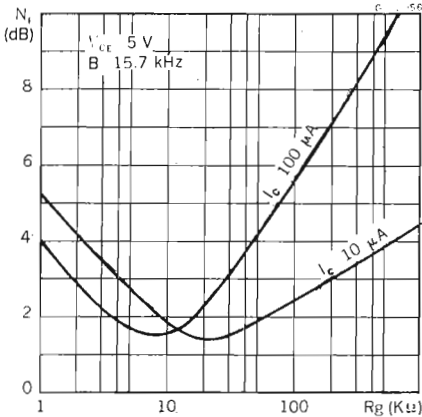


Noise figure (for BC 114 only)

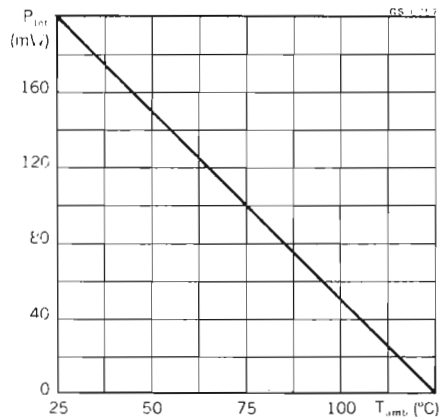


BC 113 BC 114

Noise figure (for BC 114 only)

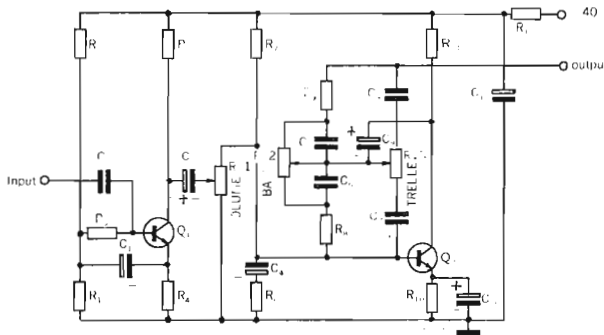


Power rating chart



TYPICAL APPLICATION

Audio preamplifier



$$R_1 = 470 \text{ k}\Omega$$

$$R_2 = 1.8 \text{ M}\Omega$$

$$R_3 = 18 \text{ k}\Omega$$

$$R_4 = 10 \text{ k}\Omega$$

$$R_5 = 33 \text{ k}\Omega$$

$$R_6 = 82 \text{ k}\Omega$$

$$R_7 = 270 \text{ k}\Omega$$

$$R_8 = 22 \text{ k}\Omega$$

$$R_9 = 2.2 \text{ k}\Omega$$

$$R_{10} = 8.2 \text{ k}\Omega$$

$$R_{11} = 22 \text{ k}\Omega$$

$$R_{12} = 27 \text{ k}\Omega$$

$$C_1 = 1 \mu\text{F} \quad 5 \text{ V}$$

$$C_2 = 0.2 \mu\text{F} \quad 12 \text{ V}$$

$$C_3 = 5 \mu\text{F} \quad 15 \text{ V}$$

$$C_4 = 5 \mu\text{F} \quad 5 \text{ V}$$

$$C_5 = 0.015 \mu\text{F}$$

$$C_6 = 0.15 \mu\text{F}$$

$$C_7 = 0.0039 \mu\text{F}$$

$$C_8 = 0.039 \mu\text{F}$$

$$C_9 = 5 \mu\text{F} \quad 10 \text{ V}$$

$$C_{10} = 100 \mu\text{F} \quad 5 \text{ V}$$

$$C_{11} = 100 \mu\text{F} \quad 50 \text{ V}$$

$$RV_1 = 25 \text{ k}\Omega \text{ Linear}$$

$$RV_2 = 100 \text{ k}\Omega \text{ Anti-log}$$

$$RV_3 = 50 \text{ k}\Omega \text{ Anti-log}$$

$$Q_1 = \text{BC 114}$$

$$Q_2 = \text{BC 113}$$

Overall performance

Sensitivity	< 150 mV for 100 μ A, output into 200 Ω load
Input impedance	1.5 M Ω
Signal/noise ratio	75 dB
Maximum input voltage	2 V
Bass boost	+ 10 dB (60 Hz)
Bass cut	- 15 dB (60 Hz)
Treble boost	+ 10 dB (12 kHz)
Treble cut	- 13 dB (12 kHz)

BC 115

SILICON PLANAR NPN

AUDIO DRIVER

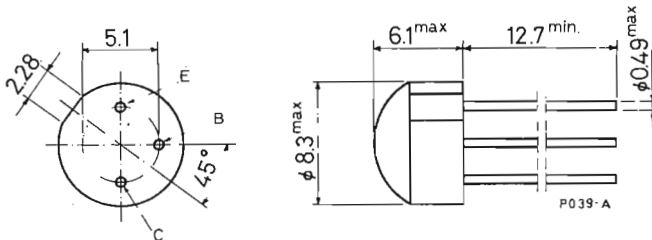
The BC 115 is a silicon planar epitaxial NPN transistor in a TO-39 epoxy package. It is particularly suited for use in audio driver circuits.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	30 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	5 V
I_C	Collector current	200 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.3 W
	at $T_{case} \leq 25^\circ\text{C}$	0.8 W
T_{stg}	Storage temperature	-55 to 125 °C
T_j	Junction temperature	125 °C

MECHANICAL DATA

Dimensions in mm



TO-39 epoxy

BC 115

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	125	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	330	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 20\text{ V}$ $V_{CB} = 20\text{ V}$ $T_{amb} = 65\text{ °C}$			100 5	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$	40			V
$V_{(BR)\ CEO}^*$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 30\text{ mA}$	30			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_C = 10\ \mu\text{A}$	5			V
$V_{CE(sat)}^*$ Collector-emitter saturation voltage	$I_C = 100\text{ mA}$ $I_B = 10\text{ mA}$		0.4	1	V
V_{BE} Base-emitter voltage	$I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 100\text{ mA}$ $V_{CE} = 10\text{ V}$		0.65 0.75		V V
$V_{BE(sat)}^*$ Base-emitter saturation voltage	$I_C = 100\text{ mA}$ $I_B = 10\text{ mA}$		0.8	0.9	V
h_{FE}^* DC current gain	$I_C = 100\ \mu\text{A}$ $V_{CE} = 10\text{ V}$ $I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 100\text{ mA}$ $V_{CE} = 10\text{ V}$			95 50 145 100 170 400 50 150	— — — —
f_T Transition frequency	$I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$		80		MHz

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 115

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
C_{CBO}	Collector-base capacitance	$I_E = 0$ $f = 1 \text{ MHz}$ $V_{CB} = 10 \text{ V}$		12	25	pF
h_{ie}	Input impedance	$I_C = 10 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = 10 \text{ V}$		550		Ω
h_{re}	Voltage feedback ratio	$I_C = 10 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = 10 \text{ V}$		0.9×10^{-4}		—
h_{oe}	Output admittance	$I_C = 10 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = 10 \text{ V}$		50		μS

BC 116A

SILICON PLANAR PNP

GENERAL PURPOSE TRANSISTOR

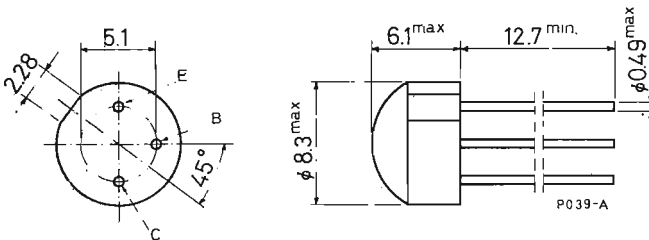
The BC 116A is a silicon planar epitaxial PNP transistor in a TO-39 epoxy package. It is designed as general purpose device for application over a wide range of collector current.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-45 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-5 V
I_C	Collector current	-500 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.3 W
	at $T_{case} \leq 25^\circ\text{C}$	0.8 W

MECHANICAL DATA

Dimensions in mm



TO-39 epoxy

BC 116A

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	125	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	330	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -20\text{ V}$ $V_{CB} = -20\text{ V}$ $T_{amb} = 75\text{ °C}$			-100 -10	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$	-45			V
$V_{(BR)\ CEO}^*$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -10\text{ mA}$	-40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_C = -10\ \mu\text{A}$	-5			V
$V_{CE(sat)}^*$ Collector-emitter saturation voltage	$I_C = -50\text{ mA}$ $I_B = -5\text{ mA}$ $I_C = -150\text{ mA}$ $I_B = -15\text{ mA}$			-0.25 -0.40	V V
V_{BE}^* Base-emitter voltage	$I_C = -10\text{ mA}$ $V_{CE} = -10\text{ V}$ $I_C = -50\text{ mA}$ $V_{CE} = -1\text{ V}$	-0.70		-0.75 -1	V V
$V_{BE(sat)}^*$ Base-emitter saturation voltage	$I_C = -50\text{ mA}$ $I_B = -5\text{ mA}$ $I_C = -150\text{ mA}$ $I_B = -15\text{ mA}$	-0.80		-1 -1.3	V V
h_{FE} DC current gain	$I_C = -100\ \mu\text{A}$ $V_{CE} = -10\text{ V}$ $I_C = -10\text{ mA}$ $V_{CE} = -1\text{ V}$ $I_C = -50\text{ mA}$ $V_{CE} = -1\text{ V}$ $I_C = -150\text{ mA}$ $V_{CE} = -10\text{ V}$	30	90		— — — —
f_T Transition frequency	$I_C = -30\text{ mA}$ $V_{CE} = -10\text{ V}$	130	200		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -10\text{ V}$ $f = 1\text{ MHz}$			5 10	pF

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 119

SILICON PLANAR NPN

AUDIO OUTPUT AMPLIFIER

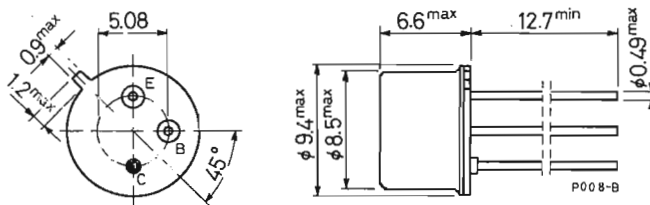
The BC 119 is a silicon planar epitaxial NPN transistor in a TO-39 metal case. It is suitable for 1 W class "A" and up to 6 W class "B" audio output stages and is available as a pair 2 BC 119.

ABSOLUTE MAXIMUM RATINGS

V_{CB0}	Collector-base voltage ($I_E = 0$)	60 V
V_{CE0}	Collector-emitter voltage ($I_B = 0$)	30 V
V_{EB0}	Emitter-base voltage ($I_C = 0$)	5 V
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.8 W
	at $T_{case} \leq 25^\circ\text{C}$	5 W
	at $T_{case} \leq 100^\circ\text{C}$	2.8 W
T_{stg}	Storage temperature	-55 to 200 °C
T_j	Junction temperature	200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 119

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	35	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	220	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 40\text{ V}$ $V_{CB} = 40\text{ V}$ $T_{amb} = 150\text{ °C}$			100 20	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$	60			V
$V_{CEO(sus)}$ * Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 30\text{ mA}$	30			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\ \mu\text{A}$	5			V
$V_{CE(sat)}$ * Collector-emitter saturation voltage	$I_C = 150\text{ mA}$ $I_B = 15\text{ mA}$ $I_C = 500\text{ mA}$ $I_B = 50\text{ mA}$ $I_C = 1\text{ A}$ $I_B = 100\text{ mA}$		0.15 0.4 0.8	0.35 1.1 1.5	V V V
V_{BE} * Base-emitter voltage	$I_C = 500\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 150\text{ mA}$ $V_{CE} = 1\text{ V}$		1 0.85	1.8 1	V V
$V_{BE(sat)}$ * Base-emitter saturator. voltage	$I_C = 150\text{ mA}$ $I_B = 15\text{ mA}$ $I_C = 1\text{ A}$ $I_B = 0.1\text{ A}$		0.9 1.4	1.2 2	V V
h_{FE} * DC current gain	$I_C = 50\text{ mA}$ $V_{CE} = 1\text{ V}$ $I_C = 150\text{ mA}$ $V_{CE} = 1\text{ V}$ $I_C = 500\text{ mA}$ $V_{CE} = 10\text{ V}$	40 40 25	100 90 60		— — —
h_{FE1}/h_{FE2} Matched pair	$I_C = 300\text{ mA}$ $V_{CE} = 5\text{ V}$			1.4	—
f_T Transition frequency	$I_C = 50\text{ mA}$ $V_{CE} = 10\text{ V}$	40			MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 10\text{ V}$		12	25	pF

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

SILICON PLANAR NPN

BC 125 BC 125B

AUDIO DRIVERS

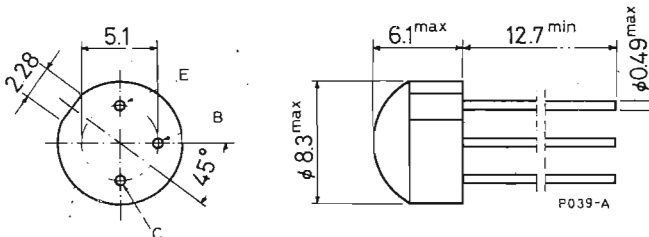
The BC 125 and BC 125 B are silicon planar epitaxial NPN transistors in TO-39 epoxy package. They are designed for use as audio drivers.

ABSOLUTE MAXIMUM RATINGS

		BC 125	BC 125 B
V_{CBO}	Collector-base voltage ($I_E = 0$)	50 V	60 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	5 V	6 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)		30 V
I_C	Collector current	0.5 A	
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.3 W	
	at $T_{case} \leq 25^\circ\text{C}$		0.8 W
T_{stg}	Storage temperature	-55 to 125 °C	
T_j	Junction temperature	125 °C	

MECHANICAL DATA

Dimensions in mm



TO-39 epoxy

BC 125

BC 125B

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	125	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	330	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{°C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	for BC 125 $V_{CB} = 20\text{ V}$ $V_{CB} = 20\text{ V}$ $T_{amb} = 75\text{°C}$ for BC 125 B $V_{CB} = 40\text{ V}$ $V_{CB} = 40\text{ V}$ $T_{amb} = 75\text{°C}$		0.5	100 20	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\ \mu\text{A}$ for BC 125 for BC 125 B	50 60			V V
$V_{CEO(sus)}$ * Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 30\text{ mA}$	30			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_C = 10\ \mu\text{A}$ for BC 125 for BC 125 B	5 6			V V
$V_{CE(sat)}$ * Collector-emitter saturation voltage	for BC 125 $I_C = 150\text{ mA}$ $I_B = 15\text{ mA}$ for BC 125 B $I_C = 150\text{ mA}$ $I_B = 15\text{ mA}$ $I_C = 500\text{ mA}$ $I_B = 50\text{ mA}$		0.2	2.5 0.25 0.8	V V V

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 125 BC 125B

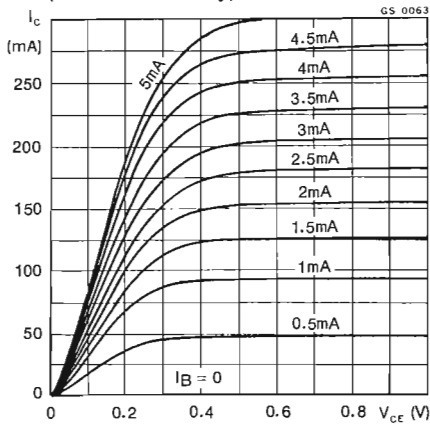
ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{BE}^* Base-emitter voltage	$I_C = 50 \text{ mA}$ $V_{CE} = 1 \text{ V}$	0.72			V
$V_{BE(sat)}^*$ Base-emitter saturation voltage	for BC 125 $I_C = 150 \text{ mA}$ $I_B = 15 \text{ mA}$	1 1.3			V
	for BC 125 B $I_C = 150 \text{ mA}$ $I_B = 15 \text{ mA}$	0.87 1			V
	$I_C = 500 \text{ mA}$ $I_B = 50 \text{ mA}$	1.1 1.3			V
h_{FE}^* DC current gain	for BC 125 $I_C = 1 \text{ mA}$ $V_{CE} = 1 \text{ V}$	50			—
	$I_C = 10 \text{ mA}$ $V_{CE} = 1 \text{ V}$	70			—
	$I_C = 50 \text{ mA}$ $V_{CE} = 1 \text{ V}$	30 75			—
	$I_C = 150 \text{ mA}$ $V_{CE} = 1 \text{ V}$	30 60			—
	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$	25 55			—
	$I_C = 10 \text{ mA}$ $V_{CE} = 10 \text{ V}$	30 75			—
	for BC 125 B $I_C = 1 \text{ mA}$ $V_{CE} = 1 \text{ V}$	85			—
	$I_C = 10 \text{ mA}$ $V_{CE} = 1 \text{ V}$	45 100			—
	$I_C = 50 \text{ mA}$ $V_{CE} = 1 \text{ V}$	95			—
	$I_C = 150 \text{ mA}$ $V_{CE} = 1 \text{ V}$	40 80 120			—
$I_C = 500 \text{ mA}$ $V_{CE} = 10 \text{ V}$	70			—	
f_T Transition frequency	$I_C = 50 \text{ mA}$ $V_{CE} = 10 \text{ V}$	200	350		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 10 \text{ V}$ $f = 1 \text{ MHz}$	6 12			pF
		5 8			pF
	for BC 125				
	for BC 125 B				

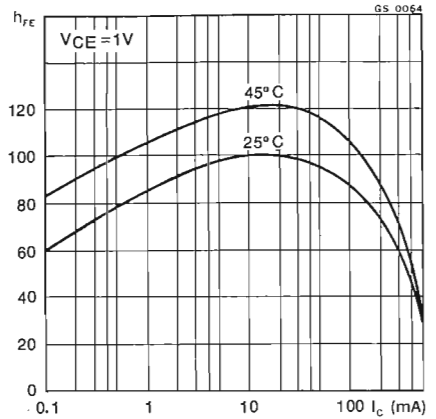
* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 125 BC 125B

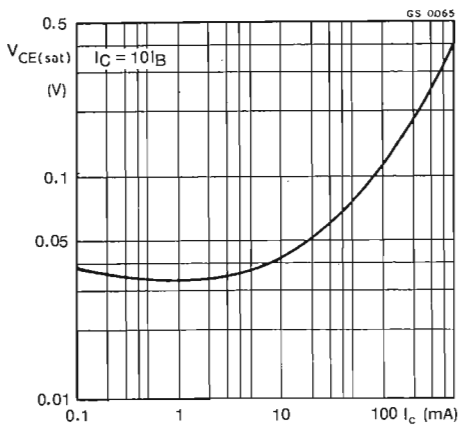
Typical output characteristics
(for BC 125 B only)



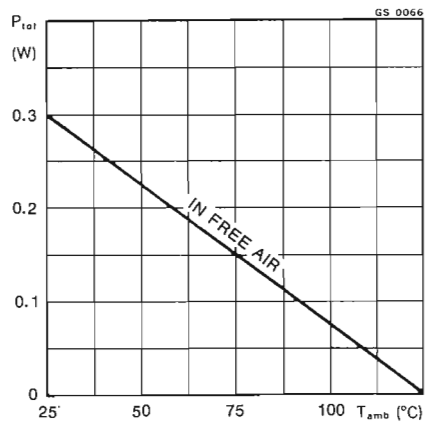
DC current gain (for BC 125 B only)



Collector-emitter saturation voltage



Power rating chart



BC 126

SILICON PLANAR PNP

AUDIO DRIVER

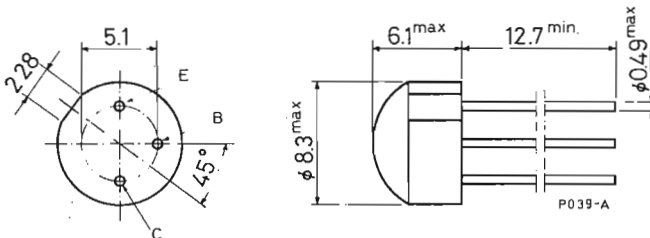
The BC 126 is a silicon planar epitaxial PNP transistor in a TO-39 epoxy package. It is designed for audio driver applications. The complementary NPN type is the BC 125.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-35 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-30 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-5 V
I_C	Collector current	-0.5 A
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$ at $T_{case} \leq 25^\circ\text{C}$	0.3 W 0.8 W
T_{stg}	Storage temperature	-55 to 125 °C
T_j	Junction temperature	125 °C

MECHANICAL DATA

Dimensions in mm



TO-39 epoxy

BC 126

THERMAL DATA

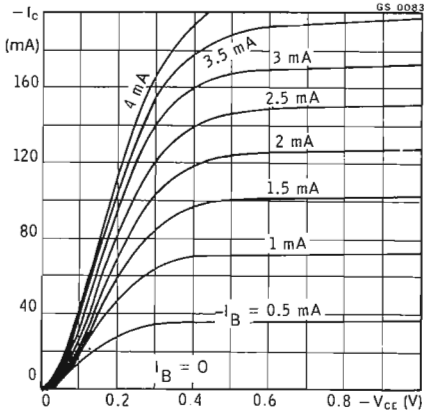
$R_{th\ j-case}$	Thermal resistance junction-case	max	125	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	330	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{°C}$ unless otherwise specified)

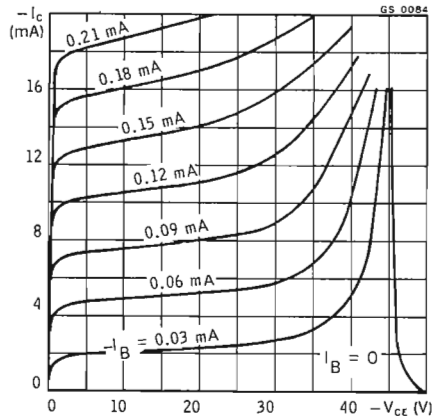
Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -20\text{ V}$ $V_{CB} = -20\text{ V}$ $T_{amb} = 75\text{°C}$			-100 -20	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$	-35			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -10\ \text{mA}$	-30			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-5			V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = -50\ \text{mA}$ $I_B = -5\ \text{mA}$ $I_C = -150\ \text{mA}$ $I_B = -15\ \text{mA}$			-0.25 -0.50	V V
V_{BE} Base-emitter voltage	$I_C = -50\ \text{mA}$ $V_{CE} = -1\ \text{V}$	-0.75		-1	V
$V_{BE(sat)}$ Base-emitter saturation voltage	$I_C = -150\ \text{mA}$ $I_B = -15\ \text{mA}$ $I_C = -50\ \text{mA}$ $I_B = -5\ \text{mA}$			-1 -0.8	V V
h_{FE} DC current gain	$I_C = -50\ \text{mA}$ $V_{CE} = -1\ \text{V}$ $I_C = -150\ \text{mA}$ $V_{CE} = -1\ \text{V}$	30	80	120	— —
f_T Transition frequency	$I_C = -50\ \text{mA}$ $V_{CE} = -20\ \text{V}$		200		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -10\ \text{V}$ $f = 1\ \text{MHz}$			5	pF

BC 126

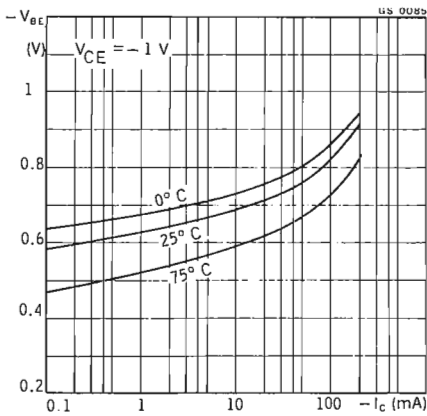
Typical output characteristics



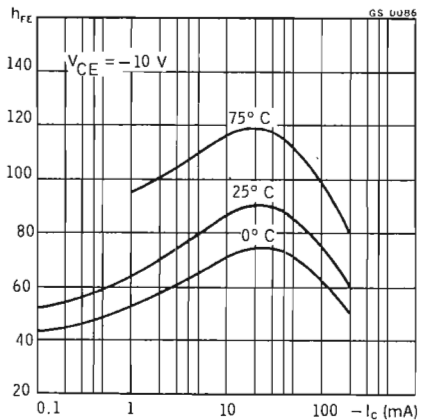
Typical output characteristics



Typical DC transconductance

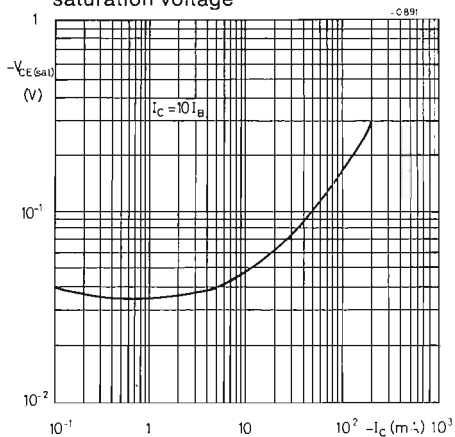


Typical DC current gain

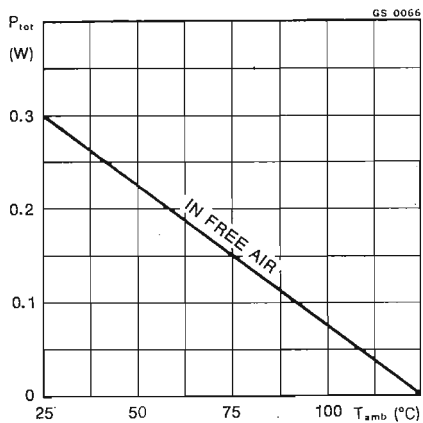


BC 126

Typical collector emitter saturation voltage



Power rating chart



SILICON PLANAR NPN

AUDIO AMPLIFIER

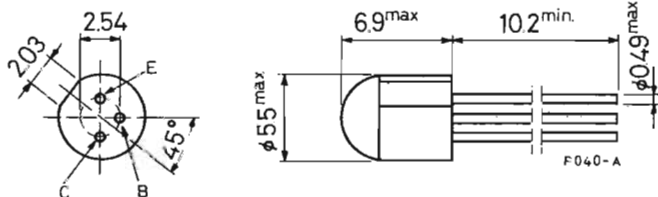
The BC 132 is a silicon planar NPN transistor in a TO-18 epoxy package. It is suitable for low level audio stages and direct coupled circuits.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	25 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	6 V
I_C	Collector current	20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.2 W
	at $T_{case} \leq 25^\circ\text{C}$	0.5 W
T_{sig}	Storage temperature	-55 to 125 $^\circ\text{C}$
T_J	Junction temperature	125 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BC 132

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200 °C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500 °C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 5\text{ V}$ $V_{CB} = 5\text{ V}$ $T_{amb} = 65\text{ °C}$			100 3	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$	30			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 10\text{ mA}$	25			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\ \mu\text{A}$	6			V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = 1\text{ mA}$ $I_B = 0.1\text{ mA}$			0.35	V
h_{FE} DC current gain	$I_C = 50\ \mu\text{A}$ $V_{CE} = 10\text{ V}$ $I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$	60	50	300	— —
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 5\text{ V}$	2.2		4	pF

BC 139

SILICON PLANAR PNP

AUDIO OUTPUT AMPLIFIER

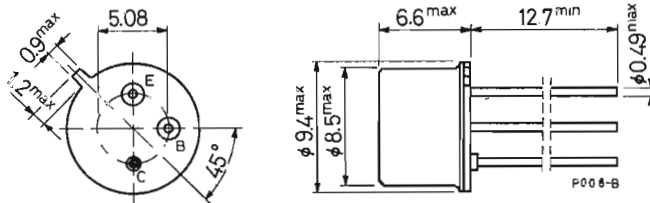
The BC 139 is a silicon planar epitaxial PNP transistor in a TO-39 metal case. It is particularly designed for use in audio output and driver stages. The complementary NPN type is the BC 119.

ABSOLUTE MAXIMUM RATINGS

V_{CB0}	Collector-base voltage ($I_E = 0$)	-40	V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-40	V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-5	V
I_C	Collector current	-0.5	A
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$ at $T_{case} \leq 25^\circ\text{C}$	0.7 3	W W
T_{stg}	Storage temperature	-55 to 200	$^\circ\text{C}$
T_j	Junction temperature	200	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 139

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	58	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	250	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -30\text{ V}$ $V_{CB} = -30\text{ V } T_{amb} = 75\text{ °C}$			-100 -50	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$	-40			V
$V_{(BR)\ CEO}$ *Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -10\text{ mA}$	-40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-5			V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = -300\text{ mA}$ $I_B = -30\text{ mA}$	-0.45	-0.8		V
	$I_C = -500\text{ mA}$ $I_B = -50\text{ mA}$		-1		V
V_{BE} Base-emitter voltage	$I_C = -10\text{ mA}$ $V_{CE} = -10\text{ V}$		-0.7		V
	$I_C = -100\text{ mA}$ $V_{CE} = -10\text{ V}$		-0.77		V
	$I_C = -300\text{ mA}$ $V_{CE} = -10\text{ V}$		-0.97		V
	$I_C = -300\text{ mA}$ $V_{CE} = -1\text{ V}$		-0.97		V

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

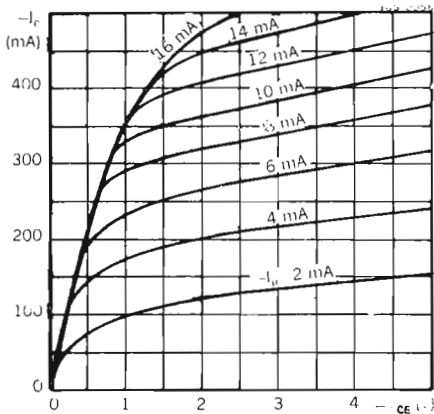
BC 139

ELECTRICAL CHARACTERISTICS (continued)

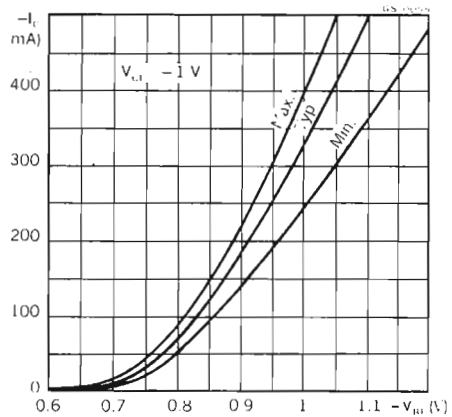
Parameter	Test conditions	Min.	Typ.	Max.	Unit
h_{FE}^* DC current gain	$I_C = -10 \text{ mA}$ $V_{CE} = -10 \text{ V}$		90		---
	$I_C = -100 \text{ mA}$ $V_{CE} = -10 \text{ V}$	40	90		---
	$I_C = -150 \text{ mA}$ $V_{CE} = -1 \text{ V}$		45		---
	$I_C = -300 \text{ mA}$ $V_{CE} = -1 \text{ V}$	20	35		---
f_T Transition frequency	$I_C = -50 \text{ mA}$ $V_{CE} = -10 \text{ V}$		200		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -10 \text{ V}$ $f = 1 \text{ MHz}$		6		pF

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

Typical output characteristics

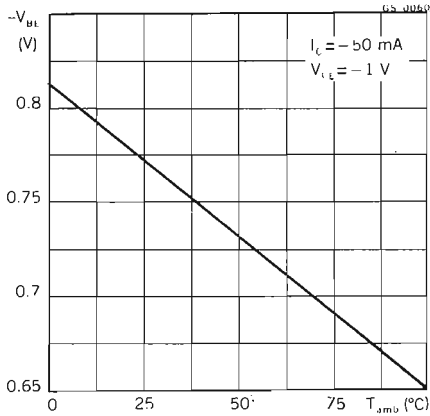


DC transconductance

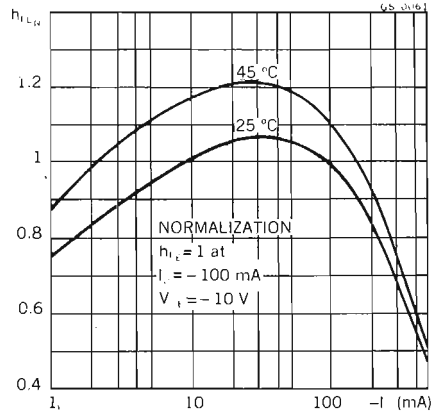


BC 139

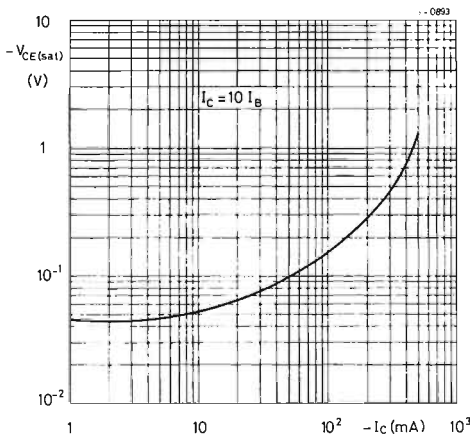
Base-emitter voltage



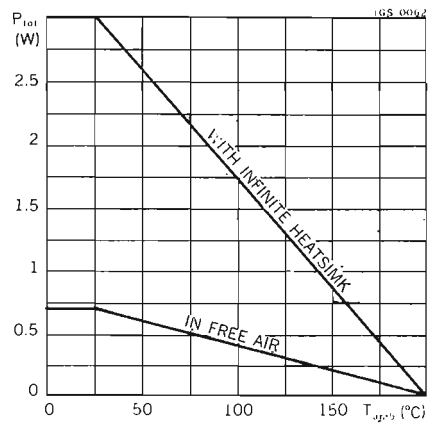
DC normalized current gain



Collector-emitter saturation voltage



Power rating chart



SILICON PLANAR NPN

BC 140 BC 141

GENERAL PURPOSE TRANSISTORS

The BC 140 and BC 141 are silicon planar epitaxial NPN transistors in TO-39 metal case. They are particularly designed for audio amplifiers and switching applications up to 1 A.

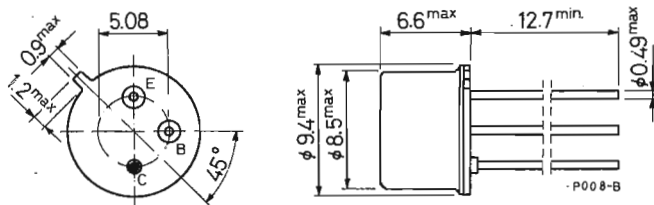
The complementary PNP types are the BC 160 and BC 161.

ABSOLUTE MAXIMUM RATINGS

		BC 140	BC 141
V_{CBO}	Collector-base voltage ($I_E = 0$)	60 V	80 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	40 V	60 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)		7 V
I_C	Collector current		1 A
I_B	Base current		0.1 A
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$		0.8 W
	at $T_{case} \leq 25^\circ\text{C}$		4 W
T_{stg}	Storage temperature	-55 to 200 °C	
T_j	Junction temperature	200 °C	

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 140 BC 141

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	44	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	220	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 50\text{ V}$ $V_{CB} = 50\text{ V}$ $T_{amb} = 150\text{ °C}$			200 200	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$ for BC 140 for BC 141	60 80			V V
$V_{(BR)\ CEO}^*$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 10\text{ mA}$ for BC 140 for BC 141	40 60			V V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\ \mu\text{A}$	7			V
$V_{CE(sat)}^*$ Collector-emitter saturation voltage	$I_C = 100\text{ mA}$ $I_B = 10\text{ mA}$ $I_C = 500\text{ mA}$ $I_B = 50\text{ mA}$ $I_C = 1\text{ A}$ $I_B = 0.1\text{ A}$		0.1 0.35 0.6		V V V
V_{BE}^* Base-emitter voltage	$I_C = 1\text{ A}$ $V_{CE} = 1\text{ V}$	1.25	1.6		V

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 140 BC 141

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
h_{FE}^* DC current gain	$I_C = 100 \mu A$ $V_{CE} = 1 V$ for BC 140-141 for BC 140-141 Gr. 6 for BC 140-141 Gr. 10		75		—
			28		—
			40		—
	$I_C = 100 mA$ $V_{CE} = 1 V$ for BC 140-141 for BC 140-141 Gr. 6 for BC 140-141 Gr. 10	40	140	250	—
		40	63	100	—
		63	100	160	—
$I_C = 1 A$ $V_{CE} = 1 V$ for BC 140-141 for BC 140-141 Gr. 6 for BC 140-141 Gr. 10		26		—	
		15		—	
		20		—	
h_{FE1}/h_{FE2} Matched pair ratio	$I_C = 100 mA$ $V_{CE} = 1 V$		1.25		—
f_T Transition frequency	$I_C = 50 mA$ $V_{CE} = 10 V$	50			MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 20 V$ $f = 1 MHz$		12		pF
t_{on} Turn-on time	$I_C = 100 mA$ $I_{B1} = 5 mA$			250	ns
t_{off} Turn-off time	$I_C = 100 mA$ $I_{B1} = I_{B2} = 5 mA$			850	ns

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

SILICON PLANAR PNP

BC 153 BC 154

LOW-NOISE AUDIO AMPLIFIERS

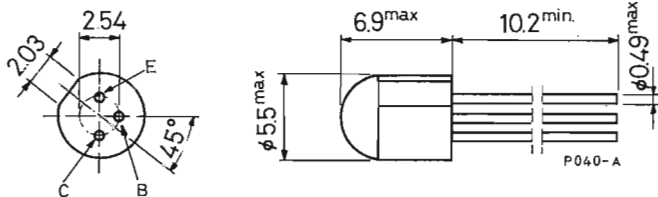
The BC 153 and BC 154 are silicon planar epitaxial PNP transistors in TO-18 epoxy package. They are specifically designed for use in low-noise audio preamplifiers.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-5 V
I_C	Collector current	-100 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.2 W
	at $T_{case} \leq 25^\circ\text{C}$	0.5 W
T_{stg}	Storage temperature	-55 to 125 $^\circ\text{C}$
T_j	Junction temperature	125 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BC 153 BC 154

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -30\text{ V}$			-50	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$	-40			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -5\text{ mA}$	-40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-5			V
$V_{CE\ (sat)}$ Collector-emitter saturation voltage	$I_C = -10\text{ mA}$ $I_B = -0.5\text{ mA}$			-0.25	V
h_{FE} DC current gain	$I_C = -10\ \mu\text{A}$ $V_{CE} = -5\text{ V}$ for BC 153 for BC 154 $I_C = -100\ \mu\text{A}$ $V_{CE} = -5\text{ V}$ for BC 153 for BC 154 $I_C = -1\text{ mA}$ $V_{CE} = -5\text{ V}$ for BC 153 for BC 154 $I_C = -10\text{ mA}$ $V_{CE} = -5\text{ V}$ for BC 153 for BC 154		115 190 50 160 50 160 50 160	125 215 135 230 135 225	— — — — — — — —
f_T Transition frequency	$I_C = -1\text{ mA}$ $V_{CE} = -5\text{ V}$		70		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -5\text{ V}$ $f = 1\text{ MHz}$		4		pF

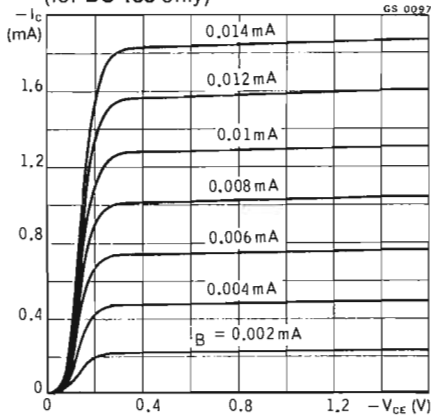
BC 153 BC 154

ELECTRICAL CHARACTERISTICS (continued)

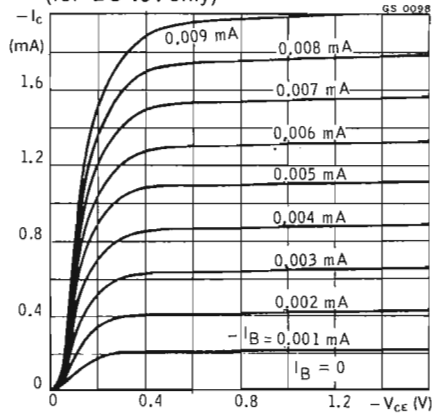
Parameter	Test conditions	Min.	Typ.	Max.	Unit
NF Noise figure	$I_C = -20 \mu\text{A}$ $V_{CE} = -5 \text{ V}$ $R_g = 10 \text{ k}\Omega$ $f = 1 \text{ kHz}$ $B = 200 \text{ Hz}$ for BC 153 for BC 154		1		dB
		0.75		2.5	dB
	$I_C = -250 \mu\text{A}$ $V_{CE} = -5 \text{ V}$ $R_g = 1 \text{ k}\Omega$ $f = 1 \text{ kHz}$ $B = 200 \text{ Hz}$ for BC 153 for BC 154		1		dB
		0.75		2.5	dB
h_{ie} Input impedance	$I_C = -1 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$ for BC 153 for BC 154		5.2		$\text{k}\Omega$
			7.1		$\text{k}\Omega$
h_{re} Reverse voltage ratio	$I_C = -1 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$ for BC 153 for BC 154		1.8×10^{-4}		—
			2.9×10^{-4}		—
h_{oe} Output admittance	$I_C = -1 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$ for BC 153 for BC 154		15		μS
			16		μS

BC 153 BC 154

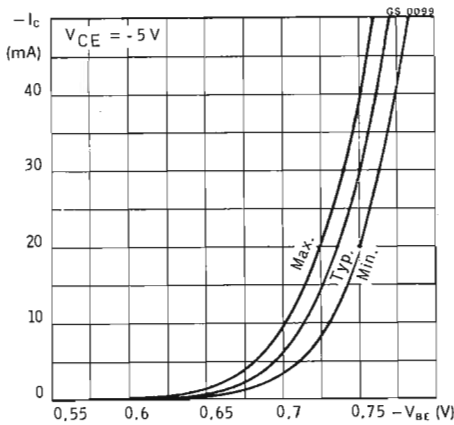
Typical output characteristics
(for BC 153 only)



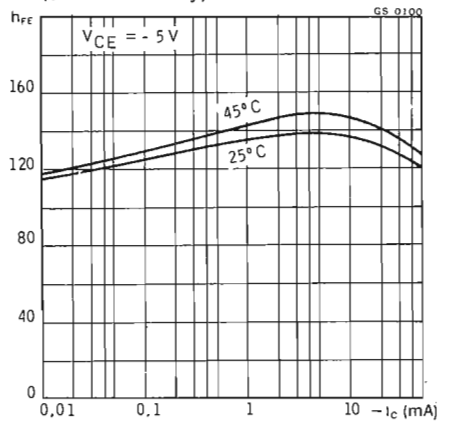
Typical output characteristics
(for BC 154 only)



DC transconductance

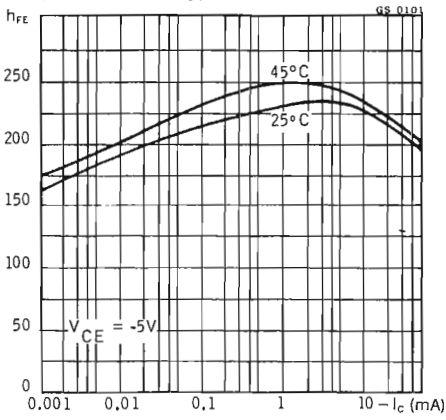


Typical DC current gain
(for BC 153 only)

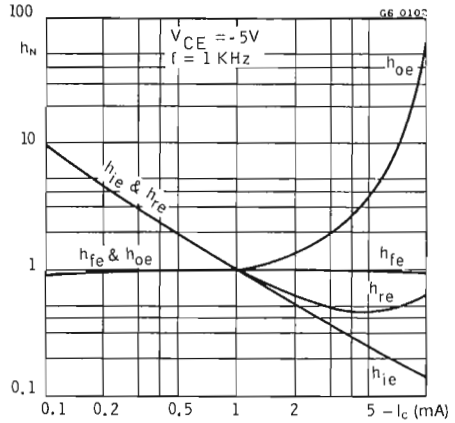


BC 153 BC 154

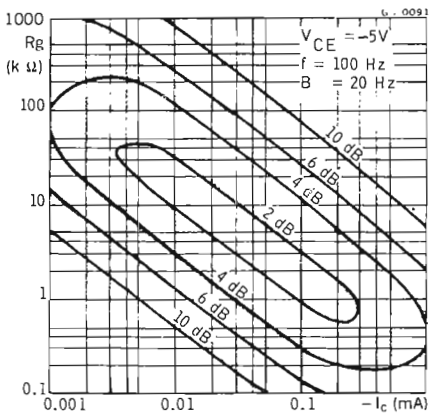
Typical DC current gain
(for BC 154 only)



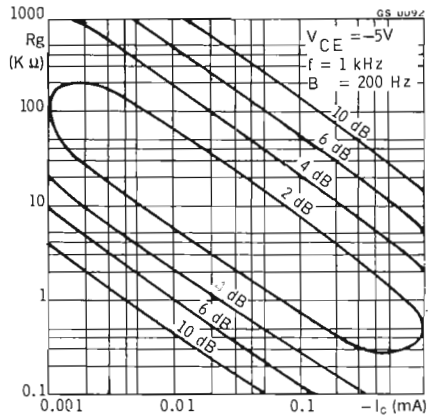
Typical normalized h parameters



Typical noise figure (for BC 154 only)

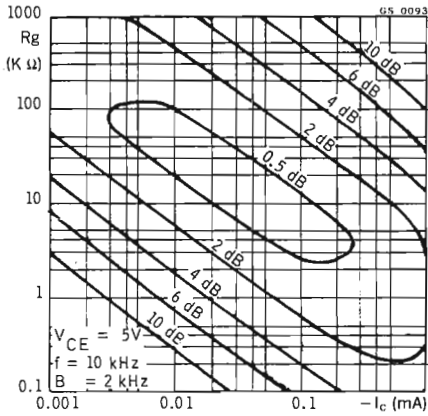


Typical noise figure (for BC 154 only)

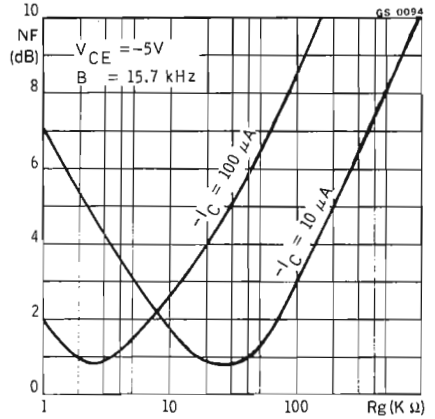


BC 153 BC 154

Typical noise figure (for BC 154 only)



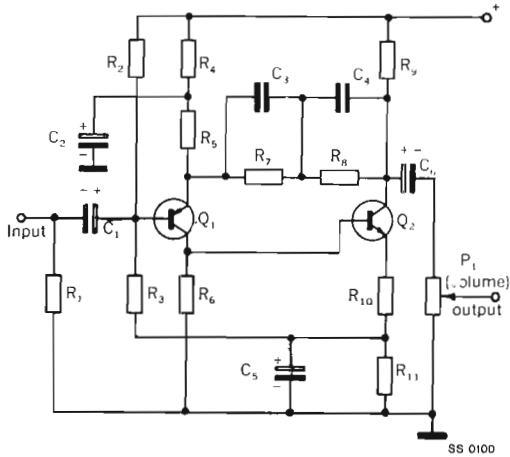
Typical noise figure (for BC 154 only)



BC 153 BC 154

TYPICAL APPLICATION FOR BC 154

Low noise preamplifier for magnetic heads



List of components

- $R_1 = 56 \text{ k}\Omega$
- $R_2 = 1.8 \text{ M}\Omega$
- $R_3 = 1.5 \text{ M}\Omega$
- $R_4 = 180 \text{ k}\Omega$
- $R_5 = 220 \Omega$
- $R_6 = 47 \text{ k}\Omega$
- $R_7 = 180 \text{ k}\Omega$
- $R_8 = 8.2 \text{ k}\Omega$
- $R_9 = 3.9 \text{ k}\Omega$
- $R_{10} = 150 \Omega$
- $R_{11} = 1 \text{ k}\Omega$
- $P_1 = 25 \text{ k}\Omega \text{ lin. (volume)}$
- $C_1 = 30 \mu\text{F}/15 \text{ V}$
- $C_2 = 50 \mu\text{F}/30 \text{ V}$
- $C_3 = 33 \text{ nF } 5\%$
- $C_4 = 10 \text{ nF } 5\%$
- $C_5 = 50 \mu\text{F}/6 \text{ V}$
- $C_6 = 30 \mu\text{F}/15 \text{ V}$
- $Q_1 = \text{BC } 154$
- $Q_2 = \text{BC } 113$

All the resistances are at 10%; 1/4 W

Overall performance

Supply Voltage	30 V
Supply Current	4 mA
Nominal Output Voltage	200 mV
THD ($V_o = 200 \text{ mV}$; $f = 1 \text{ kHz}$)	0,1 %
Sensitivity ($V_o = 200 \text{ mV}$; $f = 1 \text{ kHz}$)	4.5 mV
Signal to Noise Ratio (at nominal output voltage)	65 dB
Equalization (according to RIAA; 20 to 20000 Hz)	$\pm 1 \text{ dB}$
Input Impedance	47 k Ω
Input Overload (at nominal sensitivity)	28 dB

SILICON PLANAR PNP

BC 160 BC 161

GENERAL PURPOSE TRANSISTORS

The BC 160 and BC 161 are silicon planar epitaxial PNP transistors in TO-39 metal case. They are particularly designed for audio amplifiers and switching applications up to 1 A.

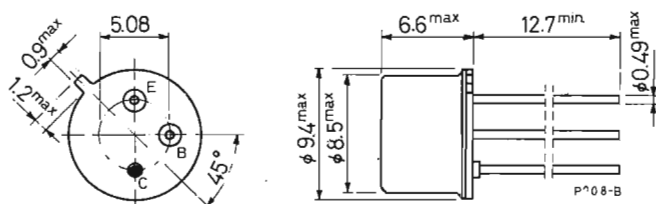
The complementary NPN types are the BC 140 and BC 141.

ABSOLUTE MAXIMUM RATINGS

		BC 160	BC 161
V_{CBO}	Collector-base voltage ($I_E = 0$)	-60 V	-80 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-40 V	-60 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)		-5 V
I_C	Collector current		-1 A
I_B	Base current		-0.1 A
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$		0.8 W
	at $T_{case} \leq 25^\circ\text{C}$		4 W
T_{stg}	Storage temperature		-55 to 200 °C
T_J	Junction temperature		200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 160

BC 161

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	44	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	220	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -50\text{ V}$ $V_{CB} = -50\text{ V } T_{amb} = 150\text{ °C}$			-200 -200	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -100\ \mu\text{A}$ for BC 160 for BC 161	-60 -80			V V
$V_{(BR)\ CEO}^*$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -10\text{ mA}$ for BC 160 for BC 161	-40 -60			V V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -100\ \mu\text{A}$	-5			V
$V_{CE(sat)}^*$ Collector-emitter saturation voltage	$I_C = -0.1\text{ A } I_B = -10\text{ mA}$ $I_C = -0.5\text{ A } I_B = -50\text{ mA}$ $I_C = -1\text{ A } I_B = -0.1\text{ A}$		-0.1 -0.35 -0.6		V V V
V_{BE}^* Base-emitter voltage	$I_C = -1\text{ A } V_{CE} = -1\text{ V}$	-1.1	-1.6		V
h_{FE}^* DC current gain	$I_C = -100\ \mu\text{A } V_{CE} = -1\text{ V}$ for BC 160-161 for BC 160-161 Gr. 6 for BC 160-161 Gr. 10 $I_C = -100\text{ mA } V_{CE} = -1\text{ V}$ for BC 160-161 for BC 160-161 Gr. 6 for BC 160-161 Gr. 10 $I_C = -1\text{ A } V_{CE} = -1\text{ V}$ for BC 160-161 for BC 160-161 Gr. 6 for BC 160-161 Gr. 10		110 46 80 40 63 63 26 15 20	250 100 160	— — — — — — — — —

BC 160 BC 161

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
h_{FE1}/h_{FE2} Matched pair ratio	$I_C = -100 \text{ mA}$ $V_{CE} = -1 \text{ V}$		1.25		—
f_T Transition frequency	$I_C = -50 \text{ mA}$ $V_{CE} = -10 \text{ V}$	50			MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -20 \text{ V}$ $f = 1 \text{ MHz}$		15		pF
t_{on} Turn-on time	$I_C = -100 \text{ mA}$ $I_{B1} = -5 \text{ mA}$		500		ns
t_{off} Turn-off time	$I_C = -100 \text{ mA}$ $I_{B1} = I_{B2} = -5 \text{ mA}$		650		ns

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

SILICON PLANAR PNP

BC 177
BC 178
BC 179

LOW NOISE GENERAL PURPOSE AUDIO AMPLIFIERS

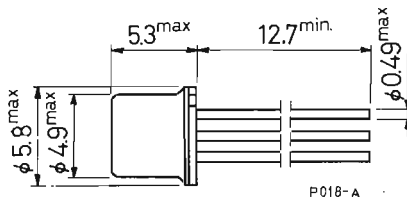
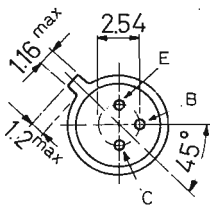
The BC 177, BC 178 and BC 179 are silicon planar epitaxial PNP transistors in TO-18 metal case. They are suitable for use in driver audio stages, low noise input audio stages and as low power, high gain general purpose transistors. The complementary NPN types are respectively the BC 107, BC 108, BC 109.

ABSOLUTE MAXIMUM RATINGS

		BC 177	BC 178	BC 179
V_{CBO}	Collector-base voltage ($I_E = 0$)	-50 V	-30 V	-25 V
$\rightarrow V_{CES}$	Collector-emitter voltage ($V_{BE} = 0$)	-45 V	-25 V	-20 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-45 V	-25 V	-20 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-5 V		
$\rightarrow I_{EM}$	Emitter peak current	200 mA		
I_C	Collector current	-100 mA		
$\rightarrow I_{CM}$	Collector peak current	-200 mA		
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$ at $T_{case} \leq 115^\circ\text{C}$	300 mW		
T_{stg}	Storage temperature	-65 to 175 °C		
T_j	Junction temperature	175 °C		

MECHANICAL DATA

Dimensions in mm



P018-A

(sim. to TO-18)

BC 177
BC 178
BC 179

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
I_{CES}	Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = -20\text{ V}$			-1 -100	nA
$V_{(BR)\ CEO}$	Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -2\text{ mA}$ for BC 177 for BC 178 for BC 179	-45 -25 -20			V V V
$V_{(BR)\ CES}$	Collector-emitter breakdown voltage ($V_{BE} = 0$)	$I_C = -10\text{ }\mu\text{A}$ for BC 177 for BC 178 for BC 179	-50 -30 -25			V V V
$V_{(BR)\ EBO}$	Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\text{ }\mu\text{A}$	-5			V
$V_{CE(sat)}$	Collector-emitter saturation voltage	$I_C = -10\text{ mA}$ $I_B = -0.5\text{ mA}$ $I_C = -100\text{ mA}$ $I_B = -5\text{ mA}$		-75	-250	mV mV
V_{BE}	Base-emitter voltage	$I_C = -2\text{ mA}$ $V_{CE} = -5\text{ V}$	-600	-640	-750	mV
$V_{BE(sat)}$	Base-emitter saturation voltage	$I_C = -10\text{ mA}$ $I_B = -0.5\text{ mA}$ $I_C = -100\text{ mA}$ $I_B = -5\text{ mA}$		-720		mV mV
h_{FE}	DC current gain	$I_C = -10\text{ }\mu\text{A}$ $V_{CE} = -5\text{ V}$	30			—

ELECTRICAL CHARACTERISTICS (continued)

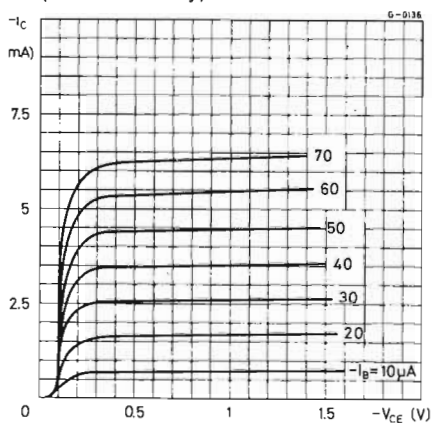
Parameter		Test conditions	Min.	Typ.	Max.	Unit
h_{fe}	Small signal current gain	$I_C = -2 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$ for BC 177 Gr. 6 for BC 177 Gr. A for BC 178 Gr. 6 for BC 178 Gr. A for BC 178 Gr. B for BC 179 Gr. A for BC 179 Gr. B	75 125 75 125 240 125 240	150 260 150 260 500 260 500	— — — — — — —	
f_T	Transition frequency	$I_C = -10 \text{ mA}$ $V_{CE} = -5 \text{ V}$		200		MHz
C_{CBO}	Collector-base capacitance	$I_E = 0$ $V_{CB} = -10 \text{ V}$		5.5		pF
NF	Noise figure	$I_C = -0.2 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $R_g = 2 \text{ k}\Omega$ $f = 1 \text{ kHz}$ $B = 200 \text{ Hz}$ for BC 177 for BC 178 for BC 179		2 2 1.2	10 10 4	dB dB dB
h_{ie}	Input impedance	$I_C = -2 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$ for BC 177 Gr. 6 for BC 177 Gr. A for BC 178 Gr. 6 for BC 178 Gr. A for BC 178 Gr. B for BC 179 Gr. A for BC 179 Gr. B		1.5 2.7 1.5 2.7 5.2 2.7 5.2		k Ω k Ω k Ω k Ω k Ω k Ω k Ω
h_{re}	Reverse voltage ratio	$I_C = -2 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$ for BC 177 Gr. 6 for BC 177 Gr. A for BC 178 Gr. 6 for BC 178 Gr. A for BC 178 Gr. B for BC 179 Gr. A for BC 179 Gr. B		1.8×10^{-4} 2.7×10^{-4} 1.8×10^{-4} 2.7×10^{-4} 4.5×10^{-4} 2.7×10^{-4} 4.5×10^{-4}		— — — — — — —

BC 177
BC 178
BC 179

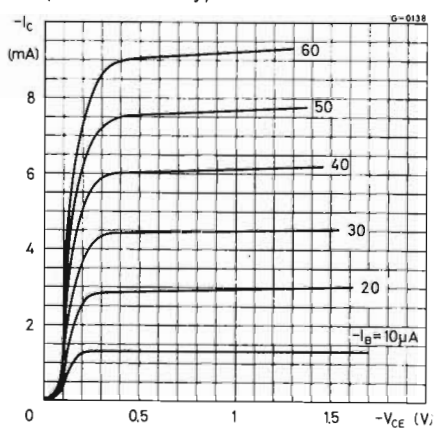
ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
h_{oe} Output admittance	$I_C = -2 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$				
	for BC 177 Gr. 6		20		μS
	for BC 177 Gr. A		25		μS
	for BC 178 Gr. 6		20		μS
	for BC 178 Gr. A		25		μS
	for BC 178 Gr. B		35		μS
	for BC 179 Gr. A		25		μS
	for BC 179 Gr. B		35		μS

Typical output characteristics
 (for **BC 177** only)

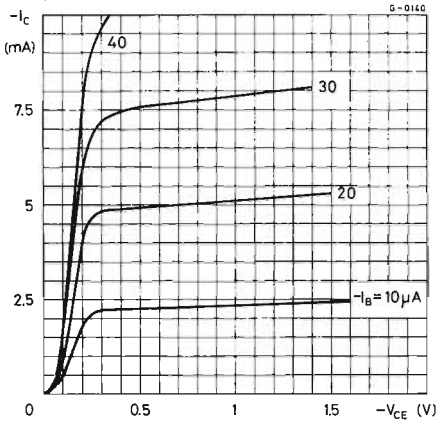


Typical output characteristics
 (for **BC 178** only)

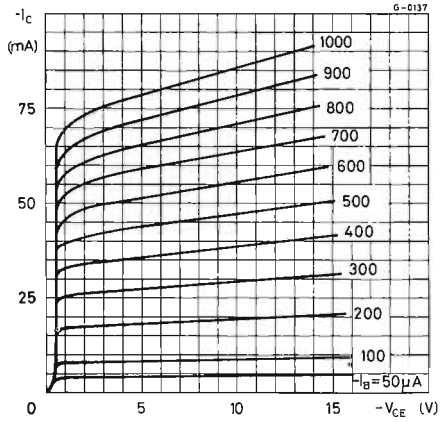


BC 177 BC 178 BC 179

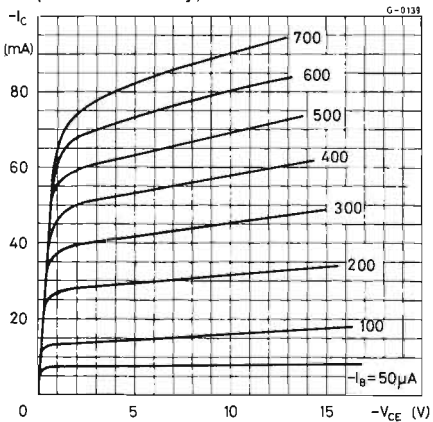
Typical output characteristics
(for BC 179 only)



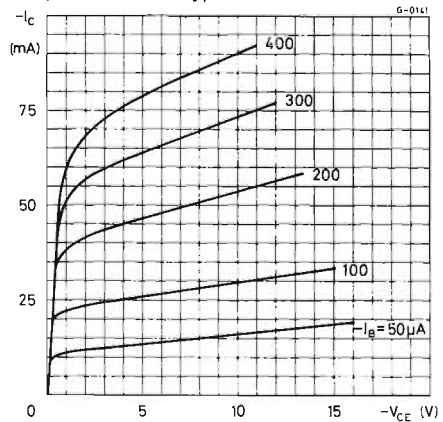
Typical output characteristics
(for BC 177 only)



Typical output characteristics
(for BC 178 only)

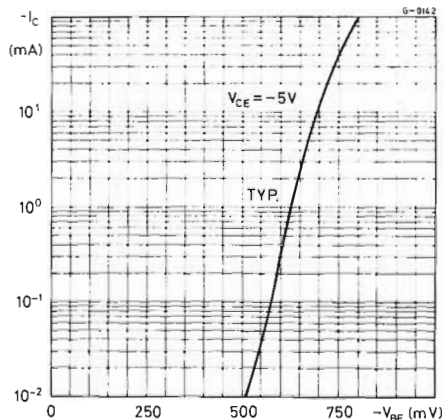


Typical output characteristics
(for BC 179 only)

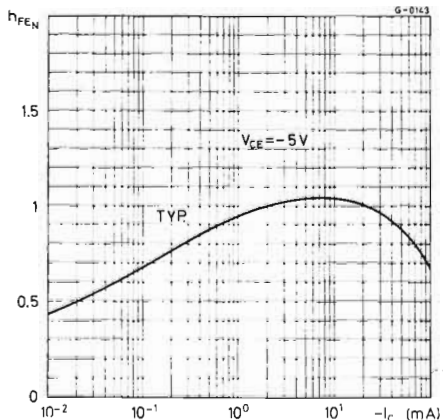


BC 177
BC 178
BC 179

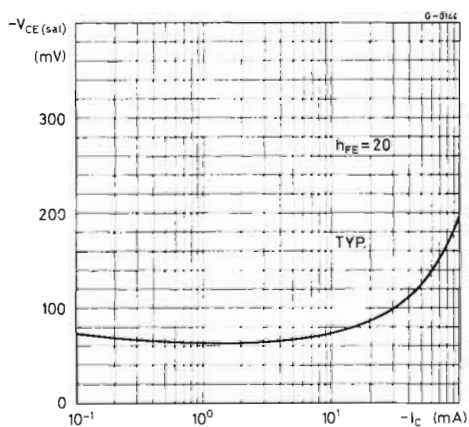
DC transconductance



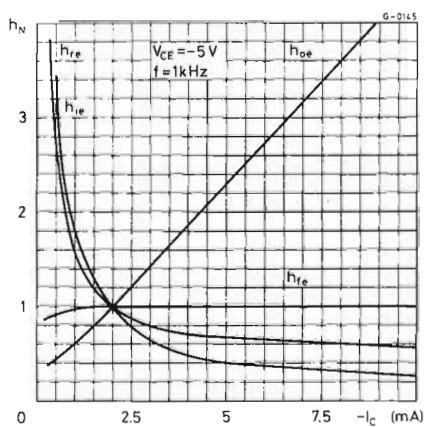
DC normalized current gain



Collector-emitter saturation voltage

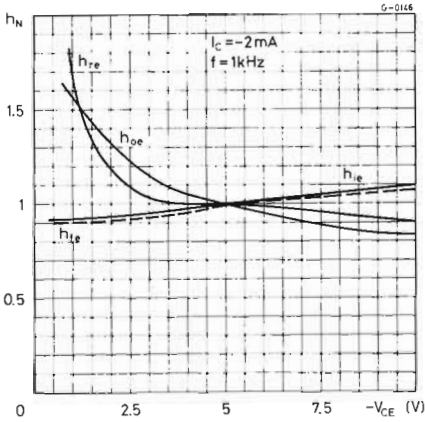


Typical normalized h parameters

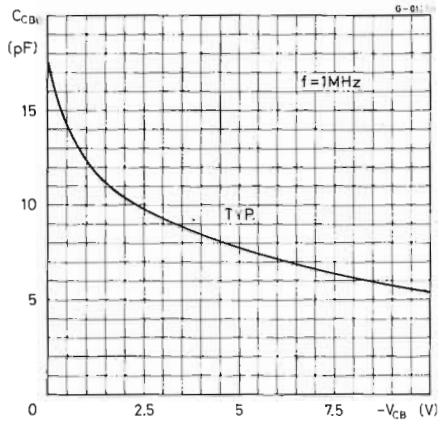


BC 177
BC 178
BC 179

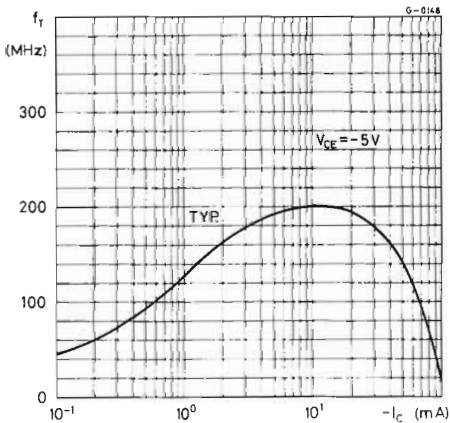
Typical normalized h parameters



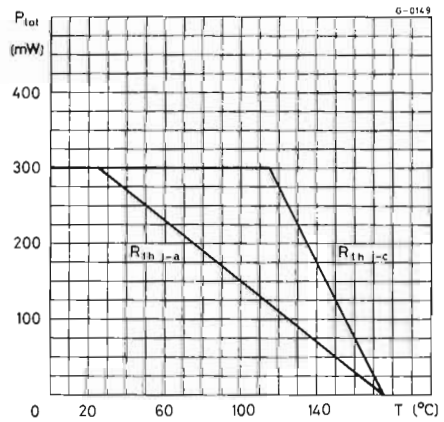
Collector-base capacitance



Transition frequency



Power rating chart



SILICON PLANAR PNP

BC 204
BC 205
BC 206

GENERAL PURPOSE AMPLIFIERS

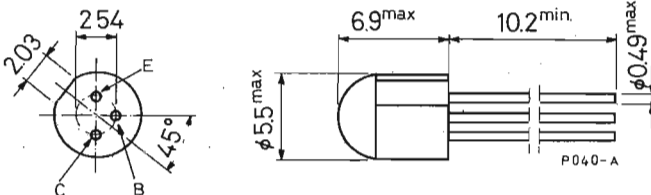
The BC 204, BC 205 and BC 206 are silicon planar epitaxial PNP transistors in TO-18 epoxy package. They are intended for general amplifier applications and TV signal processing.

ABSOLUTE MAXIMUM RATINGS

		BC 204	BC 205 BC 206
V_{CBO}	Collector-base voltage ($I_E = 0$)	-50 V	-25 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-45 V	-20 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)		-5 V
I_C	Collector current	-100 mA	
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.2 W	
	at $T_{case} \leq 25^\circ\text{C}$	0.5 W	
T_{stg}	Storage temperature	-55 to 125 °C	
T_j	Junction temperature	125 °C	

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BC 204 BC 205 BC 206

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	for BC 204 $V_{CB} = -45\text{ V}$ $V_{CB} = -45\text{ V}$ $T_{amb} = 65\text{ °C}$ for BC 205-BC 206 $V_{CB} = -20\text{ V}$ $V_{CB} = -20\text{ V}$ $T_{amb} = 65\text{ °C}$			-50 -3 -50 -3	nA μA nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\text{ }\mu\text{A}$ for BC 204 for BC 205-BC 206	-50 -25			V V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -5\text{ mA}$ for BC 204 for BC 205-BC 206	-45 -20			V V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\mu\text{A}$	-5			V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = -10\text{ mA}$ $I_B = -0.5\text{ mA}$		-0.1	-0.3	V
V_{BE} Base-emitter voltage	$I_C = -2\text{ mA}$ $V_{CE} = -5\text{ V}$	-0.55	-0.65	-0.75	V

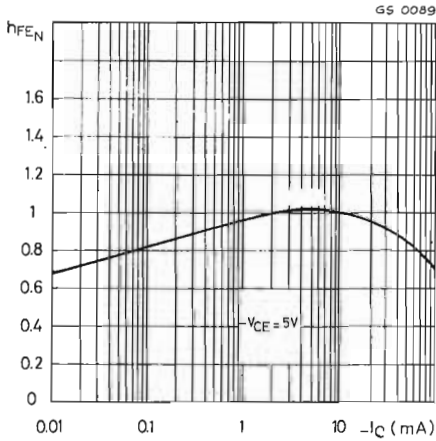
BC 204
BC 205
BC 206

ELECTRICAL CHARACTERISTICS (continued)

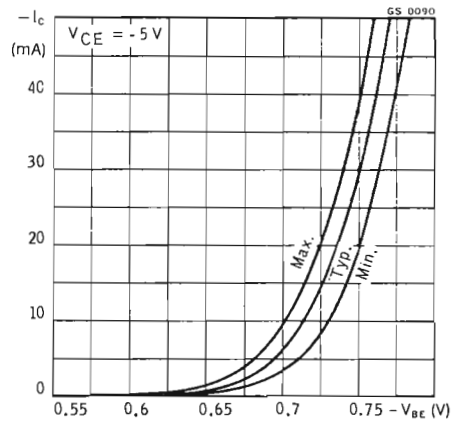
Parameter		Test conditions	Min.	Typ.	Max.	Unit		
h_{FE}	DC current gain	$I_C = -2 \text{ mA}$ $V_{CE} = -5 \text{ V}$						
			for BC 204	50	160	450	—	
			for BC 204 Gr. VI	50	90	120	—	
			for BC 204 Gr. A	110	180	220	—	
			for BC 204 Gr. B	200	300	450	—	
			for BC 205	110	270	450	—	
			for BC 205 Gr. A	110	180	220	—	
			for BC 205 Gr. B	200	350	450	—	
			for BC 206	200	400	—	—	
			for BC 206 Gr. B	200	350	450	—	
			$I_C = -10 \mu\text{A}$ $V_{CE} = -5 \text{ V}$					
			for BC 204		110	—	—	
			for BC 204 Gr. VI		80	—	—	
			for BC 204 Gr. A		130	—	—	
for BC 204 Gr. B		200	—	—				
for BC 205		200	—	—				
for BC 205 Gr. A		130	—	—				
for BC 205 Gr. B		270	—	—				
for BC 206		320	—	—				
for BC 206 Gr. B		270	—	—				
f_T	Transition frequency	$I_C = -10 \text{ mA}$ $V_{CE} = -5 \text{ V}$		160		MHz		
C_{CBO}	Collector-base capacitance	$I_E = 0$ $V_{CB} = -10 \text{ V}$ $f = 1 \text{ MHz}$		4		pF		
NF	Noise figure	$I_C = -200 \mu\text{A}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$ $B = 200 \text{ Hz}$						
			for BC 204/205	2	10	dB		
		for BC 206	1	4	dB			

BC 204
BC 205
BC 206

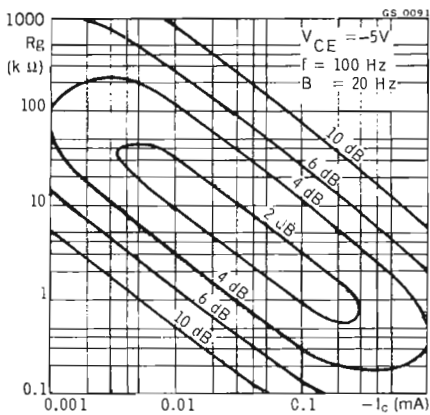
DC normalized current gain



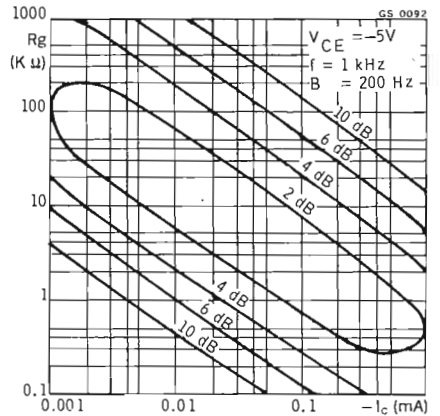
DC transconductance



Typical noise figure

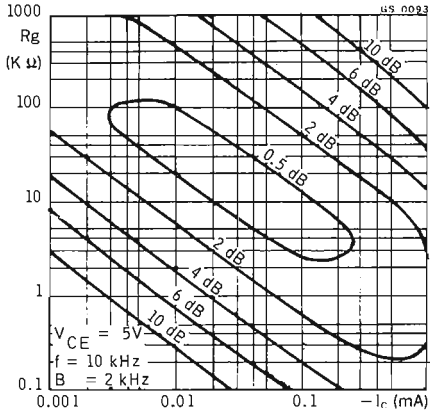


Typical noise figure

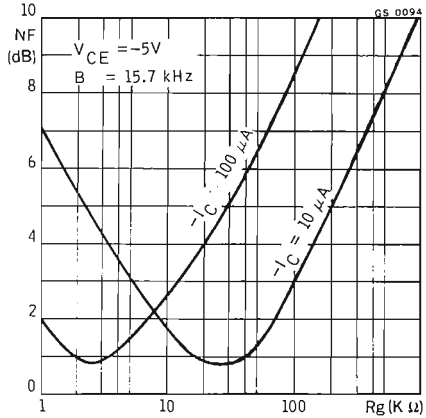


BC 204 BC 205 BC 206

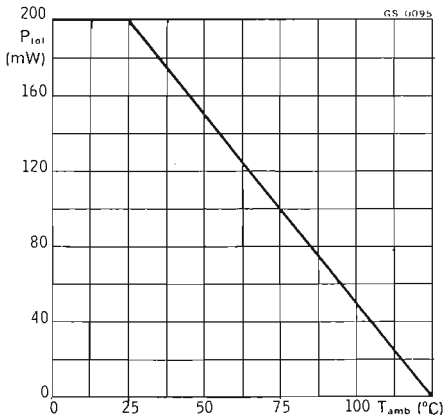
Typical noise figure



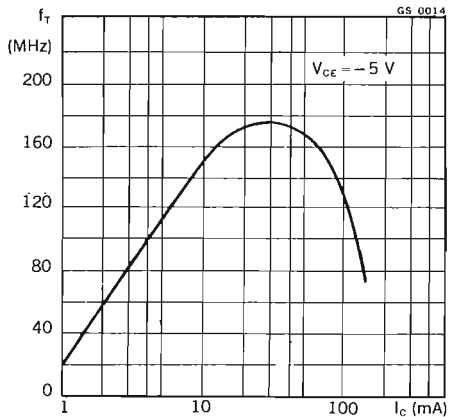
Typical noise figure



Power rating chart



Transition frequency



SILICON PLANAR NPN

BC 207
BC 208
BC 209

GENERAL PURPOSE AUDIO AMPLIFIERS

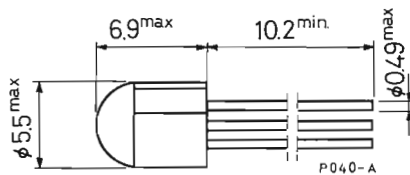
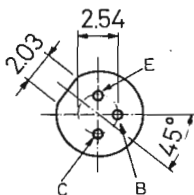
The BC 207, BC 208 and BC 209 are silicon planar epitaxial NPN transistors in TO-18 epoxy package. They are intended for use in driver or input stages of audio amplifier and in signal processing circuits of TV receivers.

ABSOLUTE MAXIMUM RATINGS

		BC 207	BC 208 BC 209
V_{CBO}	Collector-base voltage ($I_E = 0$)	50 V	25 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	45 V	20 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	5 V	
I_C	Collector current	100 mA	
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.2 W	
	at $T_{case} \leq 25^\circ\text{C}$	0.5 W	
T_{stg}	Storage temperature	-55 to 125 °C	
T_j	Junction temperature	125 °C	

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BC 207 BC 208 BC 209

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 40\text{ V}$ $V_{CB} = 40\text{ V}$ $T_{amb} = 65\text{ °C}$			50 50	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\ \mu\text{A}$ for BC 207 for BC 208-BC 209	50 25			V V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 10\text{ mA}$ for BC 207 for BC 208-BC 209	45 20			V V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu\text{A}$	5			V
$V_{CE(sat)}$ * Collector-emitter saturation voltage	$I_C = 10\text{ mA}$ $I_B = 0.5\text{ mA}$ $I_C = 100\text{ mA}$ $I_B = 5\text{ mA}$			0.25 0.6	V V
h_{FE} DC current gain	$I_C = 2\text{ mA}$ $V_{CE} = 5\text{ V}$ for BC 207 for BC 207 Gr. A for BC 207 Gr. B for BC 208 for BC 208 Gr. A for BC 208 Gr. B for BC 208 Gr. C for BC 209 for BC 209 Gr. B for BC 209 Gr. C	110 110 200 110 110 200 420 200 200 420	230 180 290 350 180 290 520 350 290 520	450 220 450 800 220 450 800 800 450 800	— — — — — — — — — —

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

ELECTRICAL CHARACTERISTICS (continued)

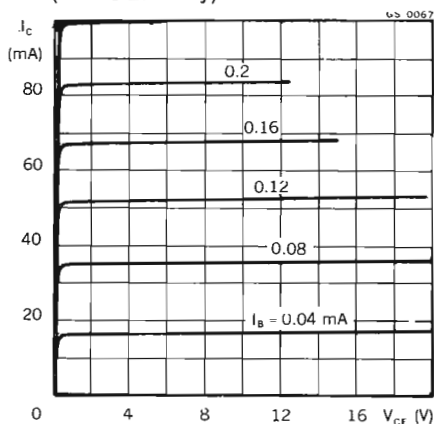
Parameter	Test conditions	Min.	Typ.	Max.	Unit
h_{FE} DC current gain	$I_C = 10 \mu A$ $V_{CE} = 5 V$ for BC 207 for BC 207 Gr. A for BC 207 Gr. B for BC 208 for BC 208 Gr. A for BC 208 Gr. B for BC 208 Gr. C for BC 209 for BC 209 Gr. B for BC 209 Gr. C	40	120 90 150 120 90 150 270 210 150 270		— — — — — — — — — —
f_T Transition frequency	$V_{CE} = 5 V$ $I_C = 10 mA$		200		MHz
NF Noise figure	$I_C = 0.2 mA$ $V_{CE} = 5 V$ $R_g = 2 k\Omega$ $f = 1 kHz$ $B = 200 Hz$ for BC 207 for BC 208 for BC 209		2 2 1.5	10 10 4	dB dB dB
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 10 V$ $f = 1 MHz$		3.1	6	pF
h_{ie} Input impedance	$I_C = 2 mA$ $V_{CE} = 5 V$ $f = 1 kHz$ for BC 207 for C 207 Gr. A for BC 207 Gr. B for BC 208 for BC 208 Gr. A for BC 208 Gr. B for BC 208 Gr. C for BC 209 for BC 209 Gr. B for BC 209 Gr. C		4 3 4.8 5.5 3 4.8 7 5.5 4.8 7		$k\Omega$ $k\Omega$ $k\Omega$ $k\Omega$ $k\Omega$ $k\Omega$ $k\Omega$ $k\Omega$ $k\Omega$ $k\Omega$
h_{oe} Output admittance	$I_C = 2 mA$ $V_{CE} = 5 V$ $f = 1 kHz$ for BC 207 for BC 207 Gr. A for BC 207 Gr. B		20 13 26		μS μS μS

BC 207
BC 208
BC 209

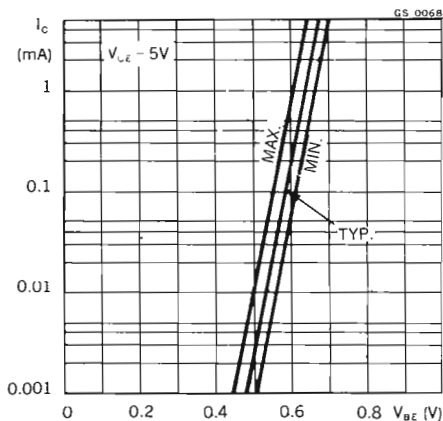
ELECTRICAL CHARACTERISTICS (continued)

h_{oe} Output admittance	for BC 208	30	μS
	for BC 208 Gr. A	13	μS
	for BC 208 Gr. B	26	μS
	for BC 208 Gr. C	34	μS
	for BC 209	30	μS
	for BC 209 Gr. B	26	μS
	for BC 209 Gr. C	34	μS
h_{re} Reverse voltage ratio	$I_C = 2 \text{ mA}$ $V_{CE} = 5 \text{ V}$ $f = 1 \text{ kHz}$		
	for BC 207	2.7×10^{-4}	—
	for BC 207 Gr. A	1.7×10^{-4}	—
	for BC 207 Gr. B	3.7×10^{-4}	—
	for BC 208	3.1×10^{-4}	—
	for BC 208 Gr. A	1.7×10^{-4}	—
	for BC 208 Gr. B	2.7×10^{-4}	—
	for BC 208 Gr. C	3.8×10^{-4}	—
	for BC 209	3.1×10^{-4}	—
	for BC 209 Gr. B	2.7×10^{-4}	—
for BC 209 Gr. C	3.8×10^{-4}	—	

Typical output characteristics
(for **BC 209** only)

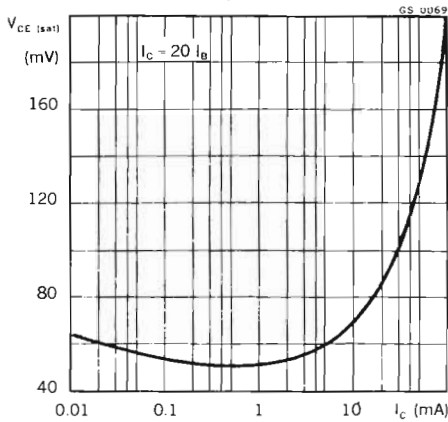


DC transconductance

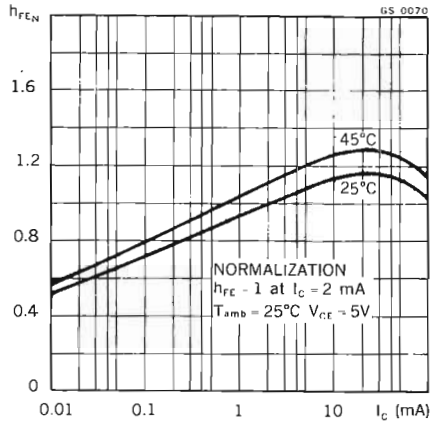


BC 207
BC 208
BC 209

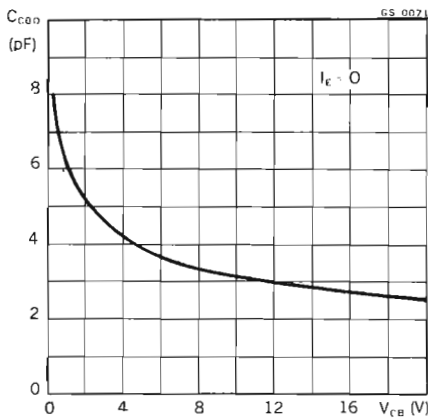
Typical collector-emitter saturation voltage



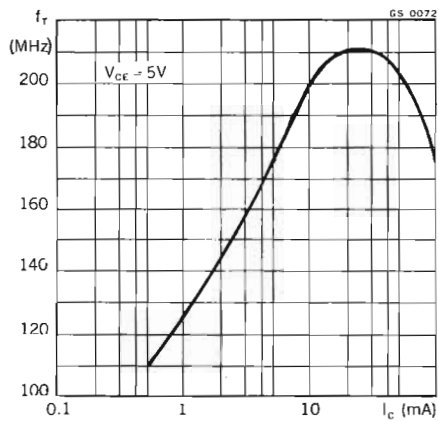
DC normalized current gain



Typical collector-base capacitance

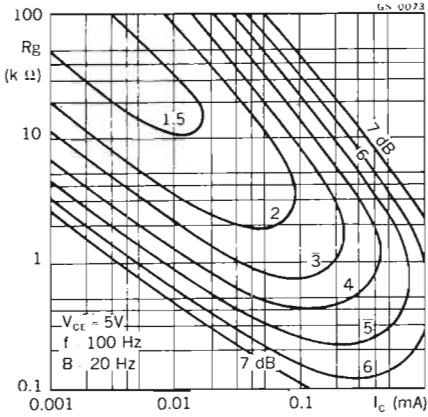


Typical transition frequency

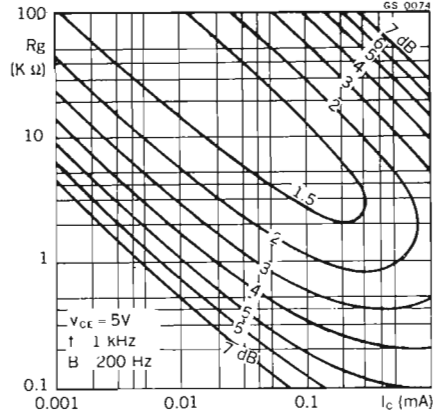


BC 207
BC 208
BC 209

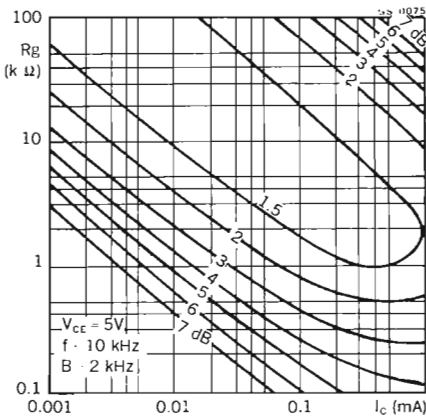
Typical noise figure (for BC 209 only)



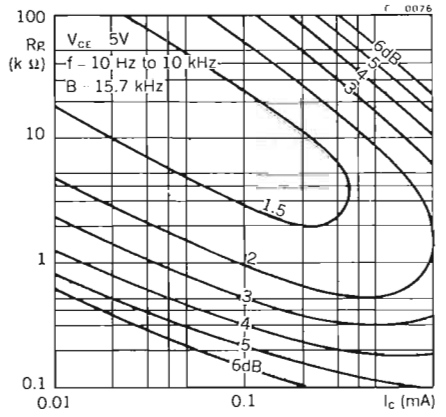
Typical noise figure (for BC 209 only)



Typical noise figure (for BC 209 only)



Typical noise figure (for BC 209 only)



BC 225

SILICON PLANAR PNP

AUDIO AMPLIFIER

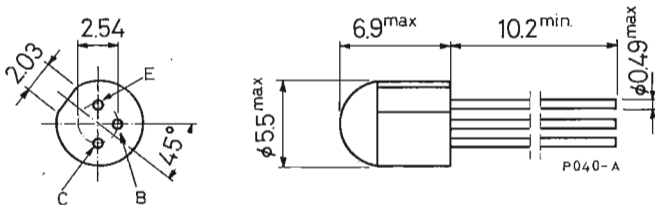
The BC 225 is a silicon planar PNP transistor in a TO-18 epoxy package. Designed for audio applications, it presents good current gain linearity from 10 μ A to 50 mA.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-5 V
I_C	Collector current	-100 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.2 W
	at $T_{case} \leq 25^\circ\text{C}$	0.5 W
T_{stg}	Storage temperature	-55 to 125 $^\circ\text{C}$
T_j	Junction temperature	125 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BC 225

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -30\text{ V}$			-100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$	-40			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -5\text{ mA}$	-40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-5			V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = -10\text{ mA}$ $I_B = -0.5\text{ mA}$		-0.1	-0.25	V
	$I_C = -50\text{ mA}$ $I_B = -5\text{ mA}$		-0.16		V
V_{BE} Base-emitter voltage	$I_C = -1\text{ mA}$ $V_{CE} = -5\text{ V}$	-0.65			V
h_{FE} DC current gain	$I_C = -10\ \mu\text{A}$ $V_{CE} = -5\text{ V}$		130		—
	$I_C = -100\ \mu\text{A}$ $V_{CE} = -5\text{ V}$		90	155	—
	$I_C = -1\text{ mA}$ $V_{CE} = -5\text{ V}$		90	170	—
	$I_C = -10\text{ mA}$ $V_{CE} = -5\text{ V}$		90	165	—
	$I_C = -50\text{ mA}$ $V_{CE} = -5\text{ V}$			140	—
f_T Transition frequency	$I_C = -1\text{ mA}$ $V_{CE} = -5\text{ V}$		70		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -5\text{ V}$ $f = 1\text{ MHz}$		4		pF
NF Noise figure	$I_C = -20\ \mu\text{A}$ $V_{CE} = -5\text{ V}$ $R_g = 10\text{ k}\Omega$ $f = 1\text{ kHz}$ $B = 200\text{ Hz}$		1		dB
	$I_C = -0.25\text{ mA}$ $V_{CE} = -5\text{ V}$ $R_g = 1\text{ k}\Omega$ $f = 1\text{ kHz}$ $B = 200\text{ Hz}$		1		dB

BC 288

SILICON PLANAR NPN

AUDIO OUTPUT AMPLIFIER

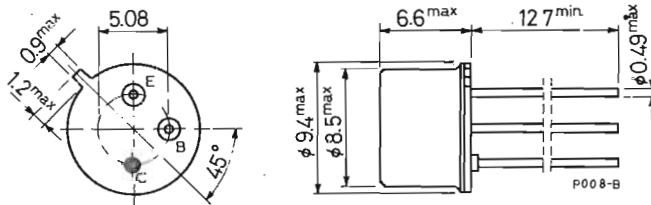
The BC 288 is a silicon planar epitaxial NPN transistor in a TO-39 metal case. It is designed to be used in low voltage audio output stages and is available as a pair 2 BC 288.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	80 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	6 V
I_C	Collector current	5 A
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.8 W
	at $T_{case} \leq 25^\circ\text{C}$	7 W
T_{stg}	Storage temperature	-55 to 200 °C
T_j	Junction temperature	200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 288

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	25	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	220	°C/W

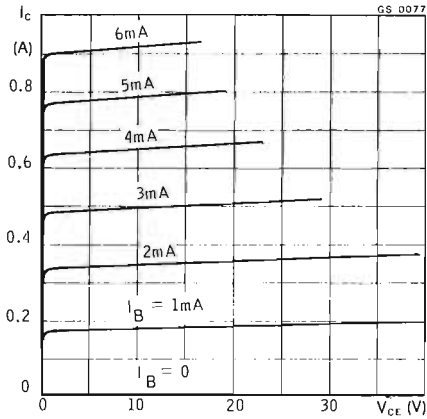
ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 30\text{ V}$			10	μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 1\text{ mA}$	80			V
$V_{CEO(sus)}$ * Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 50\text{ mA}$	40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 1\text{ mA}$	6			V
$V_{CE(sat)}$ * Collector-emitter saturation voltage	$I_C = 2\text{ A}$ $I_B = 0.2\text{ A}$	0.35	0.6		V
V_{BE} * Base-emitter voltage	$I_C = 2\text{ A}$ $V_{CE} = 2\text{ V}$	0.95			V
h_{FE} * DC current gain	$I_C = 100\text{ mA}$ $V_{CE} = 2\text{ V}$ $I_C = 500\text{ mA}$ $V_{CE} = 2\text{ V}$ $I_C = 2\text{ A}$ $V_{CE} = 2\text{ V}$	30	120	200	—
h_{FE1}/h_{FE2} Matched pair	$I_C = 300\text{ mA}$ $V_{CE} = 5\text{ V}$		1.4		—
f_T Transition frequency	$I_C = 2\text{ A}$ $V_{CE} = 2\text{ V}$		80		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 10\text{ V}$		45		pF

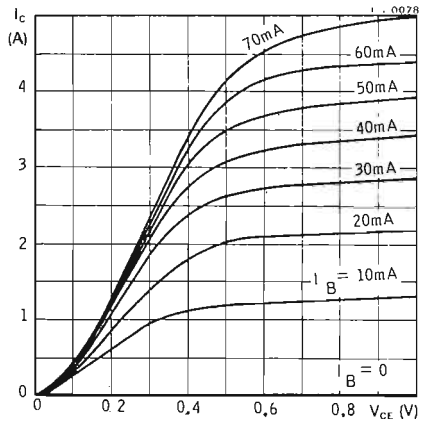
* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 288

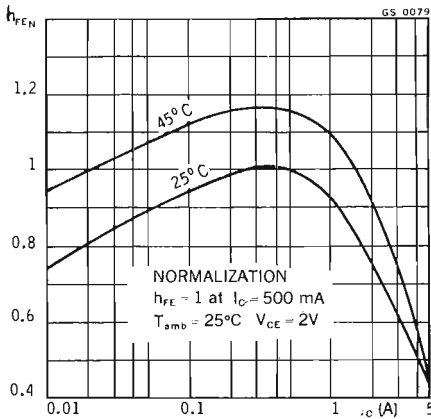
Typical output characteristics



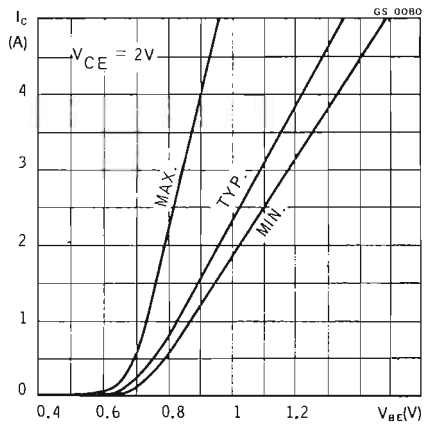
Typical output characteristics



Normalized DC current gain

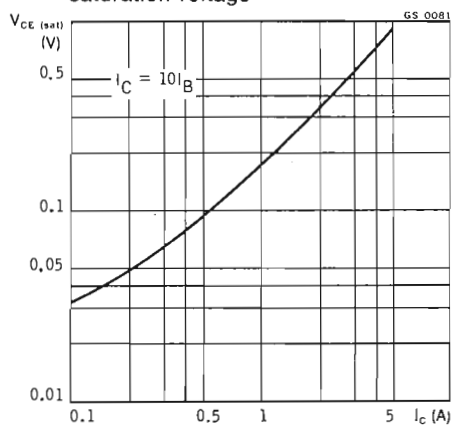


DC transconductance

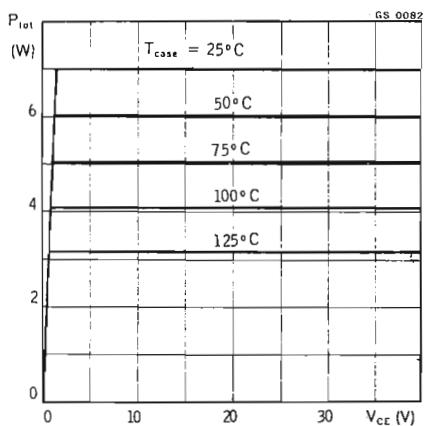


BC 288

Typical collector-emitter saturation voltage



Power rating chart



SILICON PLANAR PNP

BC 297 BC 298

AUDIO DRIVERS OR OUTPUT STAGES

The BC 297 and BC 298 are silicon planar epitaxial PNP transistors in TO-18 metal case. They are particularly intended for use in high current high gain applications, in driver stages of hi-fi equipments or in output stages of low power class B amplifiers.

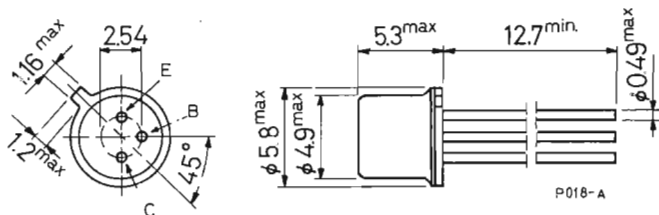
The complementary NPN types are the BC 377 and BC 378, respectively.

ABSOLUTE MAXIMUM RATINGS

		BC 297	BC 298
V_{CES}	Collector-emitter voltage ($V_{BE} = 0$)	-50 V	-30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-45 V	-25 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)		-5 V
I_E	Emitter current		1.2 A
I_C	Collector current		-1 A
I_B	Base current		-0.2 A
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$ at $T_{case} \leq 75^\circ\text{C}$		375 mW 1 W
T_{stg}	Storage temperature		-65 to 175 °C
T_j	Junction temperature		175 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-18)

BC 297 BC 298

THERMAL DATA

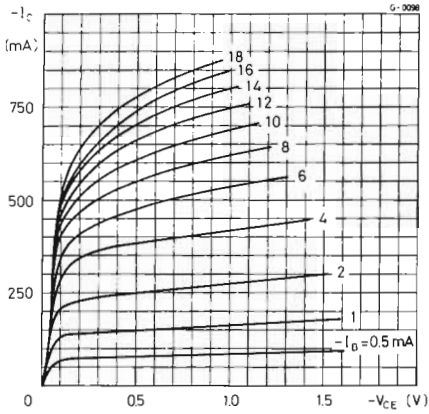
$R_{th\ j-case}$	Thermal resistance junction-case	max	100	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	400	°C/W

ELECTRICAL CHARACTERISTICS (T_{case} = 25 °C unless otherwise specified)

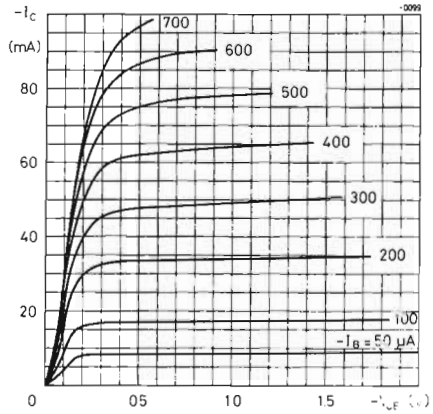
	Parameter	Test conditions	Min.	Typ.	Max.	Unit
	I_{CES} Collector cutoff current ($V_{BE} = 0$)	for BC 297 $V_{CE} = -50\text{ V}$ for BC 298 $V_{CE} = -30\text{ V}$			-100 -100	nA nA
	$V_{(BR)\ CE0}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -10\text{ mA}$ for BC 297 for BC 298	-45 -25			V V
→	$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-5			V
	$V_{CE\ (sat)}$ Collector-emitter saturation voltage	$I_C = -500\text{ mA}$ $I_B = -50\text{ mA}$			-0.7	V
	V_{BE} Base-emitter voltage	$I_C = -100\text{ mA}$ $V_{CE} = -1\text{ V}$	-770			mV
	$V_{BE\ (sat)}$ Base-emitter saturation voltage	$I_C = -500\text{ mA}$ $I_B = -50\text{ mA}$			-1.2	V
→	h_{FE} DC current gain Gr. 6 Gr. 7	$I_C = -100\text{ mA}$ $V_{CE} = -1\text{ V}$ $I_C = -100\text{ mA}$ $V_{CE} = -1\text{ V}$ $I_C = -300\text{ mA}$ $V_{CE} = -1\text{ V}$	75 125 30	150 260		— — —
	h_{FE1}/h_{FE2} Matched pair ratio	$I_C = -100\text{ mA}$ $V_{CE} = -1\text{ V}$		1.41		—
→	f_T Transition frequency	$I_C = -50\text{ mA}$ $V_{CE} = -10\text{ V}$		250		MHz
→	C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -10\text{ V}$		8		pF
→	C_{EBO} Emitter-base capacitance	$I_C = 0$ $V_{EB} = -0.5\text{ V}$		30		pF

BC 297 BC 298

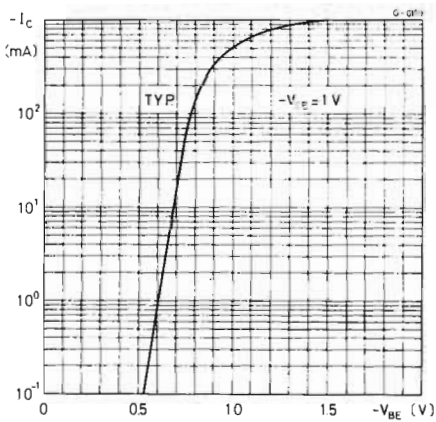
Typical output characteristics



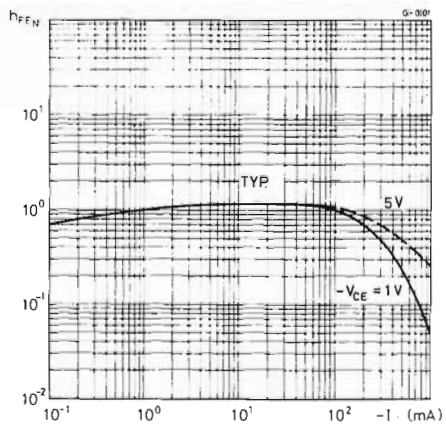
Typical output characteristics



DC transconductance

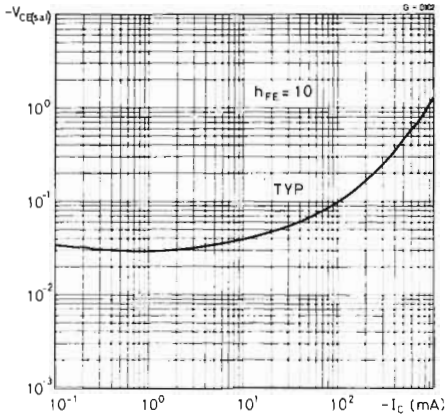


DC normalized current gain

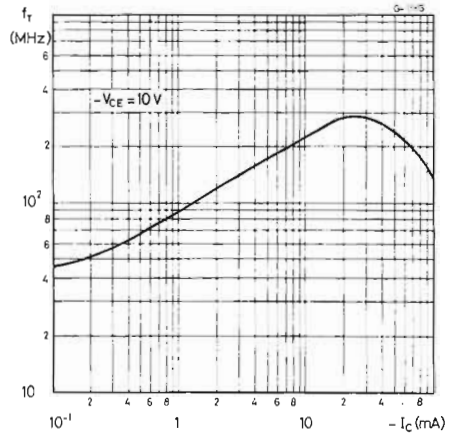


BC 297 BC 298

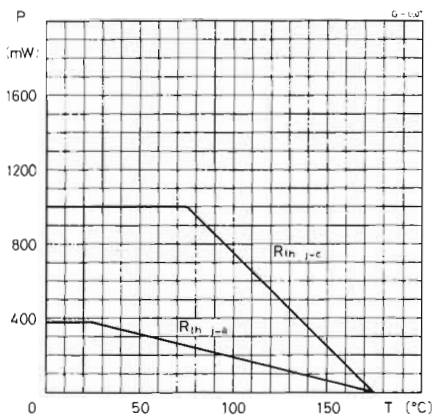
Collector-emitter saturation voltage



Typical transition frequency



Power rating chart



SILICON PLANAR NPN

BC 300
BC 301
BC 302

MEDIUM POWER AUDIO DRIVERS

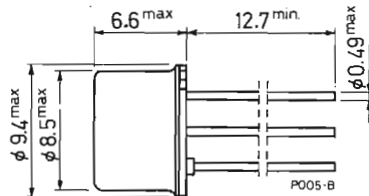
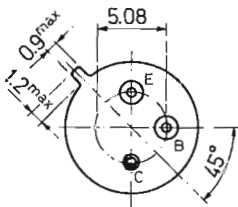
The BC 300, BC 301 and BC 302 are silicon planar epitaxial NPN transistors in TO-39 metal case. They are intended for audio driver stages in commercial and industrial equipments. In addition they are useful as high speed saturated switches and general purpose amplifiers. The PNP types complementary to BC 301 and BC 302 are respectively the BC 303 and BC 304.

ABSOLUTE MAXIMUM RATINGS

		BC 300	BC 301	BC 302
V_{CBO}	Collector-base voltage ($I_E = 0$)	120 V	90 V	60 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	80 V	60 V	45 V
V_{CEV}	Collector-emitter voltage ($V_{BE} = -1.5$ V)	120 V	90 V	—
V_{EBO}	Emitter-base voltage ($I_C = 0$)	7 V		
I_C	Collector current	0.5 A		
I_{CM}	Collector peak current	1 A		
I_{BM}	Base peak current	0.5 A		
P_{tot}	Total power dissipation at $T_{amb} \leq 25$ °C	0.85 W		
	at $T_{case} \leq 25$ °C	6 W		
T_{stg}	Storage temperature	-65 to 175 °C		
T_J	Junction temperature	175 °C		

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 300 BC 301 BC 302

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	25	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	175	°C/W

ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 60\text{ V}$		5	20	nA
I_{EBO} Emitter cutoff current ($I_C = 0$)	$V_{EB} = 7\text{ V}$			20	nA
$V_{CEO(sus)}$ *Collector-emitter voltage ($I_B = 0$)	$I_C = 100\text{ mA}$ for BC 300 for BC 301 for BC 302	80 60 45			V V V
$V_{CEV(sus)}$ *Collector-emitter voltage	$I_C = 100\text{ mA}$ $V_{BE} = -1.5\text{ V}$ for BC 300 for BC 301	120 90			V V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = 150\text{ mA}$ $I_B = 15\text{ mA}$		0.2	0.5	V
V_{BE} Base-emitter voltage	$I_C = 150\text{ mA}$ $V_{CE} = 10\text{ V}$		0.78		V
h_{FE} DC current gain	Gr. 4 $I_C = 150\text{ mA}$ $V_{CE} = 10\text{ V}$ Gr. 5 $I_C = 150\text{ mA}$ $V_{CE} = 10\text{ V}$ Gr. 6 $I_C = 150\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 0.1\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 500\text{ mA}$ $V_{CE} = 10\text{ V}$	40 70 120 20 20		80 140 240	— — — — —
f_T Transition frequency	$I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$		120		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 10\text{ V}$		10		pF
h_{ie} Input impedance	$I_C = 5\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 1\text{ kHz}$		1.1		k Ω

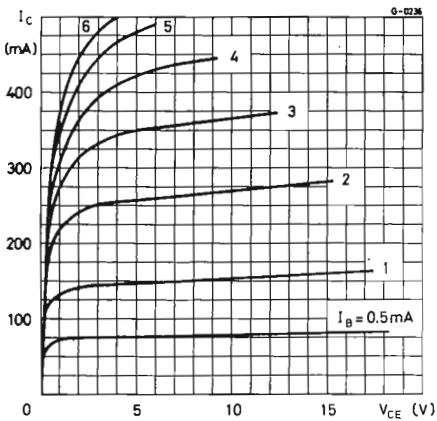
* Pulsed; pulse duration = 300 μ s, duty factor = 1.5%.

BC 300
BC 301
BC 302

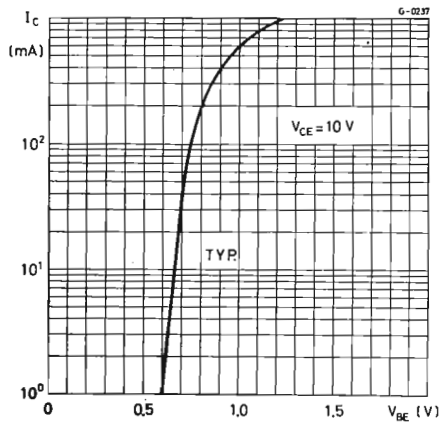
ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
h_{re} Reverse voltage ratio	$I_C = 5 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = 10 \text{ V}$		1.7×10^{-4}		—
h_{fe} Small signal current gain	$I_C = 5 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = 10 \text{ V}$		140		—
h_{oe} Output admittance	$I_C = 5 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = 10 \text{ V}$		14		μS

Typical output characteristics

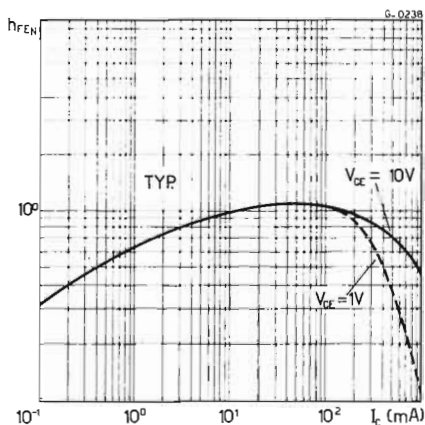


DC transconductance

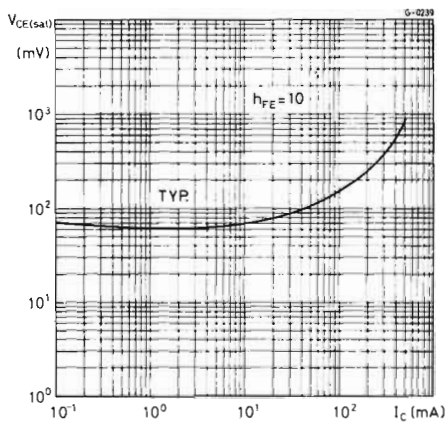


BC 300 BC 301 BC 302

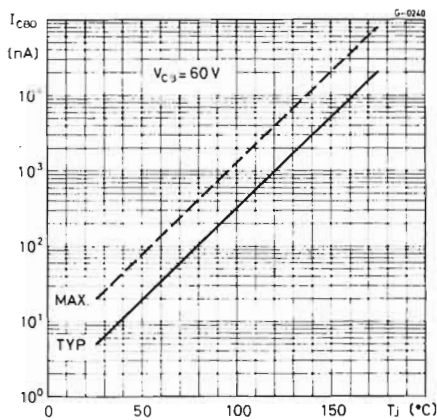
DC normalized current gain



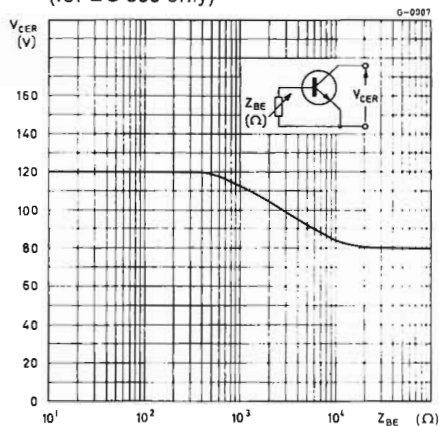
Collector-emitter saturation voltage



Collector cutoff current

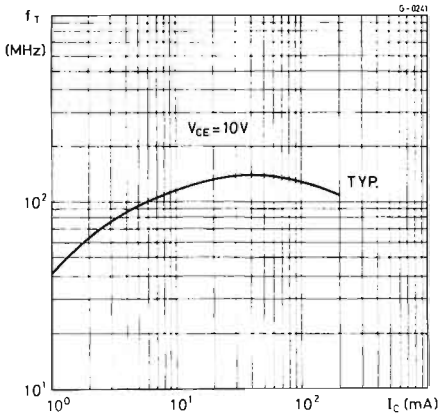


Collector-emitter breakdown voltage (for BC 300 only)

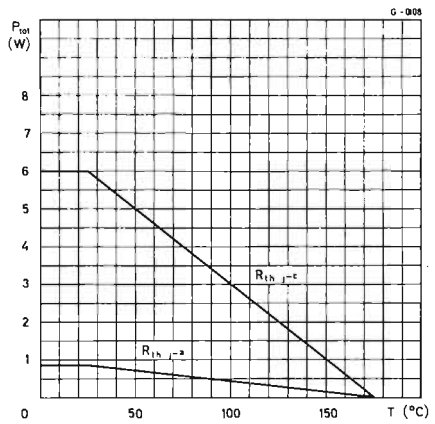


BC 300
BC 301
BC 302

Transition frequency



Power rating chart



BC 303 BC 304

SILICON PLANAR PNP

MEDIUM POWER AUDIO DRIVERS

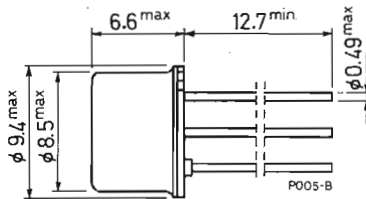
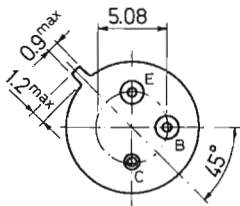
The BC 303 and BC 304 are silicon planar epitaxial PNP transistors in TO-39 metal case. They are intended particularly as audio driver stages in commercial and professional equipments. In addition they are useful as high speed saturated switches and general purpose amplifiers. The complementary NPN types are respectively the BC 301 and BC 302.

ABSOLUTE MAXIMUM RATINGS

		BC 303	BC 304
V_{CBO}	Collector-base voltage ($I_E = 0$)	-85 V	-60 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-60 V	-45 V
V_{CEV}	Collector-emitter voltage ($V_{BE} = 1.5$ V)	-85 V	—
V_{EBO}	Emitter-base voltage ($I_C = 0$)		-7 V
I_C	Collector current	-0.5 A	
I_{CM}	Collector peak current	-1 A	
I_{BM}	Base peak current	-0.5 A	
P_{tot}	Total power dissipation at $T_{amb} \leq 25$ °C at $T_{case} \leq 25$ °C	0.85 W 6 W	
T_{stg}	Storage temperature	-65 to 175 °C	
T_j	Junction temperature	175 °C	

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 303 BC 304

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	25	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	175	°C/W

ELECTRICAL CHARACTERISTICS (T_{case} = 25 °C unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -60\text{ V}$		-5	-20	nA
I_{EBO} Emitter cutoff current ($I_C = 0$)	$V_{EB} = -5\text{ V}$			-20	nA
$V_{CE0(sus)}$ *Collector-emitter voltage ($I_B = 0$)	$I_C = -100\text{ mA}$ for BC 303 for BC 304	-60 -45			V V
$V_{CEV(sus)}$ *Collector-emitter voltage (for BC 303 only)	$I_C = -100\text{ mA}$ $V_{BE} = 1.5\text{ V}$	-85			V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = -150\text{ mA}$ $I_B = -15\text{ mA}$		-0.25	-0.65	V
V_{BE} Base-emitter voltage	$I_C = -150\text{ mA}$ $V_{CE} = -10\text{ V}$		-0.78		V
h_{FE} DC current gain	Gr. 4 $I_C = -150\text{ mA}$ $V_{CE} = -10\text{ V}$ Gr. 5 $I_C = -150\text{ mA}$ $V_{CE} = -10\text{ V}$ Gr. 6 $I_C = -150\text{ mA}$ $V_{CE} = -10\text{ V}$ $I_C = 0.1\text{ mA}$ $V_{CE} = -10\text{ V}$ $I_C = -500\text{ mA}$ $V_{CE} = -10\text{ V}$	40 70 120		80 140 240	— — — — —
f_T Transition frequency	$I_C = -10\text{ mA}$ $V_{CE} = -10\text{ V}$		75		MHz
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -10\text{ V}$		15		pF
h_{ie} Input impedance	$I_C = -5\text{ mA}$ $V_{CE} = -10\text{ V}$ $f = 1\text{ kHz}$		0.9		kΩ

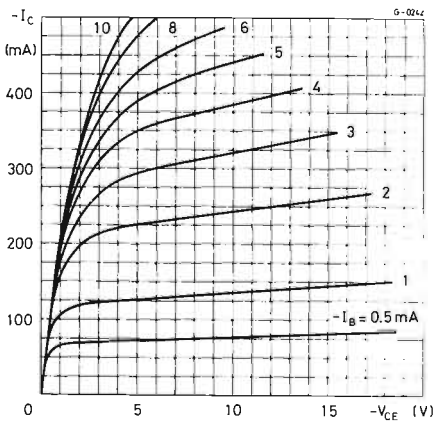
* Pulsed: pulse duration = 300 μs, duty factor = 1.5%.

BC 303 BC 304

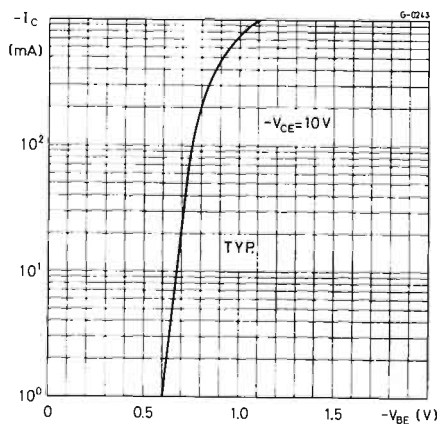
ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min. Typ. Max.	Unit
h_{re}	Reverse voltage ratio	$I_C = -5 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = -10 \text{ V}$	1.7×10^{-4}	—
h_{fe}	Small signal current gain	$I_C = -5 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = -10 \text{ V}$	140	—
h_{oe}	Output admittance	$I_C = -5 \text{ mA}$ $f = 1 \text{ kHz}$ $V_{CE} = -10 \text{ V}$	45	μS

Typical output characteristics

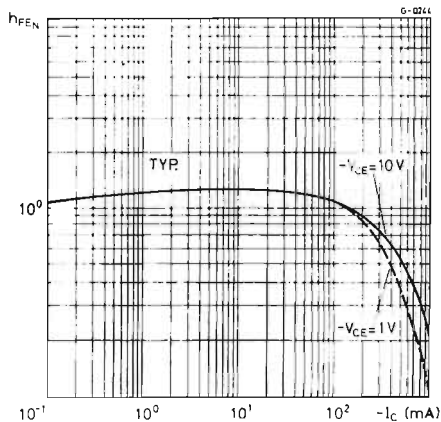


DC transconductance

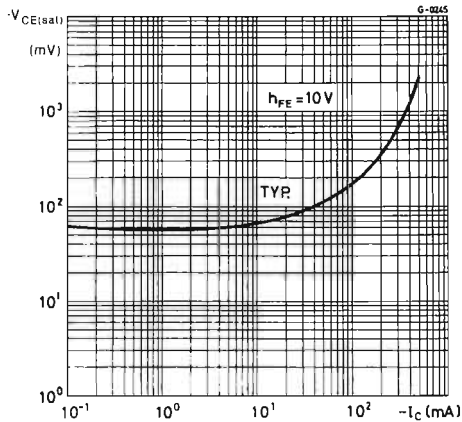


BC 303 BC 304

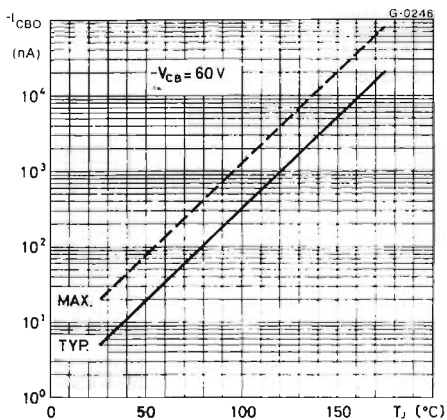
DC normalized current gain



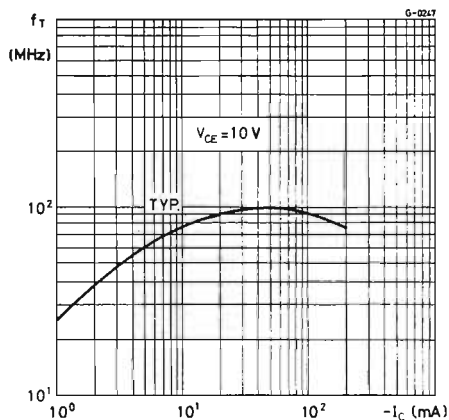
Collector-emitter saturation voltage



Collector cutoff current



Transition frequency



BC 323

SILICON PLANAR NPN

TV VERTICAL OUTPUT AMPLIFIER

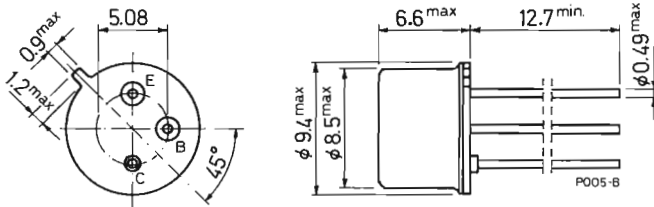
The BC 323 is a silicon planar epitaxial NPN transistor in a TO-39 metal case. It is designed as the output stage of a vertical deflection amplifier for TV receivers.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	100	V
V_{CES}	Collector-emitter voltage ($V_{BE} = 0$)	100	V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	60	V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	5	V
I_C	Collector current	5	A
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.8	W
	at $T_{case} \leq 25^\circ\text{C}$	7	W
T_{stg}	Storage temperature	-55 to 200	$^\circ\text{C}$
T_j	Junction temperature	200	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 323

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	25	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	220	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 40\text{ V}$ $V_{CB} = 40\text{ V}$ $T_{amb} = 75\text{ °C}$		0.02	10	μA μA
$V_{(BR)\ CES}$ Collector-emitter breakdown voltage ($V_{BE} = 0$)	$I_C = 1\text{ mA}$	100			V
$V_{CEO(sus)}$ Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 50\text{ mA}$	60			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 1\text{ mA}$	5			V
$V_{CE(sat)}$ * Collector-emitter saturation voltage	$I_C = 500\text{ mA}$ $I_B = 50\text{ mA}$		0.07	0.15	V
$V_{BE(sat)}$ * Base-emitter saturation voltage	$I_C = 500\text{ mA}$ $I_B = 50\text{ mA}$		0.7	0.9	V
h_{FE} * DC current gain	$I_C = 50\text{ mA}$ $V_{CE} = 1\text{ V}$ $I_C = 500\text{ mA}$ $V_{CE} = 1\text{ V}$	45	140	225	—
f_T Transition frequency	$I_C = 500\text{ mA}$ $V_{CE} = 5\text{ V}$		100		MHz
C_{EBO} Emitter-base capacitance ($I_C = 0$)	$V_{EB} = 0.5\text{ V}$			500	pF
C_{CBO} Collector-base capacitance ($I_E = 0$)	$V_{CB} = 10\text{ V}$			80	pF

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 377
BC 378

SILICON PLANAR NPN

AUDIO DRIVERS OR OUTPUT STAGES

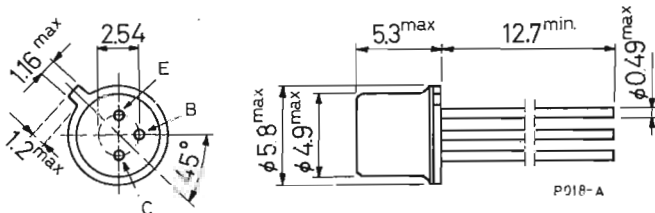
The BC 377 and BC 378 are silicon planar epitaxial NPN transistors in TO-18 metal case. They are particularly intended for use in high current, high gain applications, in driver stages of hi-fi equipments or in output stages of low power class B amplifiers. The complementary PNP types are the BC 297 and BC 298, respectively.

ABSOLUTE MAXIMUM RATINGS

		BC 377	BC 378
V_{CES}	Collector-emitter voltage ($I_{EB} = 0$)	50 V	30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	45 V	25 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	6 V	
I_E	Emitter current	-1.2 A	
I_C	Collector current	1 A	
I_B	Base current	0.2 A	
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	375 mW	
	at $T_{case} \leq 75^\circ\text{C}$	1 W	
T_{stg}	Storage temperature	-65 to 175 °C	
T_j	Junction temperature	175 °C	

MECHANICAL DATA

Dimensions in mm



(sim. to TO-18)

BC 377

BC 378

THERMAL DATA

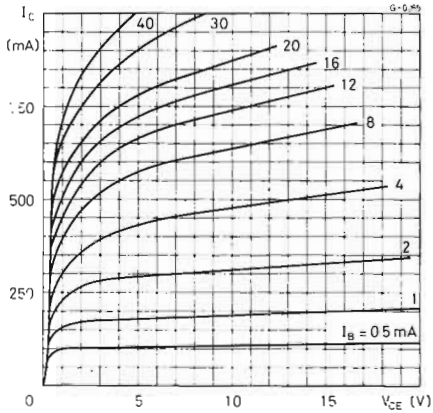
$R_{th\ j-case}$	Thermal resistance junction-case	max	100	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	400	°C/W

ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

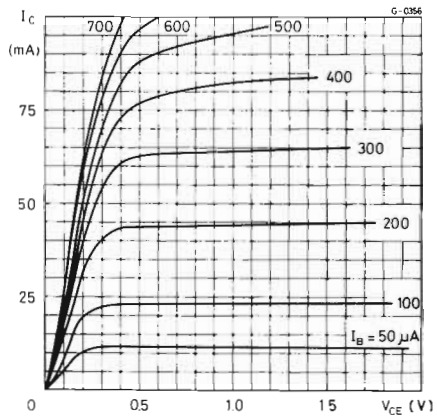
	Parameter	Test conditions	Min.	Typ.	Max.	Unit
	I_{CES} Collector cutoff current ($V_{BE} = 0$)	for BC 377 $V_{CE} = 50\text{ V}$ for BC 378 $V_{CE} = 30\text{ V}$			15 15	nA nA
→	$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu\text{A}$	6			V
	$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 2\text{ mA}$ for BC 377 for BC 378	45 25			V V
	$V_{CE\ (sat)}$ Collector-emitter saturation voltage	$I_C = 500\text{ mA}$ $I_B = 50\text{ mA}$			0.7	V
	V_{BE} Base-emitter voltage	$I_C = 100\text{ mA}$ $V_{CE} = 1\text{ V}$	740			mV
	$V_{BE\ (sat)}$ Base-emitter saturation voltage	$I_C = 500\text{ mA}$ $I_B = 50\text{ mA}$			1.2	V
→	h_{FE} DC current gain Gr. 6 Gr. 7	$I_C = 100\text{ mA}$ $V_{CE} = 1\text{ V}$ $I_C = 100\text{ mA}$ $V_{CE} = 1\text{ V}$ $I_C = 300\text{ mA}$ $V_{CE} = 1\text{ V}$	75 125 40	150 260		— — —
	h_{FE1}/h_{FE2} Matched pair ratio	$I_C = 100\text{ mA}$ $V_{CE} = 1\text{ V}$		1.41		—
→	f_T Transition frequency	$I_C = 50\text{ mA}$ $V_{CE} = 10\text{ V}$	300			MHz
→	C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 10\text{ V}$	8			pF
→	C_{EBO} Emitter-base capacitance	$I_C = 0$ $V_{EB} = 0.5\text{ V}$	30			pF

BC 377 BC 378

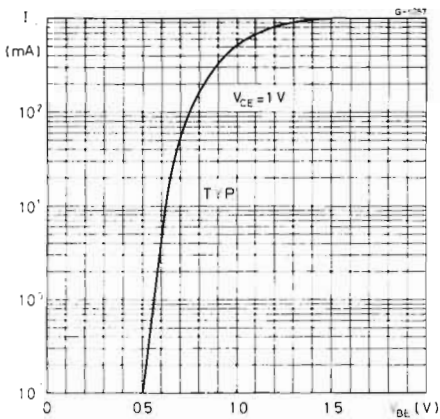
Typical output characteristics



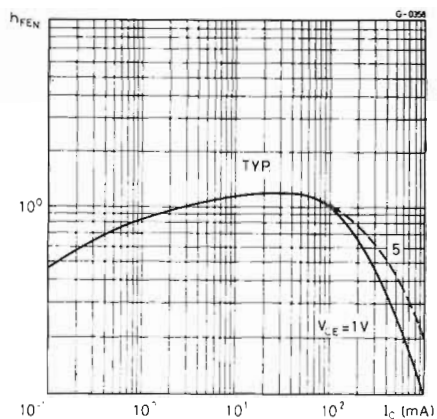
Typical output characteristics



DC transconductance

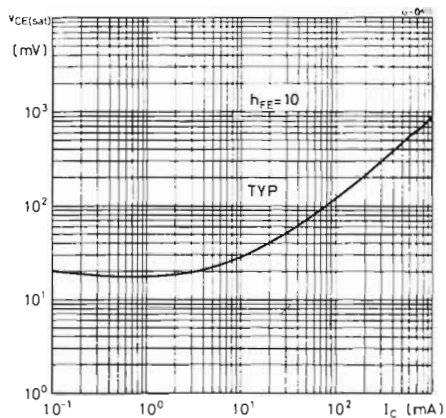


DC normalized current gain

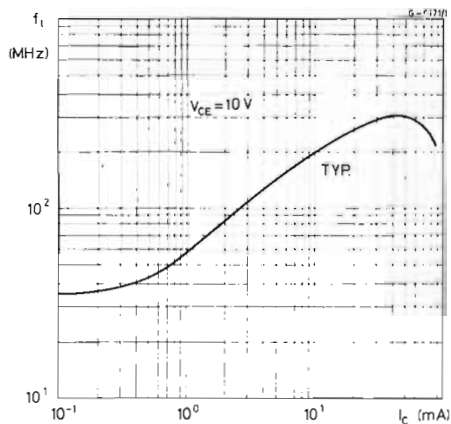


BC 377 BC 378

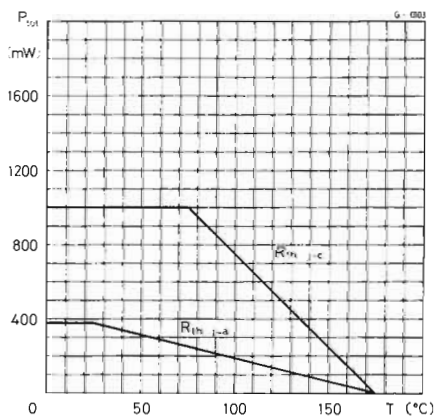
Collector-emitter saturation voltage



Transition frequency



Power rating chart



SILICON PLANAR NPN

BC 440 BC 441

MEDIUM POWER AMPLIFIER

The BC 440 and BC 441 are silicon planar epitaxial NPN transistors in TO-39 metal case. They are intended for general purpose applications, especially for driver stages.

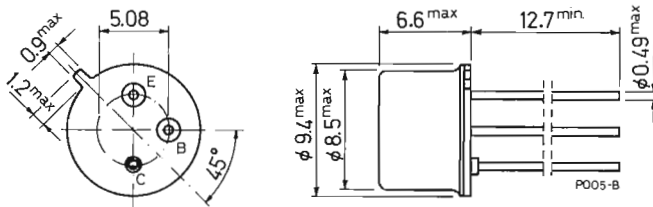
The complementary PNP types are respectively the BC 460 and BC 461.

ABSOLUTE MAXIMUM RATINGS

		BC 440	BC 441
V_{CBO}	Collector-base voltage ($I_E = 0$)	50 V	75 V
V_{CEO} (sus)	Collector-emitter voltage ($I_B = 0$)	40 V	60 V
V_{CER}	Collector-emitter voltage ($R_{BE} \leq 100 \Omega$)	50 V	75 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)		5 V
I_{CM}	Collector peak current	2 A	
I_{BM}	Base peak current	1 A	
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ C$	1 W	
	at $T_{case} \leq 25^\circ C$		10 W
T_{stg}	Storage temperature	-65 to 200 °C	
T_j	Junction temperature	200 °C	

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 440 BC 441

THERMAL DATA

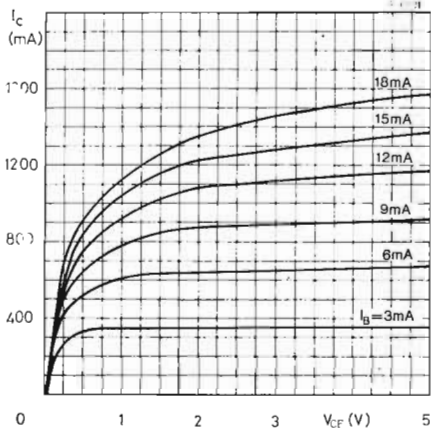
$R_{th\ j-case}$	Thermal resistance junction-case	max	17.5	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	175	°C/W

ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

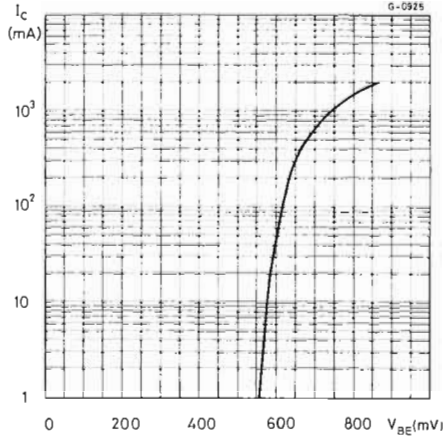
Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 40\text{ V}$			100	nA
I_{CER} Collector cutoff current ($R_{BE} = 100\ \Omega$)	for BC 440 $V_{CE} = 50\text{ V}$ for BC 441 $V_{CE} = 70\text{ V}$			10	μA
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\ \mu\text{A}$		5		V
$V_{CEC(sus)}$ Collector-emitter voltage ($I_B = 0$)	$I_C = 100\text{ mA}$ for BC 440 for BC 441	40		60	V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = 1\text{ A}$ $I_B = 100\text{ mA}$			1	V
$V_{BE(sat)}$ Base-emitter saturation voltage	$I_C = 1\text{ A}$ $I_B = 100\text{ mA}$			1.5	V
h_{FE} DC current gain	Gr. 4 $I_C = 500\text{ mA}$ $V_{CE} = 4\text{ V}$ Gr. 5 $I_C = 500\text{ mA}$ $V_{CE} = 4\text{ V}$ Gr. 6 $I_C = 500\text{ mA}$ $V_{CE} = 4\text{ V}$ $I_C = 1\text{ A}$ $V_{CE} = 2\text{ V}$ (for BC 440 only)	40		70	—
h_{FE1}/h_{FE2} Matched pair ratio	$I_C = 500\text{ mA}$ $V_{CE} = 4\text{ V}$			1.4	—
f_T Transition frequency	$I_C = 50\text{ mA}$ $V_{CE} = 4\text{ V}$		50		MHz

BC 440 BC 441

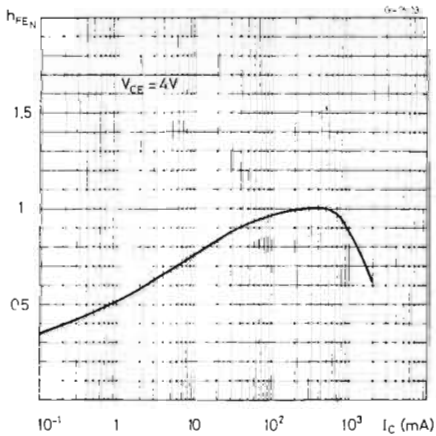
Typical output characteristics



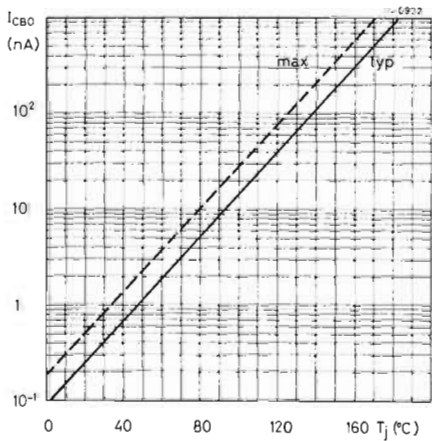
Typical DC transconductance



Typical DC normalized current gain

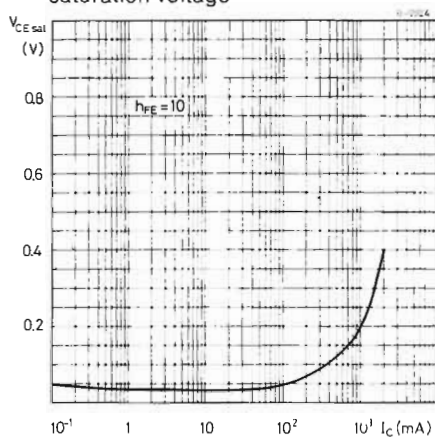


Collector cutoff current



BC 440 BC 441

Typical collector-emitter
saturation voltage



SILICON PLANAR PNP

BC 460 BC 461

MEDIUM POWER AMPLIFIER

The BC 460 and BC 461 are silicon planar epitaxial PNP transistors in TO-39 metal case. They are intended for general purpose applications, especially for driver stages.

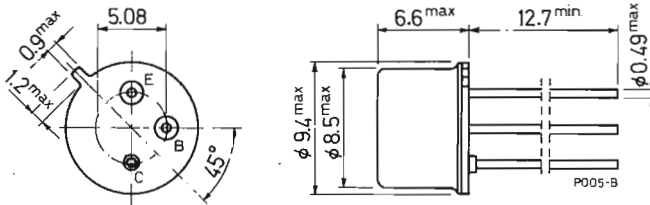
The complementary NPN types are respectively the BC 440 and BC 441.

ABSOLUTE MAXIMUM RATINGS

		BC 460	BC 461
V_{CBO}	Collector-base voltage ($I_E = 0$)	-50 V	-75 V
$V_{CEO (sus)}$	Collector-emitter voltage ($I_B = 0$)	-40 V	-60 V
V_{CER}	Collector-emitter voltage ($R_{BE} \leq 100 \Omega$)	-50 V	-75 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)		-5 V
I_{CM}	Collector peak current		-2 A
I_{BM}	Base peak current		-1 A
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ C$		1 W
	at $T_{case} \leq 25^\circ C$		10 W
T_{stg}	Storage temperature		-65 to 200 °C
T_j	Junction temperature		200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BC 460 BC 461

THERMAL DATA

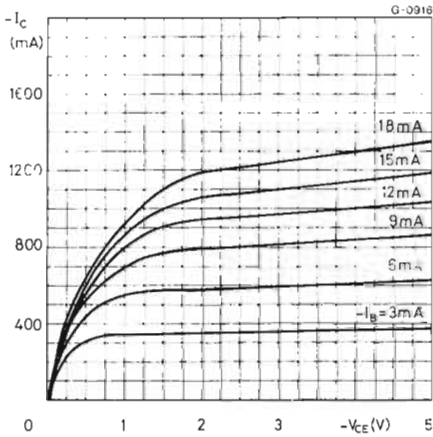
$R_{th\ j-case}$	Thermal resistance junction-case	max	17.5	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	175	°C/W

ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

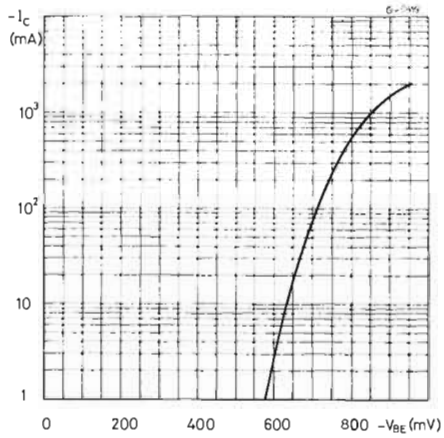
Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -40\text{ V}$			-100	nA
I_{CER} Collector cutoff current ($R_{BE} = 100\ \Omega$)	for BC 460 $V_{CE} = -50\text{ V}$ for BC 461 $V_{CE} = -70\text{ V}$			-10 -10	μA μA
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -100\ \mu\text{A}$			-5	V
$V_{CEO(sus)}$ Collector-emitter voltage ($I_B = 0$)	$I_C = -100\text{ mA}$ for BC 460 for BC 461	-40 -60			V V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = -1\text{ A}$ $I_B = -100\text{ mA}$			-1	V
$V_{BE(sat)}$ Base-emitter saturation voltage	$I_C = -1\text{ A}$ $I_B = -100\text{ mA}$			-1.5	V
h_{FE} DC current gain	Gr. 4 $I_C = -500\text{ mA}$ $V_{CE} = -4\text{ V}$ Gr. 5 $I_C = -500\text{ mA}$ $V_{CE} = -4\text{ V}$ Gr. 6 $I_C = -500\text{ mA}$ $V_{CE} = -4\text{ V}$ $I_C = -1\text{ A}$ $V_{CE} = -2\text{ V}$ (for BC 460 only)		40 60 115 20	70 130 250	— — — —
$h_{FE1}^{h_{FE2}}$ Matched pair ratio	$I_C = -500\text{ mA}$ $V_{CE} = -4\text{ V}$			1.4	—
f_T Transition frequency	$I_C = -50\text{ mA}$ $V_{CE} = -4\text{ V}$		50		MHz

BC 460 BC 461

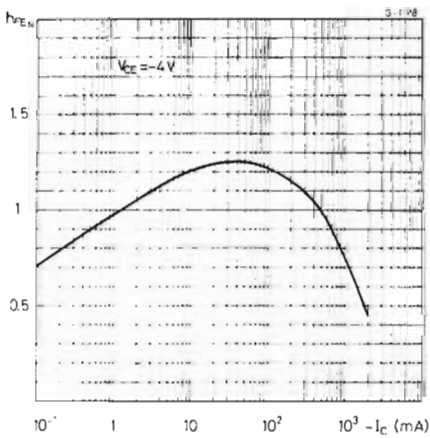
Typical output characteristics



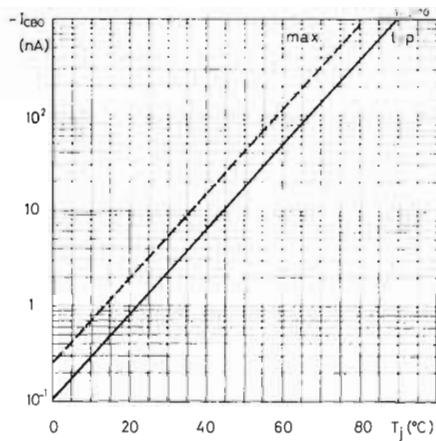
Typical DC transconductance



Typical DC normalized current gain

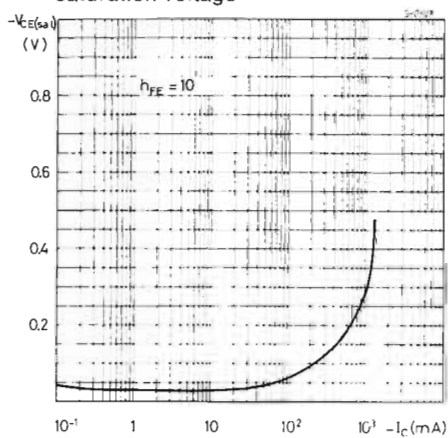


Collector cutoff current



BC 460 BC 461

Typical collector-emitter
saturation voltage



SILICON PLANAR PNP

BC 477
BC 478
BC 479

LOW NOISE GENERAL PURPOSE AUDIO AMPLIFIERS

The BC 477, BC 478 and BC 479 are silicon planar epitaxial PNP transistors in TO-18 metal case.

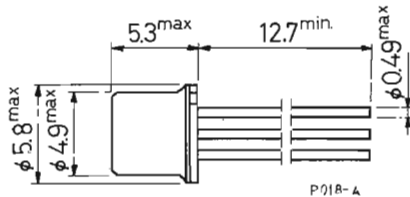
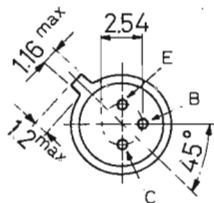
The BC 477 is a high voltage type designed for use in audio amplifiers or driver stages, and in the signal processing circuits of TV sets. The BC 478 and BC 479 are respectively low noise and very low noise types, designed for general preamplifier or amplifier applications.

ABSOLUTE MAXIMUM RATINGS

		BC 477	BC 478	BC 479
V_{CES}	Collector-emitter voltage ($V_{BE} = 0$)	-90 V	-50 V	-40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-80 V	-50 V	-40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-6 V		
I_C	Collector current	-150 mA		
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$ at $T_{case} \leq 25^\circ\text{C}$	0.36 W		
		1.2 W		
T_{stg}	Storage temperature	-55 to 200°C		
T_j	Junction temperature	200 $^\circ\text{C}$		

MECHANICAL DATA

Dimensions in mm



(sim. to TO-18)

BC 477
BC 478
BC 479

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	146	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	480	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	for BC 477 $V_{CE} = -70\text{ V}$ $V_{CE} = -70\text{ V}$ $T_{amb} = 125\text{ °C}$ for BC 478 $V_{CE} = -40\text{ V}$ $V_{CE} = -40\text{ V}$ $T_{amb} = 125\text{ °C}$ for BC 479 $V_{CE} = -30\text{ V}$ $V_{CE} = -30\text{ V}$ $T_{amb} = 125\text{ °C}$			-10 -10 -10 -10 -10 -10	nA μA nA μA nA μA
I_{EBO} Emitter cutoff current ($I_C = 0$)	$V_{EB} = -4\text{ V}$			-10	nA
$V_{(BR)CES}$ Collector-emitter breakdown voltage ($V_{BE} = 0$)	$I_C = -10\ \mu\text{A}$ for BC 477 for BC 478 for BC 479	-90 -50 -40			V V V
$V_{(BR)CEO}$ *Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -5\text{ mA}$ for BC 477 for BC 478 for BC 479	-80 -50 -40			V V V
$V_{(BR)EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$			-6	V

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BC 477
BC 478
BC 479

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{CE(sat)}$ * Collector-emitter saturation voltage	$I_C = -10\text{ mA}$ $I_B = -0.5\text{ mA}$ $I_C = -100\text{ mA}$ $I_B = -5\text{ mA}$		-0.1	-0.25	V
V_{BE} * Base-emitter voltage	$I_C = -2\text{ mA}$ $V_{CE} = -5\text{ V}$	-0.55	-0.65	-0.75	V
$V_{BE(sat)}$ * Base-emitter saturation voltage	$I_C = -10\text{ mA}$ $I_B = -0.5\text{ mA}$ $I_C = -100\text{ mA}$ $I_B = -5\text{ mA}$		-0.75	-0.9	V
h_{FE} * DC current gain	$I_C = -10\text{ }\mu\text{A}$ $V_{CE} = -5\text{ V}$ for BC 477 for BC 477 Gr. VI for BC 477 Gr. A for BC 478 for BC 478 Gr. A for BC 478 Gr. B for BC 479 for BC 479 Gr. B $I_C = -2\text{ mA}$ $V_{CE} = -5\text{ V}$ for BC 477 for BC 477 Gr. VI for BC 477 Gr. A for BC 478 for BC 478 Gr. A for BC 478 Gr. B for BC 479 for BC 479 Gr. B $I_C = -10\text{ mA}$ $V_{CE} = -5\text{ V}$ for BC 477 for BC 477 Gr. VI for BC 477 Gr. A for BC 478 for BC 478 Gr. A for BC 478 Gr. B for BC 479 for BC 479 Gr. B		30 30 50 50 50 100 100 100 70 70 110 110 110 220 220 220 160 100 180 270 180 350 400 350	115 70 130 195 130 250 290 250 250 130 250 450 250 450 450 250 180 180 350 400 350	— —

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

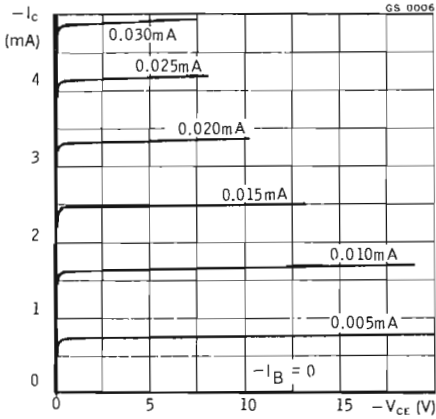
BC 477
BC 478
BC 479

ELECTRICAL CHARACTERISTICS (continued)

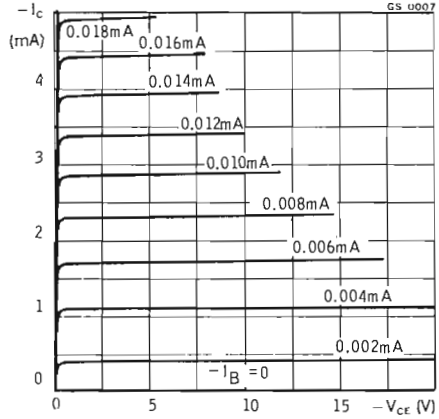
Parameter	Test conditions	Min.	Typ.	Max.	Unit
h_{fe} Small signal current gain	$I_C = -2 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 1 \text{ kHz}$ for BC 477	75	260	—	
	for BC 477 Gr. VI for BC 477 Gr. A	75 125	150 260	— —	
	for BC 478	125	500	—	
	for BC 478 Gr. A for BC 478 Gr. B	125 240	260 500	— —	
	for BC 479	240	—	—	
	for BC 479 Gr. B	240	500	—	
	$I_C = -10 \text{ mA}$ $V_{CE} = -5 \text{ V}$ $f = 20 \text{ MHz}$		7.5	—	
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = -5 \text{ V}$		4	6	pF
C_{EBO} Emitter-base capacitance	$I_C = 0$ $V_{EB} = -0.5 \text{ V}$		11	15	pF
NF Noise figure	$I_C = -20 \mu\text{A}$ $V_{CE} = -5 \text{ V}$ $R_g = 10 \text{ k}\Omega$ $f = 10 \text{ Hz to } 10 \text{ kHz}$ $B = 15.7 \text{ kHz}$ for BC 479		0.8	3.5	dB
	$I_C = -200 \mu\text{A}$ $V_{CE} = -5 \text{ V}$ $R_g = 2 \text{ k}\Omega$ $f = 10 \text{ Hz to } 10 \text{ kHz}$ $B = 15.7 \text{ kHz}$ for BC 478 for BC 479		1.5 1	4	dB dB
	$I_C = -20 \mu\text{A}$ $V_{CE} = -5 \text{ V}$ $R_g = 10 \text{ k}\Omega$ $f = 1 \text{ kHz}$ $B = 200 \text{ Hz}$ for BC 479		0.5	2.5	dB
	$I_C = -200 \mu\text{A}$ $V_{CE} = -5 \text{ V}$ $R_g = 2 \text{ k}\Omega$ $f = 1 \text{ kHz}$ $B = 200 \text{ Hz}$ for BC 477		2	10	dB
	for BC 478		1.2	6	dB
	for BC 479		0.8	4	dB

BC 477 BC 478 BC 479

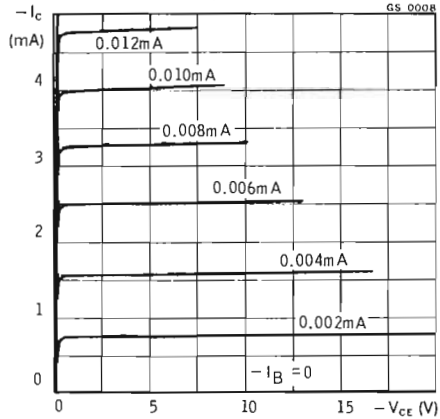
Typical output characteristics
(for BC 477 only)



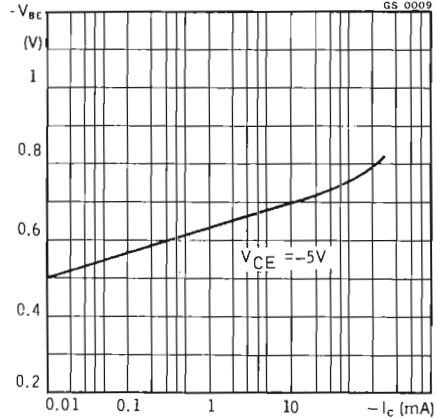
Typical output characteristics
(for BC 478 only)



Typical output characteristics
(for BC 479 only)

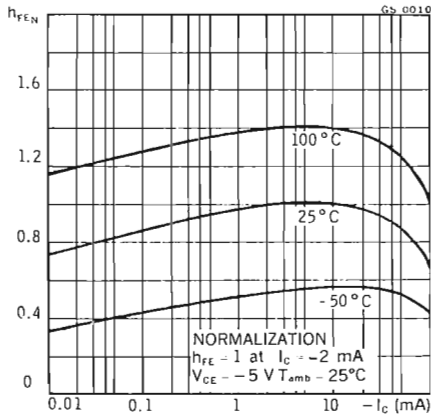


DC transconductance

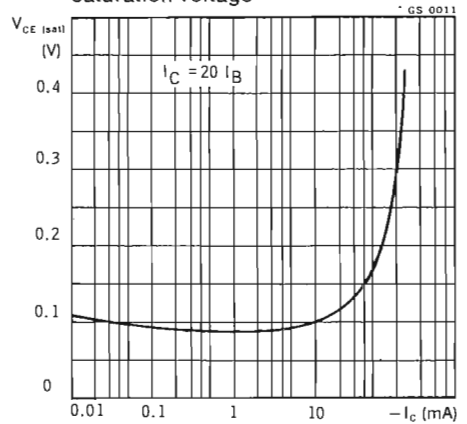


BC 477 BC 478 BC 479

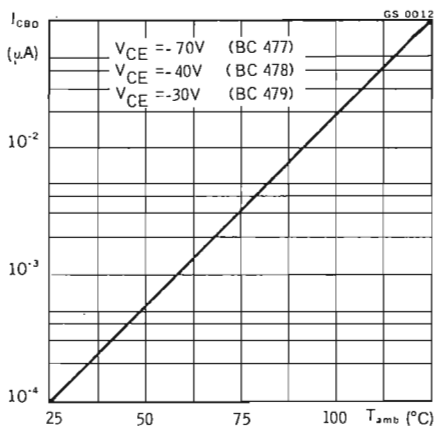
DC normalized current gain



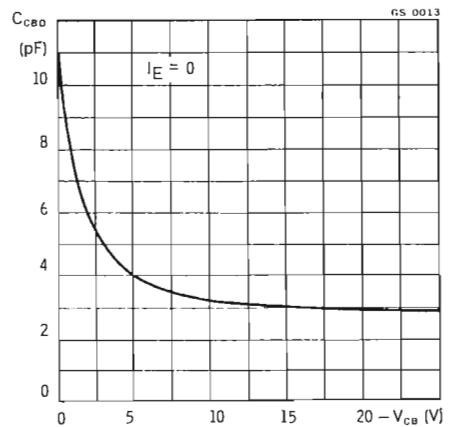
Typical collector-emitter saturation voltage



Typical collector cutoff current

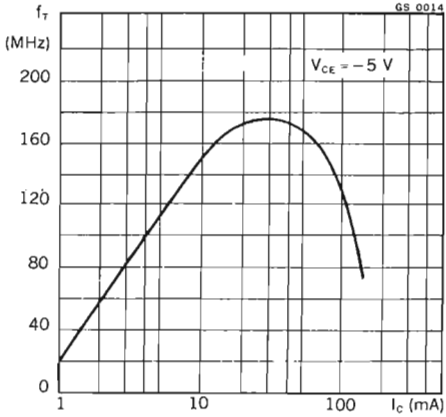


Typical collector-base capacitance

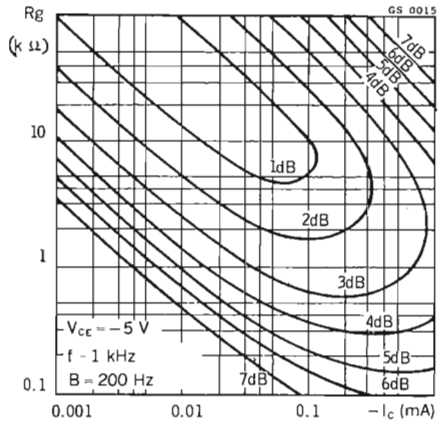


BC 477 BC 478 BC 479

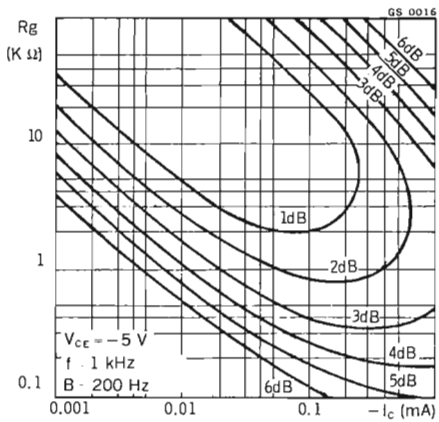
Transition frequency



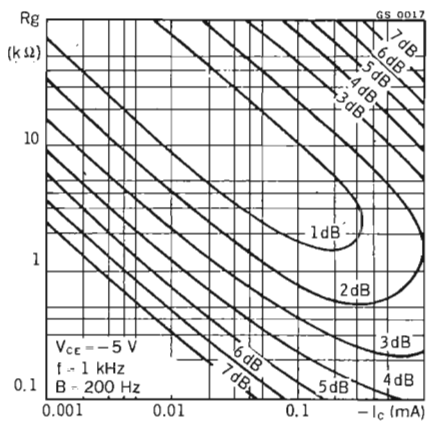
Noise figure (for BC 477 only)



Noise figure (for BC 478 only)

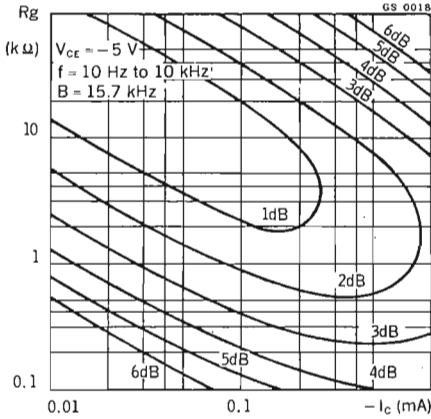


Noise figure (for BC 479 only)

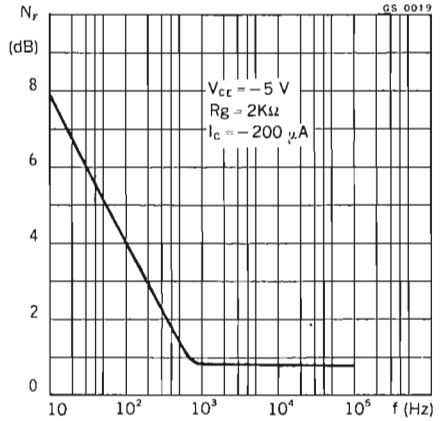


BC 477
BC 478
BC 479

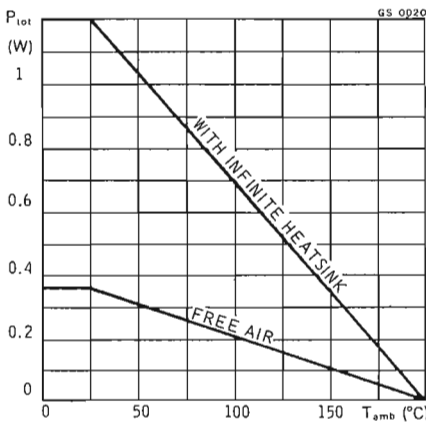
Noise figure (for **BC 479** only)



Noise figure (for **BC 479** only)



Power rating chart



BF 155

SILICON PLANAR NPN

UHF AMPLIFIER AND MIXER-OSCILLATOR

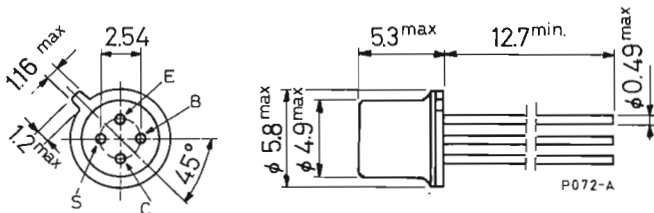
The BF 155 is a silicon planar NPN transistor in a TO-72 metal case. It is specifically designed for UHF amplifier and mixer-oscillator applications.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	3 V
I_C	Collector current	20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200 mW
	at $T_{case} \leq 25^\circ\text{C}$	300 mW
T_{stg}	Storage temperature	-55 to 200 °C
T_j	Junction temperature	200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 155

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	580 °C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	875 °C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 10\text{ V}$			100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$	40			V
$V_{(BR)\ CEO}$ *Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 5\text{ mA}$	40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\ \mu\text{A}$	3			V
V_{BE} Base-emitter voltage	$I_C = 2.5\text{ mA}$ $V_{CE} = 12\text{ V}$			0.85	V
h_{FE} * DC current gain	$I_C = 2.5\text{ mA}$ $V_{CE} = 12\text{ V}$	20	70		—
f_T Transition frequency	$I_C = 2.5\text{ mA}$ $V_{CE} = 12\text{ V}$	400	600		MHz
$-C_{re}$ Feedback capacitance	$I_C = 2.5\text{ mA}$ $V_{CE} = 12\text{ V}$ $f = 1\text{ MHz}$		0.4		pF
NF Noise figure	$I_C = 2.5\text{ mA}$ $V_{CB} = 12\text{ V}$ $R_g = 50\ \Omega$ $f = 800\text{ MHz}$		7	9	dB
G_{pb} Power gain	$I_C = 2.5\text{ mA}$ $V_{CB} = 12\text{ V}$ $f = 800\text{ MHz}$	8	10		dB
f_{max} Maximum oscillation frequency	$I_C = 2.5\text{ mA}$ $V_{CB} = 12\text{ V}$		2.5		GHz

* Pulsed: pulse duration = 300 μs ; duty factor = 1%.

BF 158

SILICON PLANAR NPN

IF AMPLIFIER FOR TV

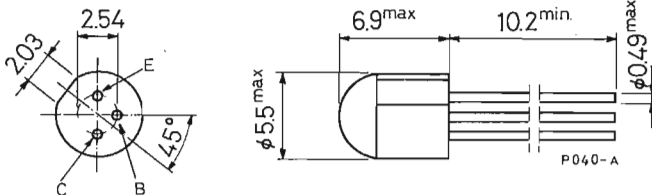
The BF 158 is a silicon planar NPN transistor in a TO-18 epoxy package. It is designed for use as IF amplifier in TV receiver.

ABSOLUTE MAXIMUM RATINGS

V_{CB0}	Collector-base voltage ($I_E = 0$)	30	V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	12	V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	2	V
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	0.2	W
	at $T_{case} \leq 25^\circ\text{C}$	0.5	W
T_{stg}	Storage temperature	-55 to 125	$^\circ\text{C}$
T_j	Junction temperature	125	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BF 158

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 15\text{ V}$			100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$	30			V
$V_{CEO(sus)}$ Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 3\text{ mA}$	12			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\ \mu\text{A}$	2			V
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_C = 10\text{ mA}$ $I_B = 1\text{ mA}$			0.5	V
h_{FE} DC current gain	$I_C = 4\text{ mA}$ $V_{CE} = 10\text{ V}$	20	50		—
f_T Transition frequency	$I_C = 5\text{ mA}$ $V_{CE} = 10\text{ V}$	700			MHz
$-C_{re}$ Feedback capacitance	$I_C = 5\text{ mA}$ $V_{CE} = 10\text{ V}$	0.8	1.2		pF
NF Noise figure	$I_C = 4\text{ mA}$ $V_{CE} = 10\text{ V}$ $R_g = 400\ \Omega$ $f = 40\text{ MHz}$		3.5		dB
G_{de} Power gain	$I_C = 5\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 40\text{ MHz}$	22	26		dB
g_{oe} Output conductance	$I_C = 5\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 40\text{ MHz}$	0.2	0.3		mS

BF 160

SILICON PLANAR NPN

IF AMPLIFIER FOR AM/FM RADIOS

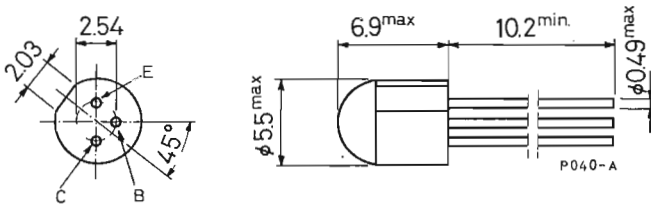
The BF 160 is a silicon planar NPN transistor in a TO-18 epoxy package. It is designed for intermediate frequency (5.5 MHz TV - 10.7 MHz FM) and for general AM-FM applications.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	12 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	2 V
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200 mW
	at $T_{case} \leq 25^\circ\text{C}$	500 mW
T_{stg}	Storage temperature	-55 to 125 °C
T_J	Junction temperature	125 °C

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BF 160

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	200	$^{\circ}C/W$
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	$^{\circ}C/W$

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25^{\circ}C$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 15\ V$ $V_{CB} = 15\ V$ $T_{amb} = 65^{\circ}C$			100 5	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu A$	30			V
$V_{(BR)\ CEO}$ *Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 3\ mA$	12			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\ \mu A$	2			V
h_{FE}^* DC current gain	$I_C = 3\ mA$ $V_{CE} = 10\ V$	20	50		—
f_T Transition frequency	$I_C = 3\ mA$ $V_{CE} = 10\ V$	400	600		MHz
$-C_{re}$ Feedback capacitance	$I_C = 3\ mA$ $V_{CE} = 10\ V$		0.8	1.2	pF
G_{pe} Power gain	$I_C = 3\ mA$ $V_{CE} = 8\ V$ $f = 10.7\ MHz$	28	32		dB

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BF 161

SILICON PLANAR NPN

UHF AMPLIFIER, OSCILLATOR AND MIXER

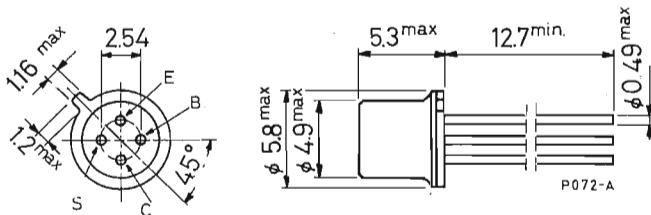
The BF 161 is a silicon planar NPN transistor in a TO-72 metal case, intended for UHF tuner applications.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	50 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	50 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	175 mW
	at $T_{case} \leq 25^\circ\text{C}$	260 mW
T_{stg}	Storage temperature	-55 to 175 °C
T_j	Junction temperature	175 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 161

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	580	°C/W
------------------	----------------------------------	-----	-----	------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 10\text{ V}$			100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 50\ \mu\text{A}$	50			V
$V_{CEO\ (sus)}$ Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 5\text{ mA}$	50			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 50\ \mu\text{A}$	5			V
V_{BE} Base-emitter voltage	$I_C = 3\text{ mA}$ $V_{CE} = 24\text{ V}$		0.74		V
h_{FE} DC current gain	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$	20	60		—
f_T Transition frequency	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$	400	550		MHz
$-C_{re}$ Feedback capacitance	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 1\text{ MHz}$		0.3	0.45	pF
NF Noise figure	$I_C = 1.5\text{ mA}$ $V_{CB} = 24\text{ V}$ $f = 800\text{ MHz}$		6.5		dB
G_{pb} Power gain	$I_C = 1.5\text{ mA}$ $V_{CB} = 24\text{ V}$ $f = 800\text{ MHz}$		12		dB
Collector current for $\Delta G_{db} = 30\text{ dB}$	$V_{CC} = 12\text{ V}$ $f = 800\text{ MHz}$		8		mA

SILICON PLANAR NPN

HIGH FREQUENCY GENERAL PURPOSE AMPLIFIER

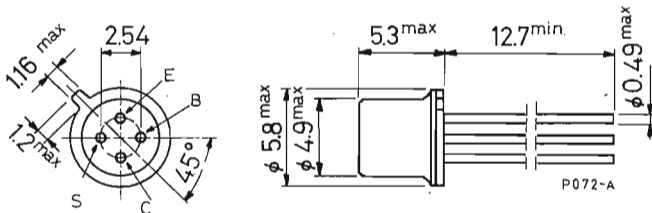
The BF 166 is a silicon planar NPN transistor in a TO-72 metal case. It is designed to be used as a gain-controlled VHF amplifier.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	3 V
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	175 mW
	at $T_{case} \leq 25^\circ\text{C}$	260 mW
T_{stg}	Storage temperature	-55 to 175 °C
T_j	Junction temperature	175 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 166

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	580 °C/W
------------------	----------------------------------	-----	----------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 10\text{ V}$			100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$	40			V
$V_{CEO(sus)}$ Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 1\text{ mA}$	40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu\text{A}$	3			V
V_{BE}^* Base-emitter voltage	$I_C = 2.5\text{ mA}$ $V_{CE} = 12\text{ V}$			0.9	V
h_{FE}^* DC current gain	$I_C = 2.5\text{ mA}$ $V_{CE} = 12\text{ V}$	20	50		—
f_T Transition frequency	$I_C = 3\text{ mA}$ $V_{CE} = 12\text{ V}$	400	500		MHz
$-C_{re}$ Feedback capacitance	$I_C = 2.5\text{ mA}$ $V_{CE} = 12\text{ V}$		0.4	0.6	pF
NF Noise figure	$I_C = 2.5\text{ mA}$ $V_{CE} = 12\text{ V}$ $R_g = 50\ \Omega$ $f = 200\text{ MHz}$		3	5	dB
G_{pe} Power gain (neutralized)	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 200\text{ MHz}$	16	18		dB
$I_C(AGC)$ Collector current for $\Delta G_{pb} = 30\text{ dB}$	$V_{CC} = 12\text{ V}$ $f = 200\text{ MHz}$			14	mA

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BF 167

SILICON PLANAR NPN

TV AGC IF AMPLIFIER

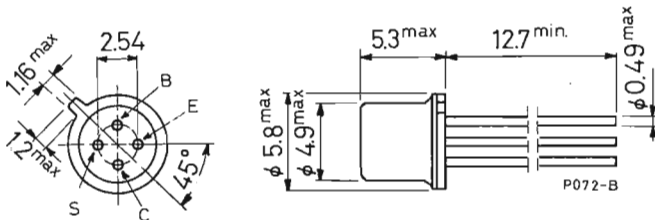
The BF 167 is a silicon planar NPN transistor in a TO-72 metal case. It is particularly designed for use in forward AGC IF amplifiers of TV receivers. It is characterized by very low feedback capacitance due to a screening diffusion under the base pad.

ABSOLUTE MAXIMUM RATINGS

V_{CES}	Collector-emitter voltage ($I_{BE} = 0$)	40	V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	30	V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4	V
I_C	Collector current	25	mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	150	mW
T_{stg}	Storage temperature	-55 to 175	$^\circ\text{C}$
T_J	Junction temperature	175	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 167

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	1000	°C/W
-----------------	-------------------------------------	-----	------	------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 10\text{ V}$ $V_{CE} = 10\text{ V}$ $T_{amb} = 100\text{ °C}$			50 5	nA μA
$V_{(BR)\ CES}$ Collector-emitter breakdown voltage ($V_{BE} = 0$)	$I_C = 10\ \mu\text{A}$	40			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 5\text{ mA}$	30			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu\text{A}$	4			V
V_{BE}^* Base-emitter voltage	$I_C = 4\text{ mA}$ $V_{CE} = 10\text{ V}$		0.74		V
h_{FE}^* DC current gain	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 4\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$		30	35 45 20	— — —
f_T Transition frequency	$I_C = 4\text{ mA}$ $V_{CE} = 10\text{ V}$		600		MHz
$-C_{re}$ Feedback capacitance	$I_C = 0$ $V_{CE} = 10\text{ V}$ $f = 1\text{ MHz}$		0.15		pF
NF Noise figure	$I_C = 4\text{ mA}$ $V_{CE} = 10\text{ V}$ $R_g = 100\ \Omega$ $f = 36\text{ MHz}$		3		dB
G_{de}^{**} Power gain	$I_E = 4\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 36\text{ MHz}$	24	28		dB

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

** See test circuit.

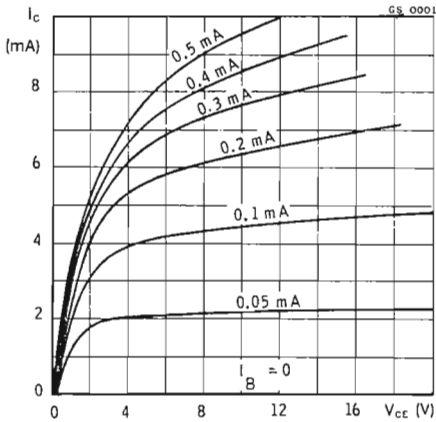
BF 167

ELECTRICAL CHARACTERISTICS (continued)

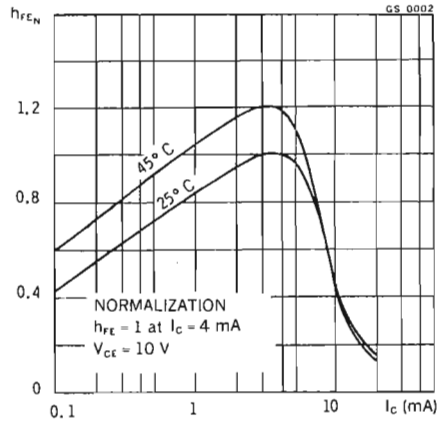
Parameter	Test conditions	Min.	Typ.	Max.	Unit
ΔG_{pe} Power gain control	$V_{EE} = -25 \text{ V}$ $R_{EE} = 3.9 \text{ k}\Omega$ $f = 36 \text{ MHz}$		60		dB
g_{ie} Input conductance	$I_C = 4 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		3.8		mS
b_{ie} Input susceptance	$I_C = 4 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		5		mS
g_{fe} Forward transconductance	$I_C = 4 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		95		mS
b_{fe} Forward transusceptance	$I_C = 4 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		34		mS
g_{oe} Output conductance	$I_C = 4 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		62		μS
b_{oe} Output susceptance	$I_C = 4 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		270		μS

BF 167

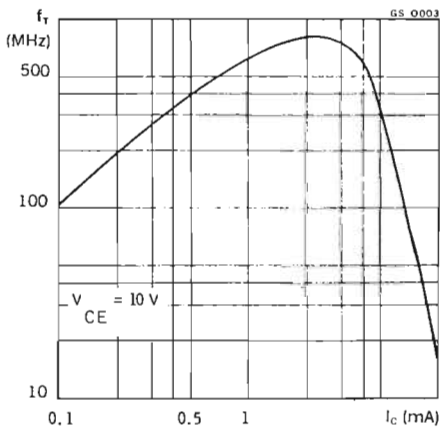
Typical output characteristics



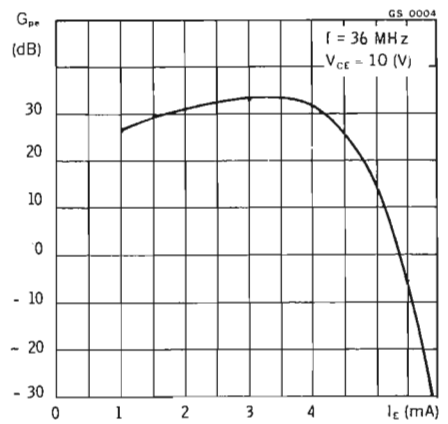
DC normalized current gain



Transition frequency

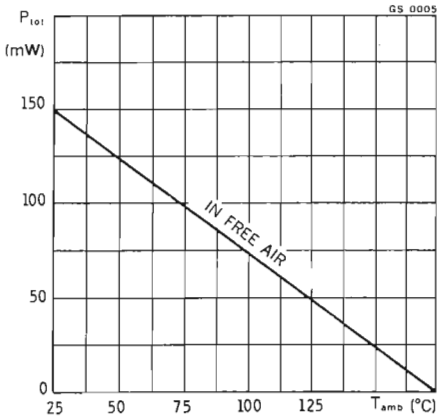


Power gain



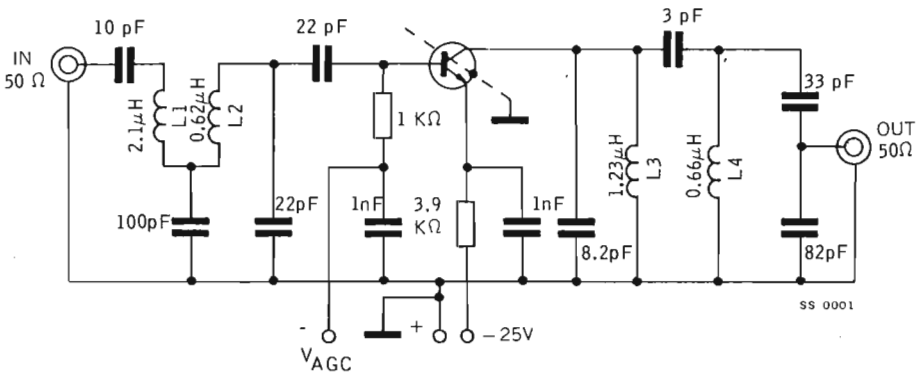
BF 167

Power rating chart



TEST CIRCUIT

Power gain ($f = 36$ MHz)



BF 173

SILICON PLANAR NPN

VIDEO IF AMPLIFIER

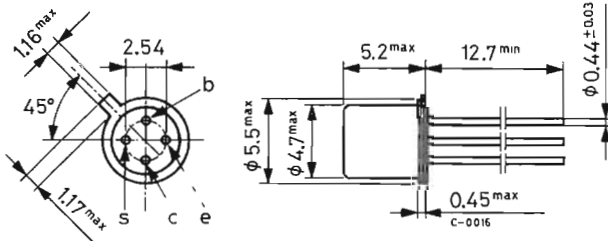
The BF 173 is a silicon planar epitaxial NPN transistor in a Jedec TO-72 metal case with a very low feedback capacitance. This transistor is intended for use in video IF amplifiers, particularly for the output stage.

ABSOLUTE MAXIMUM RATINGS

V_{CB0}	Collector-base voltage ($I_E = 0$)	40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	25 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	25 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	175 mW
	at $T_{case} \leq 25^\circ\text{C}$	230 mW
T_{stg}	Storage temperature	-55 to 175 °C
T_j	Junction temperature	175 °C

MECHANICAL DATA

Dimensions in mm



TO-72

BF 173

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	850	°C/W
-----------------	-------------------------------------	-----	-----	------

ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 20\text{ V}$			20	nA
I_{EBO} Emitter cutoff current ($I_C = 0$)	$V_{EB} = 4\text{ V}$			100	μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\text{ μA}$	40			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 2\text{ mA}$	25			V
V_{BE} Base-emitter voltage	$I_C = 7\text{ mA}$ $V_{CE} = 10\text{ V}$			0.9	V
f_T Transition frequency	$I_C = 5\text{ mA}$ $V_{CE} = 10\text{ V}$		1000		MHz
$-C_{re}$ Reverse capacitance	$I_C = 5\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 0.5\text{ MHz}$		0.23		pF
I_B Base current	$I_C = 7\text{ mA}$ $V_{CE} = 10\text{ V}$		61	185	μA
V_o^* Output voltage	$I_C = 7.2\text{ mA}$ $V_{CE} = 12\text{ V}$ $f = 38.9\text{ MHz}$	6	7.7		V
G_{tr} Transducer power gain	$I_C = 7.2\text{ mA}$ $V_{CE} = 12\text{ V}$ $f = 36.4\text{ MHz}$		26		dB
g_{ie} Input conductance	$I_C = 7\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 35\text{ MHz}$		3		mS
C_{ie} Input capacitance	$I_C = 7\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 35\text{ MHz}$		22		pF

* Voltage across the detector load $R_L = 2.7\text{ k}\Omega$ for 30% synchronisation pulse compression.

BF 173

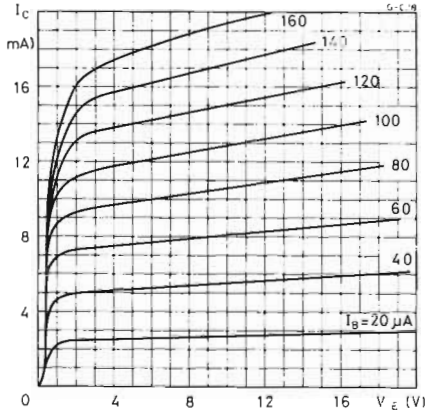
ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min. Typ. Max.	Unit
$ y_{re} $ Reverse transadmittance	$I_C = 7 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 35 \text{ MHz}$	55	μS
φ_{re} Phase angle of reverse transadmittance	$I_C = 7 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 35 \text{ MHz}$	267°	—
$ y_{fe} $ Forward transadmittance	$I_C = 7 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 35 \text{ MHz}$	165	mS
φ_{fe} Phase angle of forward transadmittance	$I_C = 7 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 35 \text{ MHz}$	336°	—
g_{oe} Output conductance	$I_C = 7 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 35 \text{ MHz}$	65	μS
C_{oe} Output capacitance	$I_C = 7 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 35 \text{ MHz}$	1.9	pF
G_{UM}^* Maximum unilateralized power gain	$I_C = 7 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 35 \text{ MHz}$	44.5	dB

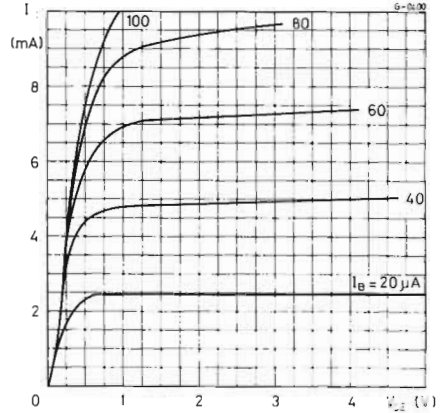
$$* G_{UM} = 10 \log \frac{|y_{fe}|^2}{4 g_{ie} g_{oe}}$$

BF 173

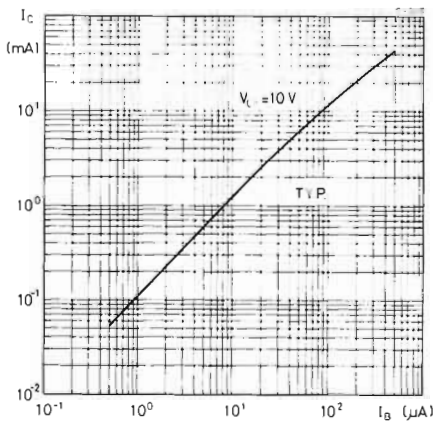
Typical output characteristics



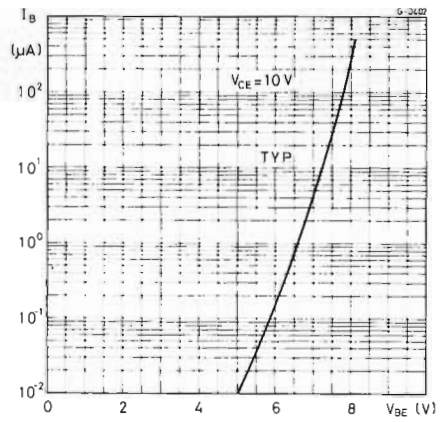
Typical output characteristics



Collector characteristic

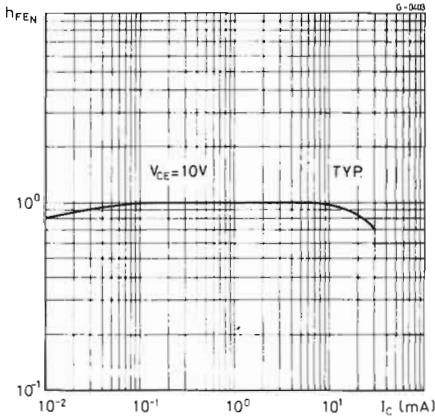


Input characteristic

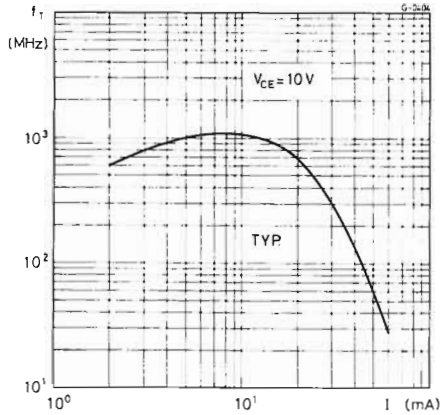


BF 173

DC normalized current gain

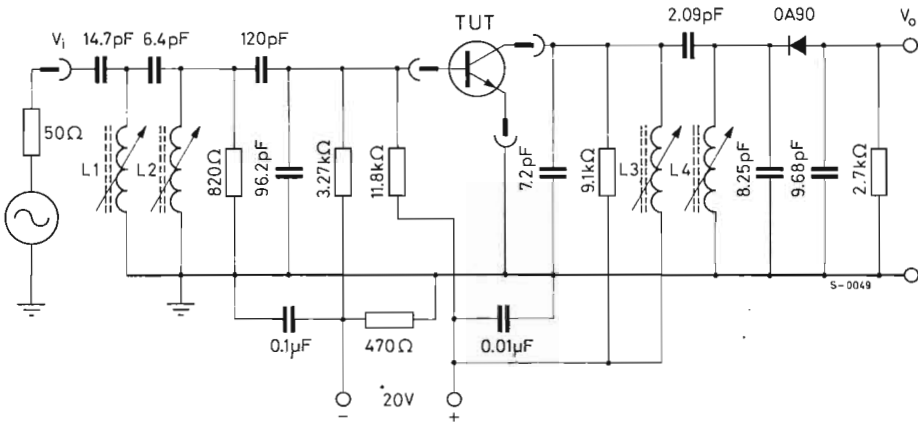


Transition frequency



TEST CIRCUIT

G_{tr} test circuit



$L_1 = 0.8 \mu\text{H}$, 9 turns $\varnothing 0.15$ mm. enameled silk-covered copper wire. $L_2 = 0.25 \mu\text{H}$, 4 turns $\varnothing 0.15$ mm. enameled silk-covered copper wire. $L_3 = 1.7 \mu\text{H}$, 12.5 turns $\varnothing 0.15$ mm. enameled silk-covered copper wire. $L_4 = 1.3 \mu\text{H}$, 11 turns $\varnothing 0.15$ mm. enameled silk-covered copper wire.

BF 222

SILICON PLANAR NPN

AMPLIFIER AND CONVERTER FOR FM TUNERS

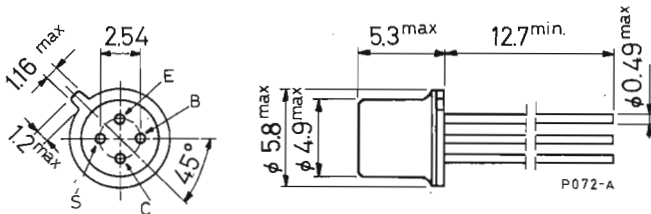
The BF 222 is a silicon planar NPN transistor in a TO-72 metal case. This device is designed for tuners of FM receivers, and features low noise, high gain and excellent forward AGC.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	50 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	175 mW
T_{stg}	Storage temperature	-55 to 175 °C
T_j	Junction temperature	175 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 222

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	875 °C/W
-----------------	-------------------------------------	-----	----------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 10\text{ V}$		1		nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 50\ \mu\text{A}$	50			V
$V_{(BR)\ CEO}^*$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 5\text{ mA}$	40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 50\ \mu\text{A}$	4			V
V_{BE} Base-emitter voltage	$I_C = 2\text{ mA}$ $V_{CE} = 7\text{ V}$		0.74		V
h_{FE} DC current gain	$I_C = 2\text{ mA}$ $V_{CE} = 7\text{ V}$	20	60		—
f_T Transition frequency	$I_C = 2\text{ mA}$ $V_{CE} = 7\text{ V}$		400		MHz
$-C_{re}$ Feedback capacitance	$I_C = 2\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 1\text{ MHz}$		0.4		pF
NF Noise figure	$I_C = 4\text{ mA}$ $V_{CE} = 5\text{ V}$ $R_g = 150\ \Omega$ $f = 100\text{ MHz}$			5	dB
G_{pe} Power gain	$I_C = 4\text{ mA}$ $V_{CE} = 5\text{ V}$ $f = 100\text{ MHz}$		20		dB
ΔG_{pe} Power gain control	$I_E = 9\text{ mA}$ $V_{CC} = 7\text{ V}$ $R_{DC} = 510\ \Omega$ $f = 100\text{ MHz}$		30		dB

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

SILICON PLANAR NPN

BF 233 BF 234

AM MIXER OSCILLATOR, AM-FM IF AMPLIFIER

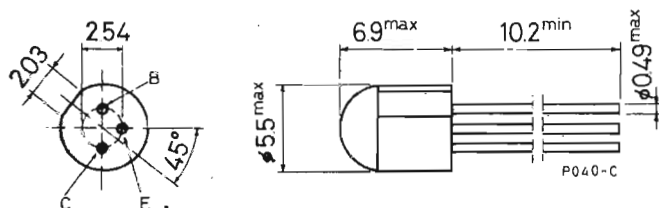
The BF 233 and BF 234 are silicon planar epitaxial NPN transistors in TO-18 epoxy package. They are intended for use in AM mixer/oscillator stages, IF amplifiers for AM/FM radio receivers and in sound IF stages for TV receivers.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	30 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	50 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200 mW
T_{stg}	Storage temperature	-55 to 125 $^\circ\text{C}$
T_j	Junction temperature	125 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BF 233 BF 234

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W
-----------------	-------------------------------------	-----	-----	------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit		
I_{CBO}	Collector cutoff current ($I_E = 0$)			200	nA		
$V_{(BR)\ CBO}$	Collector-base breakdown voltage ($I_E = 0$)		30		V		
$V_{CEO\ (sus)}$	Collector-emitter sustaining voltage ($I_B = 0$)		30		V		
V_{EBO}	Emitter-base voltage ($I_C = 0$)		4		V		
V_{BE}	Base-emitter voltage	$I_C = 1\text{ mA}$	$V_{CE} = 10\text{ V}$	0.64	0.7	0.74	V
h_{FE}	DC current gain	$I_C = 1\text{ mA}$	$V_{CE} = 10\text{ V}$				
	for BF 233 Gr. 2			40	60	70	—
	for BF 233 Gr. 3			60	80	100	—
	for BF 233 Gr. 4			90	115	150	—
	for BF 233 Gr. 5			140	175	220	—
	for BF 233 Gr. 6			200	245	350	—
	for BF 234			90	120	330	—
f_T	Transition frequency	$I_C = 1\text{ mA}$	$V_{CE} = 10\text{ V}$	150	500		MHz
$-C_{re}$	Feedback capacitance	$I_C = 1\text{ mA}$	$V_{CE} = 10\text{ V}$		0.5	1	pF
NF	Noise figure	$I_C = 1\text{ mA}$	$V_{CE} = 10\text{ V}$		1.2		dB
		$R_g = 300\ \Omega$					
		$f = 470\text{ kHz}$					
NF_C	Conversion noise figure	$I_C = 1\text{ mA}$	$V_{CE} = 10\text{ V}$		4		dB
		$R_g = 500\ \Omega$			3.5		dB
		$f = 200\text{ kHz}$					
		$f = 1\text{ MHz}$					
G_{pe}	Power gain	$I_C = 1\text{ mA}$	$V_{CE} = 10\text{ V}$		40		dB
		$f = 470\text{ kHz}$					

BF 233

BF 234

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
g_{ie} Input conductance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$ $f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		0.24 0.28 0.30		mS mS mS
b_{ie} Input susceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$ $f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		22 260 500		μS μS μS
b_{re} Reverse transusceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$ $f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		-1.6 -17 -34		μS μS μS
φ_{re} Phase angle of reverse transadmittance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$ $f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		-90° -90° -90°		— — —
g_{fe} Forward transconductance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$ $f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		35 35 35		mS mS mS
b_{fe} Forward transusceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$ $f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		0 -0.5 -1		mS mS mS
g_{oe} Output conductance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$ $f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		7 9 11		μS μS μS
b_{oe} Output susceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$ $f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		4.4 52 100		μS μS μS

SILICON PLANAR NPN

BF 257
BF 258
BF 259

HIGH VOLTAGE VIDEO AMPLIFIERS

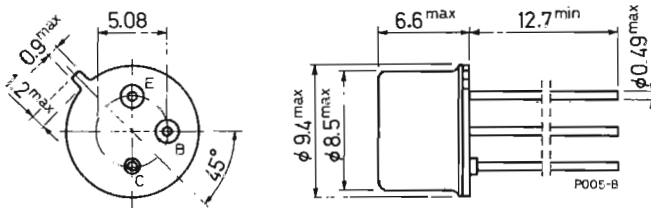
The BF 257, BF 258 and BF 259 are silicon planar epitaxial NPN transistors in TO-39 metal case. They are particularly designed for video output stages in CTV and MTV sets, class A audio output stages and drivers for horizontal deflection circuits.

ABSOLUTE MAXIMUM RATINGS

		BF 257	BF 258	BF 259
V_{CBO}	Collector-base voltage ($I_E = 0$)	160 V	250 V	300 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	160 V	250 V	300 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	5 V		
I_C	Collector current	100 mA		
I_{CM}	Collector peak current	200 mA		
P_{tot}	Total power dissipation at $T_{case} \leq 50^\circ C$	5 W		
T_{stg}	Storage temperature	-55 to 200 °C		
T_j	Junction temperature	200 °C		

MECHANICAL DATA

Dimensions in mm



(sim. to TO-39)

BF 257
BF 258
BF 259

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	30 °C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	175 °C/W

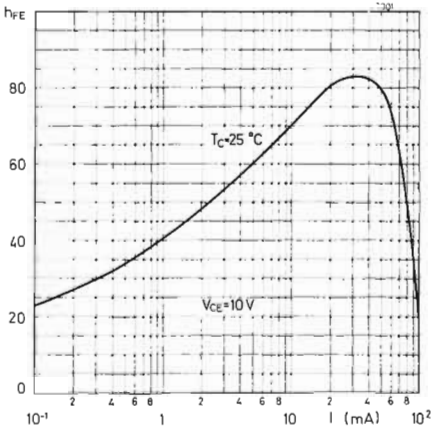
ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	for BF 257 $V_{CB} = 100\text{ V}$ for BF 258 $V_{CB} = 200\text{ V}$ for BF 259 $V_{CB} = 250\text{ V}$			50 50 50	nA nA nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\text{ }\mu\text{A}$ for BF 257 for BF 258 for BF 259	160 250 300			V V V
$V_{(BR)\ CEO}$ * Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 10\text{ mA}$ for BF 257 for BF 258 for BF 259	160 250 300			V V V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\text{ }\mu\text{A}$	5			V
$V_{CE(sat)}$ * Collector-emitter saturation voltage	$I_C = 30\text{ mA}$ $I_B = 6\text{ mA}$			1	V
h_{FE} * DC current gain	$I_C = 30\text{ mA}$ $V_{CE} = 10\text{ V}$	25			—
f_T Transition frequency	$I_C = 15\text{ mA}$ $V_{CE} = 10\text{ V}$		90		MHz
$-C_{re}$ Feedback capacitance	$I_C = 0$ $V_{CE} = 30\text{ V}$ $f = 1\text{ MHz}$		3		pF

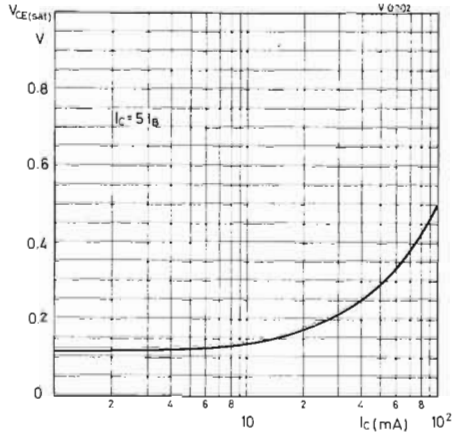
* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BF 257 BF 258 BF 259

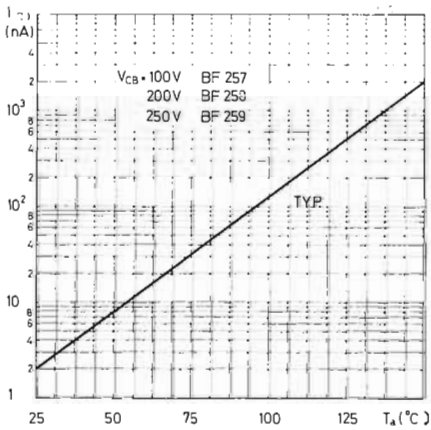
DC current gain



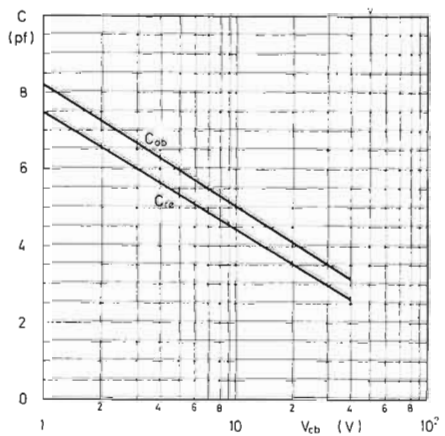
Collector-emitter saturation voltage



Collector cutoff current

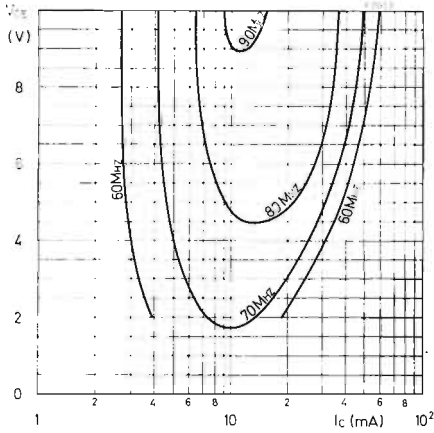


Collector-base capacitance

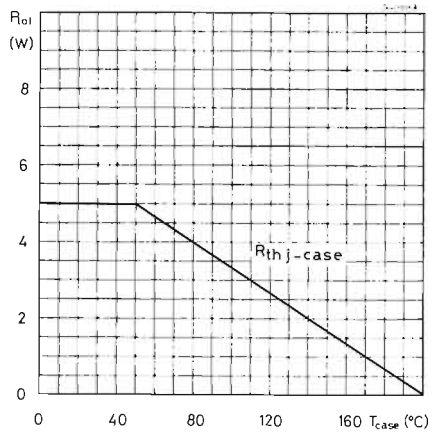


BF 257 BF 258 BF 259

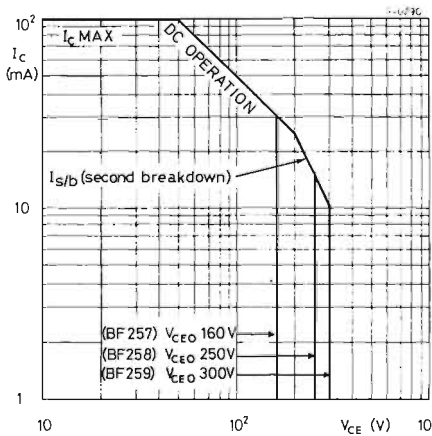
Transition frequency



Power rating chart



Safe operating area



BF 260

SILICON PLANAR NPN

AGC VHF AMPLIFIER

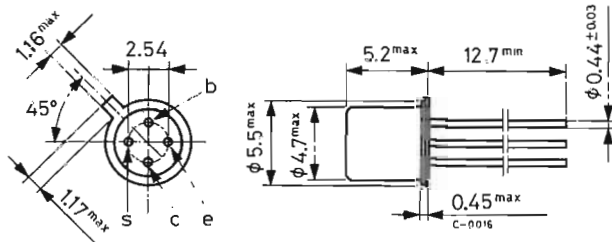
The BF 260 is a silicon planar NPN transistor in a Jeduc TO-72 metal case, with a very low feedback capacitance. It is intended primarily for use as RF amplifier in television tuners up to 260 MHz.

ABSOLUTE MAXIMUM RATINGS

V_{CB0}	Collector-base voltage ($I_E = 0$)	45 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	30 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	50 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	150 mW
	at $T_{case} \leq 25^\circ\text{C}$	230 mW
T_{stg}	Storage temperature	-65 to 175 $^\circ\text{C}$
T_j	Junction temperature	175 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-72

BF 260

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max 1000 °C/W
-----------------	-------------------------------------	---------------

ELECTRICAL CHARACTERISTICS ($T_{case} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 1\text{ V}$			20	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 1\text{ }\mu\text{A}$	45			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\text{ }\mu\text{A}$	4			V
V_{BE} Base-emitter voltage	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$	0.68			V
h_{FE} DC current gain	$I_C = 1\text{ mA}$ $V_{CE} = 6\text{ V}$		70		—
f_T Transition frequency	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 100\text{ MHz}$		800		MHz
$-C_{re}$ Reverse capacitance	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 0.5\text{ MHz}$		0.16		pF
NF Noise figure	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$			4	dB
G_{De} Power gain	$I_C = 3\text{ mA}$ $V_{CE} = 10\text{ V}$		21		dB
ΔG_{De} Power gain control	$\Delta I_C = 8\text{ mA}$ $V_{CE} = 12\text{ V}$ $R_E + R_L = 680\text{ }\Omega$ $f = 200\text{ MHz}$		30		dB
g_{ie} Input conductance	$I_C = 2\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 50\text{ MHz}$		1.5		mS
b_{ie} Input susceptance	$I_C = 2\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 50\text{ MHz}$		3.6		mS

BF 260

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min. Typ. Max.	Unit
$ y_{re} $	Reverse transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	50	μS
φ_{re}	Phase angle of reverse transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	270°	—
$ y_{fe} $	Forward transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	66	mS
φ_{fe}	Phase angle of forward transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	340°	—
g_{oe}	Output conductance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	15	μS
b_{oe}	Output susceptance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	0.34	mS
g_{ie}	Input conductance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	3.2	mS
b_{ie}	Input susceptance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	6.5	mS
$ y_{re} $	Reverse transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	94	μS
φ_{re}	Phase angle of reverse transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	270°	—
$ y_{fe} $	Forward transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	65	mS
φ_{fe}	Phase angle of forward transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	335°	—
g_{oe}	Output conductance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	30	μS
b_{oe}	Output susceptance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	0.9	mS

BF 260

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min. Typ. Max.	Unit
g_{ie}	Input conductance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	8	mS
b_{ie}	Input susceptance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	10	mS
$ y_{re} $	Reverse transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	170	μS
φ_{re}	Phase angle of reverse transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	270°	—
$ y_{fe} $	Forward transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	62	mS
φ_{fe}	Phase angle of forward transadmittance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	318°	—
g_{oe}	Output conductance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	130	μS
b_{oe}	Output susceptance	$I_C = 2 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	1.7	mS
g_{ie}	Input conductance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	2.4	mS
b_{ie}	Input susceptance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	5	mS
$ y_{re} $	Reverse transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$	50	μS
φ_{re}	Phase angle of reverse transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	270°	—
$ y_{fe} $	Forward transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	92	mS
φ_{fe}	Phase angle of forward transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$	343°	—

BF 260

ELECTRICAL CHARACTERISTICS (continued)

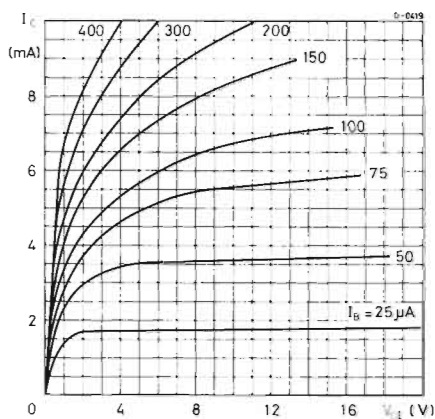
Parameter	Test conditions	Min.	Typ.	Max.	Unit
g_{oe} Output conductance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$		20		μS
b_{oe} Output susceptance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		0.34		mS
g_{ie} Input conductance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		4.5		mS
b_{ie} Input susceptance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$		7.2		mS
$ y_{re} $ Reverse transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 50 \text{ MHz}$		94		μS
φ_{re} Phase angle of reverse transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		270°		—
$ y_{fe} $ Forward transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		87		mS
φ_{fe} Phase angle of forward transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		333°		—
g_{oe} Output conductance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		50		μS
b_{oe} Output susceptance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		0.9		mS
g_{ie} Input conductance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$		9		mS
b_{ie} Input susceptance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$		11.5		mS
$ y_{re} $ Reverse transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$		170		μS
φ_{re} Phase angle of reverse transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$		270°		—
$ y_{fe} $ Forward transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$		80		mS

BF 260

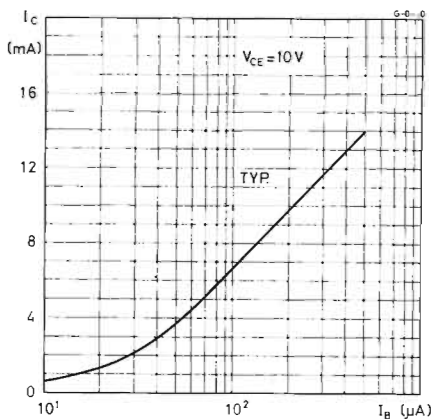
ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min. Typ. Max.	Unit
φ_{fe} Phase angle of forward transadmittance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	310°	—
g_{oe} Output conductance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	180	μS
b_{oe} Output susceptance	$I_C = 3 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 200 \text{ MHz}$	1.7	mS

Typical output characteristics

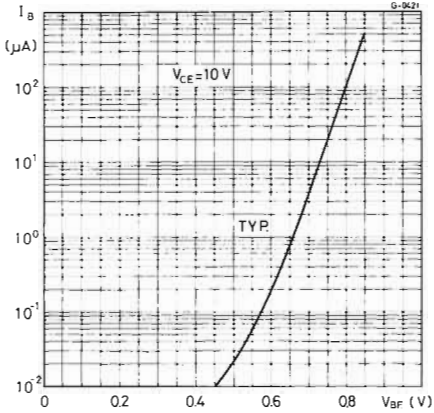


Collector characteristic

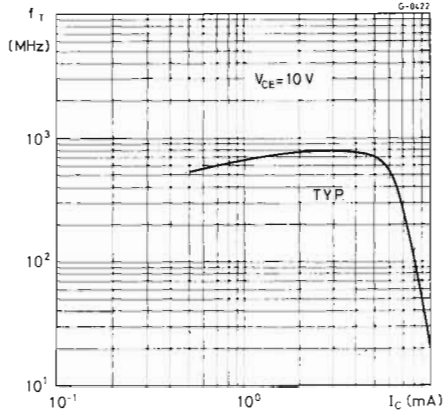


BF 260

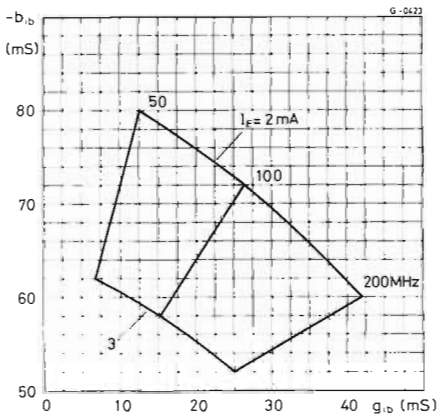
Input characteristic



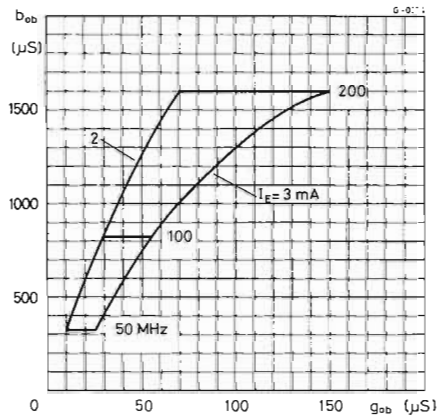
Transition frequency



Typical input admittance *



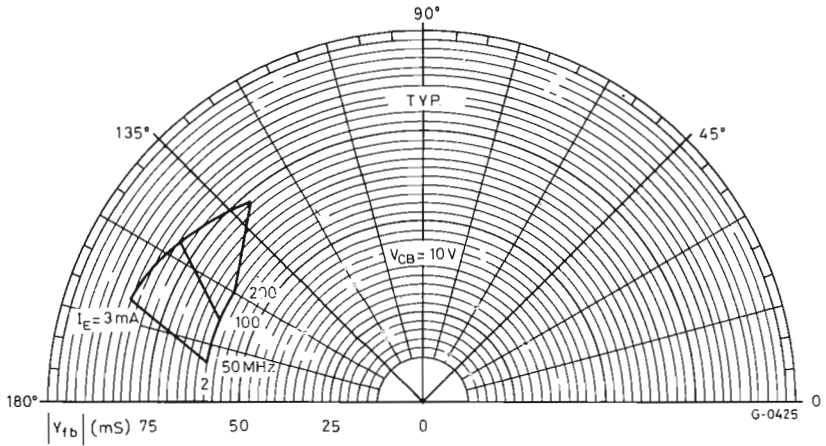
Typical output admittance *



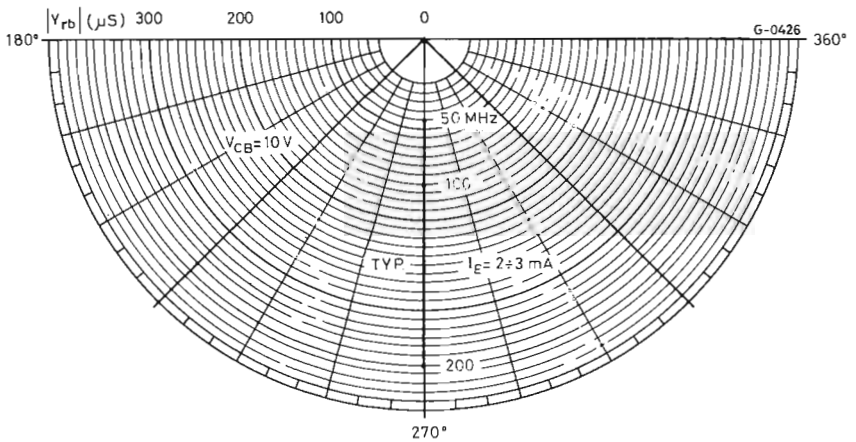
* Lead length = 3 mm.

BF 260

Forward transadmittance *



Reverse transadmittance *

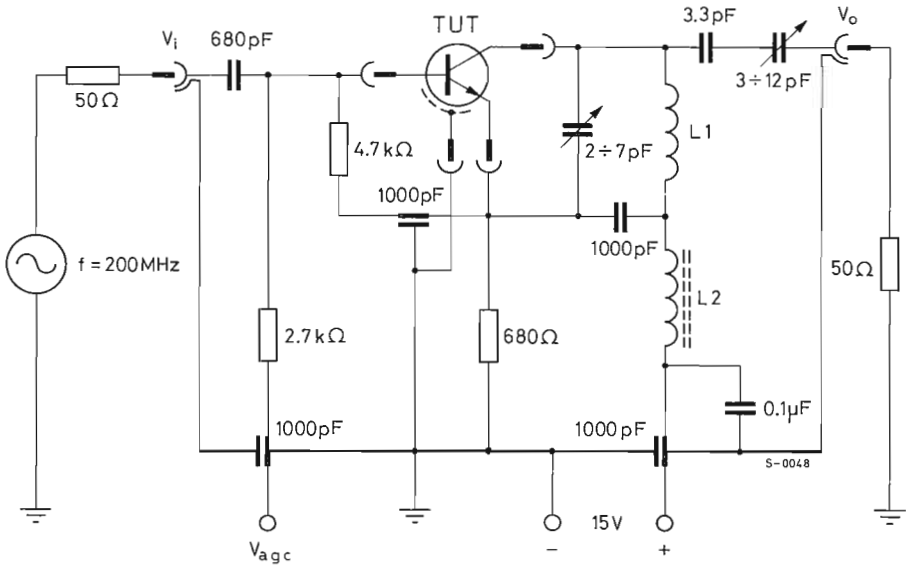


* Lead length = 3 mm.

BF 260

TEST CIRCUIT

200 MHz G_{pb} , AGC, and NF test circuit



BF 271

SILICON PLANAR NPN

VIDEO IF AMPLIFIER

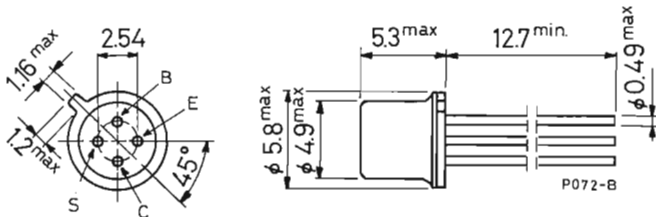
The BF 271 is a silicon planar NPN transistor in a TO-72 metal case. This device has been specifically designed for use in output stages of IF vision amplifiers. It features high power gain, low feedback capacitance and excellent linearity.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	25 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	25 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	250 mW
	at $T_{case} \leq 25^\circ\text{C}$	430 mW
T_{stg}	Storage temperature	-55 to 200 °C
T_j	Junction temperature	200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 271

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	400	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	700	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 10\text{ V}$		100		nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\ \mu\text{A}$	30			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 1\text{ mA}$	25			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu\text{A}$	4			V
V_{BE} Base-emitter voltage	$I_C = 10\text{ mA}$ $V_{CE} = 5\text{ V}$		780		mV
h_{FE}^* DC current gain	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$	30	55 75		— —
f_T Transition frequency	$I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$		900		MHz
$-C_{re}$ Feedback capacitance	$I_C = 0$ $V_{CE} = 10\text{ V}$ $f = 1\text{ MHz}$		0.22		pF
G_{pe} Power gain	$I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 36\text{ MHz}$	24	27		dB
g_{ie} Input conductance	$I_C = 10\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 36\text{ MHz}$		4.8		mS

* Pulsed: pulse duration = 300 μs ; duty factor = 1%.

BF 271

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
b_{ie} Input susceptance	$I_C = 10 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		5.2		mS
g_{fe} Forward transconductance	$I_C = 10 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		200		mS
b_{fe} Forward transusceptance	$I_C = 10 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		80		mS
g_{oe} Output conductance	$I_C = 10 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		80		μS
b_{oe} Output susceptance	$I_C = 10 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 36 \text{ MHz}$		380		μS

BF 272A

SILICON PLANAR PNP

UHF-VHF AGC AMPLIFIER

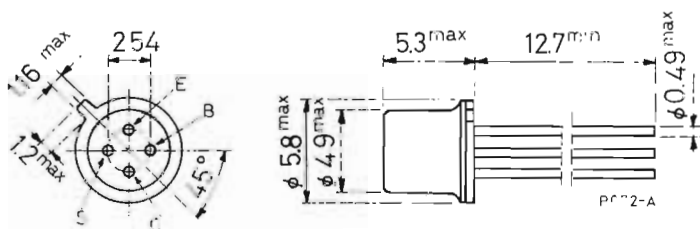
The BF 272A is a silicon planar epitaxial PNP transistor in a TO-72 metal case. This device is specifically designed for RF stages of UHF-VHF tuners. It features high gain, low feedback capacitance and very low noise figure.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-35 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-3 V
I_C	Collector current	-20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200 mW
T_{stg}	Storage temperature	-55 to 200 $^\circ\text{C}$
T_J	Junction temperature	200 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 272A

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	875 °C/W
-----------------	-------------------------------------	-----	----------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -20\text{ V}$			-100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$	-40			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -3\text{ mA}$	-35			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-3			V
V_{BE} Base-emitter voltage	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$	-0.75			V
h_{FE} DC current gain	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$	25	50		—
f_T Transition frequency	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$	700	850		MHz
$-C_{re}$ Feedback capacitance	$I_C = 0$ $V_{CE} = -10\text{ V}$ $f = 1\text{ MHz}$		0.3		pF
C_{rb} Feedback capacitance	$I_C = 0$ $V_{CB} = -10\text{ V}$ $f = 1\text{ MHz}$	0.05	0.09		pF
NF^* Noise figure	$I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_g = 50\ \Omega$ $f = 800\text{ MHz}$ $I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_g = 50\ \Omega$ $f = 200\text{ MHz}$		3.5	5.5	dB
G_{pb}^* Power gain	$I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_L = 2\text{ k}\Omega$ $f = 800\text{ MHz}$ $I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_L = 2\text{ k}\Omega$ $f = 200\text{ MHz}$	12	15		dB
			19		dB

BF 272A

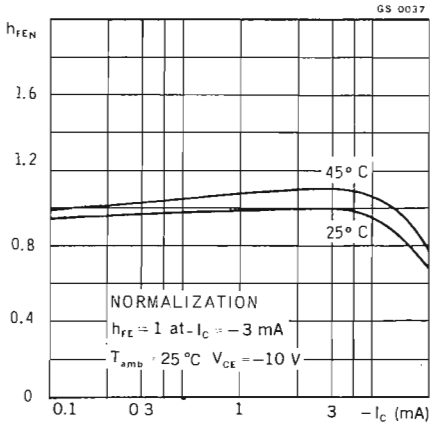
ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_C^{*} (AGC) Collector current for $\Delta G_{pb} = 30$ dB	$f = 800$ MHz $V_{CC} = 12$ V	6.6		8	mA
g_{ib} Input conductance	$I_C = -3$ mA $V_{CB} = -10$ V $f = 800$ MHz		7		mS
	$I_C = -3$ mA $V_{CB} = -10$ V $f = 200$ MHz		60		mS
b_{ib} Input susceptance	$I_C = -3$ mA $V_{CB} = -10$ V $f = 800$ MHz		-26		mS
	$I_C = -3$ mA $V_{CB} = -10$ V $f = 200$ MHz		-36		mS
g_{ob} Output conductance	$I_C = -3$ mA $V_{CB} = -10$ V $f = 800$ MHz		0.77		mS
	$I_C = -3$ mA $V_{CB} = -10$ V $f = 200$ MHz		0.1		mS
b_{ob} Output susceptance	$I_C = -3$ mA $V_{CB} = -10$ V $f = 800$ MHz		5		mS
	$I_C = -3$ mA $V_{CB} = -10$ V $f = 200$ MHz		1.3		mS
g_{fb} Forward transconductance	$I_C = -3$ mA $V_{CB} = -10$ V $f = 800$ MHz		11		mS
	$I_C = -3$ mA $V_{CB} = -10$ V $f = 200$ MHz		-51		mS
b_{fb} Forward transusceptance	$I_r = -3$ mA $V_{CB} = -10$ V $f = 800$ MHz		23		mS
	$I_C = -3$ mA $V_{CB} = -10$ V $f = 200$ MHz		45		mS
g_{rb} Reverse transconductance	$I_C = -3$ mA $V_{CB} = -10$ V $f = 800$ MHz		-0.1		mS
	$I_C = -3$ mA $V_{CB} = -10$ V $f = 200$ MHz		-0.02		mS
b_{rb} Reverse transusceptance	$I_C = -3$ mA $V_{CB} = -10$ V $f = 800$ MHz		-0.35		mS
	$I_C = -3$ mA $V_{CB} = -10$ V $f = 200$ MHz		-0.1		mS

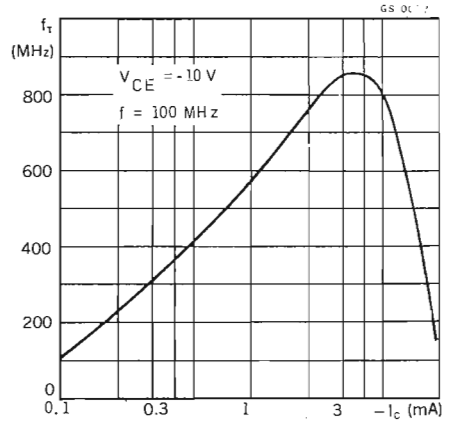
* See TEST CIRCUIT

BF 272A

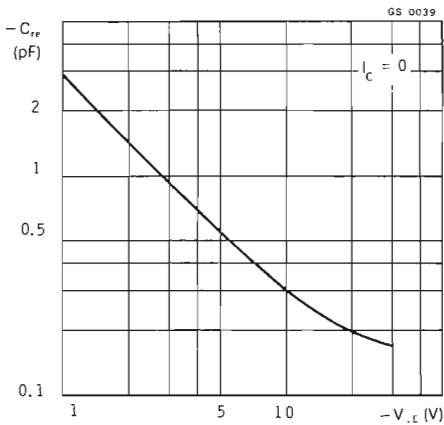
Normalized DC current gain



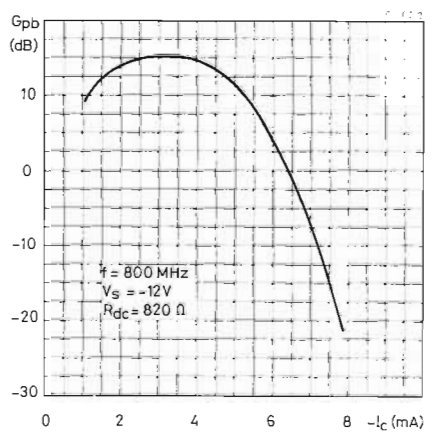
Transition frequency



Feedback capacitance

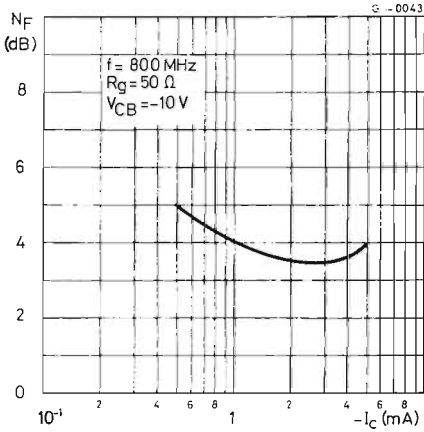


Power gain

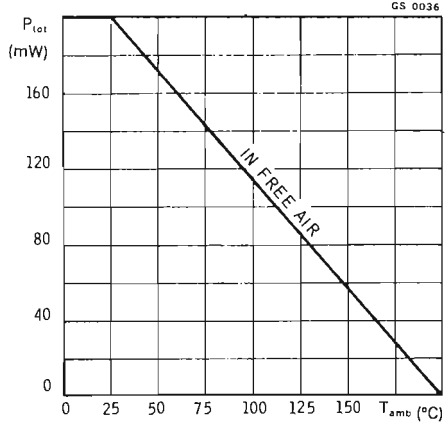


BF 272A

Noise figure

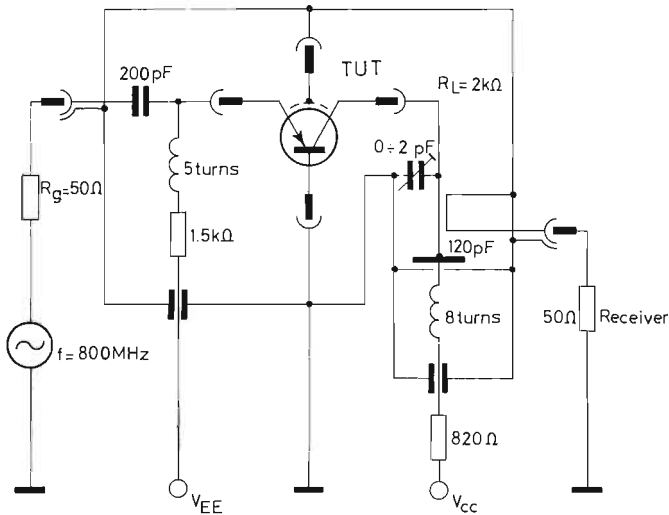


Power rating chart



TEST CIRCUIT

Power gain, AGC and noise figure ($f = 800 \text{ MHz}$)



BF 273

SILICON PLANAR NPN

AM CONVERTER AND AM-FM IF AMPLIFIER

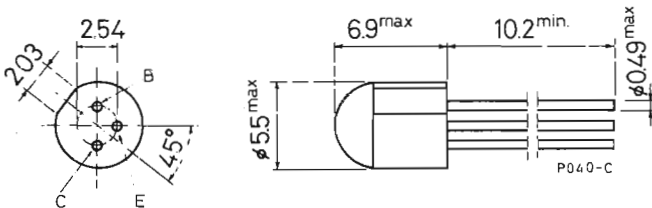
The BF 273 is a silicon planar NPN transistor in a TO-18 epoxy package, intended for use in AM converters and IF amplifiers for AM and AM/FM radios.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	25	V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	20	V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4	V
I_C	Collector current	30	mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200	mW
T_{stg}	Storage temperature	-55 to 125	$^\circ\text{C}$
T_J	Junction temperature	125	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BF 273

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	$^{\circ}C/W$
-----------------	-------------------------------------	-----	-----	---------------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25^{\circ}C$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 10\ V$ $V_{CE} = 10\ V$ $T_{amb} = 100^{\circ}C$			100 50	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\ \mu A$	25			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 1\ mA$	20			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu A$	4			V
V_{BE} Base-emitter voltage	$I_C = 1\ mA$ $V_{CE} = 10\ V$	0.70			V
h_{FE} DC current gain	$I_C = 1\ mA$ $V_{CE} = 10\ V$ for BF 273 for BF 273 Gr. C for BF 273 Gr. D	35 70 35		120 75	— — —
f_T Transition frequency	$I_C = 1\ mA$ $V_{CE} = 10\ V$	400	600		MHz
NF Noise figure	$I_C = 1\ mA$ $V_{CE} = 10\ V$ $R_g = 400\ \Omega$ $f = 100\ MHz$		2		dB
$-C_{re}$ Feedback capacitance	$I_C = 0$ $V_{CE} = 10\ V$ $f = 1\ MHz$		0.41		pF
G_{pe} Power gain	$I_C = 1\ mA$ $V_{CE} = 10\ V$ $f = 470\ kHz$ $I_C = 1\ mA$ $V_{CE} = 10\ V$ $f = 10.7\ MHz$ $I_C = 1\ mA$ $V_{CE} = 10\ V$ $f = 100\ MHz$		40 30 21		dB dB dB

BF 273

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
g_{ie}	Input conductance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$		240		μS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 10.7 \text{ MHz}$		300		μS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		900		μS
b_{ie}	Input susceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$		22		μS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 10.7 \text{ MHz}$		500		μS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		4.8		mS
b_{re}	Reverse transusceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$		-1.2		μS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 10.7 \text{ MHz}$		-27.6		μS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		-260		μS
ϕ_{re}	Reverse transadmittance phase	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$		-90°		—
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 10.7 \text{ MHz}$		-90°		—
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		-90°		—
g_{fe}	Forward transconductance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$		35		mS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 10.7 \text{ MHz}$		35		mS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		32		mS
b_{fe}	Forward transusceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 10.7 \text{ MHz}$		-1		mS
		$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		-9		mS

BF 273

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
g_{oe} Output conductance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$		7		μS
	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 10.7 \text{ MHz}$		11		μS
	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		75		μS
b_{oe} Output susceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 470 \text{ kHz}$		4.4		μS
	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 10.7 \text{ MHz}$		100		μS
	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		940		μS

BF 274

SILICON PLANAR NPN

GAIN CONTROLLED AM-FM IF AMPLIFIER

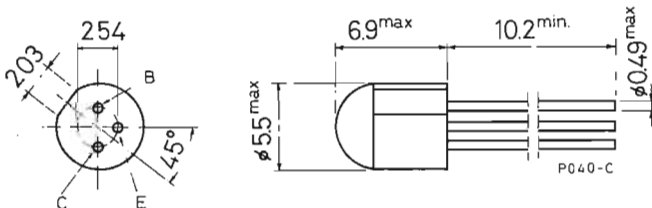
The BF 274 is a silicon planar NPN transistor in a TO-18 epoxy package, primarily intended for use in the gain controlled IF stages of AM and AM/FM radio receivers.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	25	V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	20	V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4	V
I_C	Collector current	30	mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200	mW
T_{stg}	Storage temperature	-55 to 125	$^\circ\text{C}$
T_J	Junction temperature	125	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BF 274

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W
-----------------	-------------------------------------	-----	-----	------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25^\circ\text{C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 10\text{ V}$ $V_{CE} = 10\text{ V}$ $T_{amb} = 100^\circ\text{C}$			100 50	nA μA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\ \mu\text{A}$	25			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = 1\text{ mA}$	20			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu\text{A}$	4			V
V_{BE} Base-emitter voltage	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$	0.70			V
h_{FE} DC current gain	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ for BF 274 for BF 274 Gr. B for BF 274 Gr. C	70 100 70		250 120	— — —
f_T Transition frequency	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$	400	700		MHz
$-C_{re}$ Feedback capacitance	$I_C = 0$ $V_{CE} = 10\text{ V}$ $f = 1\text{ MHz}$		0.41		pF
G_{pe} Power gain	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 470\text{ kHz}$ $I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 10.7\text{ MHz}$		40 30		dB dB
ΔG_{pe} Power gain control	$I_C = 100\ \mu\text{A}$ $V_{CE} = 10\text{ V}$ $f = 470\text{ kHz}$		20		dB

BF 287

SILICON PLANAR NPN

AM MIXER-OSCILLATOR AND AM-FM AMPLIFIER

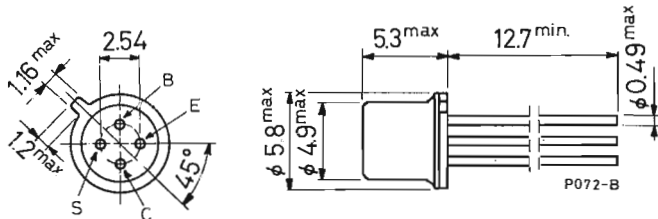
The BF 287 is a silicon planar NPN transistor in a TO-72 metal case. It is primarily intended for use in the AM mixer-oscillator stage and as IF amplifier of AM-FM radios.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	250 mW
	at $T_{case} \leq 45^\circ\text{C}$	220 mW
T_{stg}	Storage temperature	-55 to 200 °C
T_j	Junction temperature	200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 287

THERMAL DATA

$R_{th j-amb}$	Thermal resistance junction-ambient	max	700	°C/W
----------------	-------------------------------------	-----	-----	------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 10\text{ V}$			100	nA
$V_{(BR) CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\text{ }\mu\text{A}$	40			V
$V_{CEO (sus)}$ Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 5\text{ mA}$	40			V
$V_{(BR) EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\text{ }\mu\text{A}$	4			V
V_{BE} Base-emitter voltage	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $I_C = 2\text{ mA}$ $V_{CE} = 10\text{ V}$		710 740		mV mV
h_{FE} DC current gain	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $I_C = 2\text{ mA}$ $V_{CE} = 10\text{ V}$	30 40	50 60		— —
f_T Transition frequency	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 100\text{ MHz}$ $I_C = 2\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 100\text{ MHz}$		600 700		MHz MHz
G_{pe} Power gain	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 470\text{ kHz}$ $f = 10.7\text{ MHz}$ $I_C = 2\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 5.5\text{ MHz}$	42 18 25	45 22 29		dB dB dB
g_{ie} Input conductance	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 470\text{ kHz}$ $f = 10.7\text{ MHz}$		0.17 0.25		mS mS
b_{ie} Input susceptance	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 470\text{ kHz}$ $f = 10.7\text{ MHz}$		24 0.52		μS mS

BF 287

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
g_{fe} Forward transconductance	$I_C = 1 \text{ mA}$ $V_{CE} = 7 \text{ V}$ $f = 470 \text{ kHz}$ $f = 10.7 \text{ MHz}$		35	35	mS mS
$-b_{fe}$ Forward transusceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 7 \text{ V}$ $f = 470 \text{ kHz}$ $f = 10.7 \text{ MHz}$		40	0.96	μS mS
g_{oe} Output conductance	$I_C = 1 \text{ mA}$ $V_{CE} = 7 \text{ V}$ $f = 470 \text{ kHz}$ $f = 10.7 \text{ MHz}$		6	11	μS μS
b_{oe} Output susceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 7 \text{ V}$ $f = 470 \text{ kHz}$ $f = 10.7 \text{ MHz}$		4.5	100	μS μS

BF 288

SILICON PLANAR NPN

GAIN CONTROLLED AM-FM IF AMPLIFIER

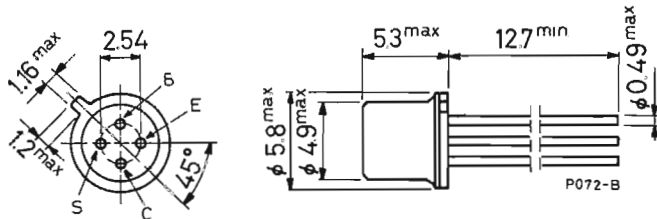
The BF 288 is a silicon planar NPN transistor in a TO-72 metal case. It is primarily intended for use in the gain controlled IF stages of AM and AM/FM radio receivers.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	40 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	250 mW
	at $T_{amb} \leq 45^\circ\text{C}$	220 mW
T_{stg}	Storage temperature	-55 to 200 °C
T_j	Junction temperature	200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 288

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	700 °C/W
-----------------	-------------------------------------	-----	----------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CES} Collector cutoff current ($V_{BE} = 0$)	$V_{CE} = 7\text{ V}$			100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 10\ \mu\text{A}$	40			V
$V_{CEO\ (sus)}$ Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 5\text{ mA}$	40			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 100\ \mu\text{A}$	4			V
V_{BE} Base-emitter voltage	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$		740		mV
h_{FE} DC current gain	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$	65	90		—
f_T Transition frequency	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$		500		MHz
$-C_{re}$ Feedback capacitance	$V_{CE} = 7\text{ V}$ $f = 1\text{ MHz}$		0.24		pF
G_{pe} Power gain	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 470\text{ kHz}$ $f = 10.7\text{ MHz}$	42 18	45 22		dB dB
g_{ie} Input conductance	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 470\text{ kHz}$ $f = 10.7\text{ MHz}$		0.17 0.25		mS mS
b_{ie} Input susceptance	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 470\text{ kHz}$ $f = 10.7\text{ MHz}$		24 0.52		μS mS
g_{fe} Forward transconductance	$I_C = 1\text{ mA}$ $V_{CE} = 7\text{ V}$ $f = 470\text{ kHz}$ $f = 10.7\text{ MHz}$		35 35		mS mS

BF 288

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min. Typ. Max.	Unit
$-b_{fe}$	Forward transusceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 7 \text{ V}$ $f = 470 \text{ kHz}$ $f = 10.7 \text{ MHz}$	40 0.95	μS mS
g_{oe}	Output conductance	$I_C = 1 \text{ mA}$ $V_{CE} = 7 \text{ V}$ $f = 470 \text{ kHz}$ $f = 10.7 \text{ MHz}$	6 11	μS μS
b_{oe}	Output susceptance	$I_C = 1 \text{ mA}$ $V_{CE} = 7 \text{ V}$ $f = 470 \text{ kHz}$ $f = 10.7 \text{ MHz}$	4.5 100	μS μS

BF 316A

SILICON PLANAR PNP

UHF MIXER OSCILLATOR

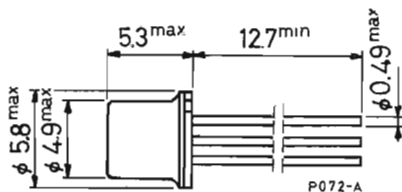
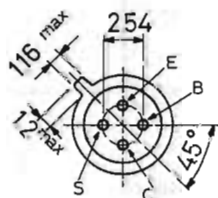
The BF 316 A is a silicon planar epitaxial PNP transistor in a TO-72 metal case. It is specifically designed for use as oscillator-mixer in UHF tuners.

ABSOLUTE MAXIMUM RATINGS

V_{CB0}	Collector-base voltage ($I_E = 0$)	-40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-35 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-3 V
I_C	Collector current	-20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200 mW
T_{stg}	Storage temperature	-55 to 200 °C
T_j	Junction temperature	200 °C

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 316A

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	875 °C/W
-----------------	-------------------------------------	-----	----------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	
I_{CBO}	Collector cutoff current ($I_E = 0$)	$V_{CB} = -20\text{ V}$			-100	nA
$V_{(BR)\ CBO}$	Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$			-40	V
$V_{(BR)\ CEO}$	Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -3\text{ mA}$			-35	V
$V_{(BR)\ EBO}$	Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$			-3	V
V_{BE}	Base-emitter voltage	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$			-0.75	V
h_{FE}	DC current gain	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$			30 50	—
f_T	Transition frequency	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$			600	MHz
$-C_{re}$	Feedback capacitance	$I_C = 0$ $V_{CE} = -10\text{ V}$ $f = 1\text{ MHz}$			0.25	pF
NF	Noise figure	$I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_g = 50\ \Omega$ $f = 800\text{ MHz}$ $I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_g = 50\ \Omega$ $f = 500\text{ MHz}$			5 3.5	dB dB
G_{pb}	Power gain	$I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_L = 2\text{ k}\Omega$ $f = 800\text{ MHz}$ $I_C = -3\text{ mA}$ $V_{CB} = -10\text{ V}$ $R_L = 2\text{ k}\Omega$ $f = 500\text{ MHz}$			12 17	dB dB

BF 316A

ELECTRICAL CHARACTERISTICS (continued)

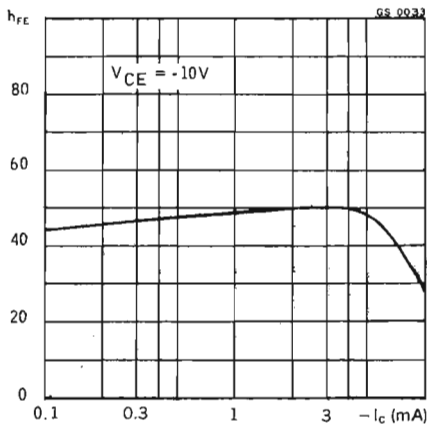
Parameter	Test conditions	Min.	Typ.	Max.	Unit
g_{ib} Input conductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		4.6		mS
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		17		mS
b_{ib} Input susceptance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		-23		mS
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		-37		mS
g_{ob} Output conductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		0.6		mS
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		0.32		mS
b_{ob} Output susceptance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		5		mS
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		3.2		mS
g_{fb} Forward transconductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		16		mS
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		10		mS
b_{fb} Forward transusceptance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		13		mS
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		39		mS
g_{rb} Reverse transconductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		-0.1		mS
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		-0.04		mS
b_{rb} Reverse transusceptance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		-0.32		mS
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		-0.26		mS

BF 316A

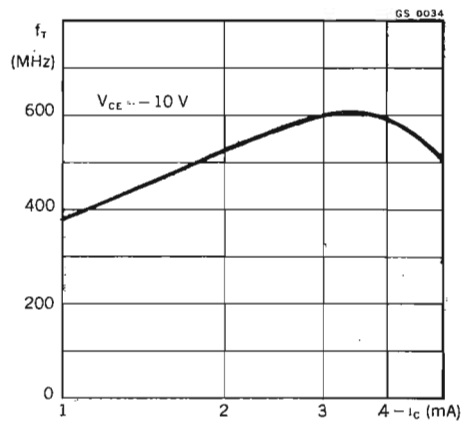
ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
$\varphi_{fb}-\varphi_{ib}$ Phase difference	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$		118°		—
	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 500 \text{ MHz}$		141°		—
R_{ob} Output resistance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 43 \text{ MHz}$		0.02		mS
g_{ob} Output conductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 43 \text{ MHz}$		1		pF

DC current gain

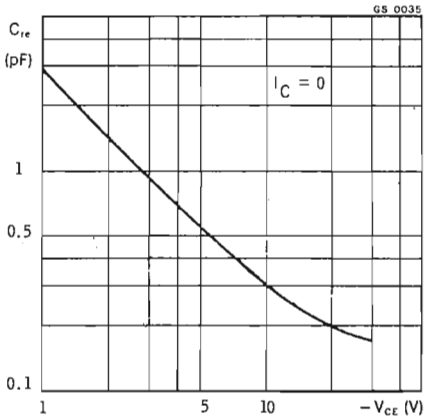


Transition frequency

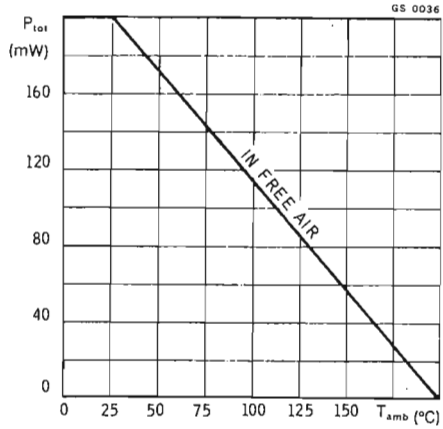


BF 316A

Feedback capacitance

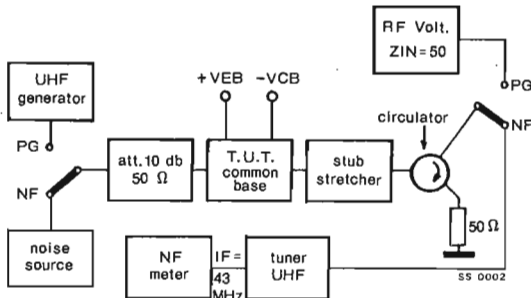


Power rating chart



TEST CIRCUIT

Power gain, noise figure ($f = 800$ MHz)



BF 454

SILICON PLANAR NPN

AM/FM IF AMPLIFIER

The BF 454 is a silicon planar NPN transistor in a TO-18 epoxy package, with low reverse capacitance, very low noise, high output impedance.

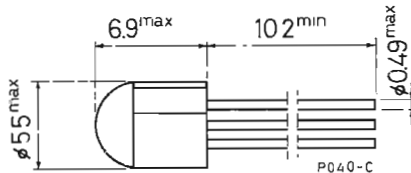
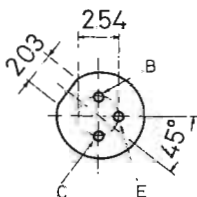
The BF 454 is especially suited for FM tuner stages, AM mixer/oscillators and for AM/FM IF amplifiers.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	35 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	25 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$ at $T_{case} \leq 25^\circ\text{C}$	200 mW 500 mW
T_{stg}	Storage temperature	-55 to 125 °C
T_J	Junction temperature	125 °C

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BF 454

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W
-----------------	-------------------------------------	-----	-----	------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{°C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 10\text{ V}$			200	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$	35			V
$V_{CEO(sus)}$ *Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 1\text{ mA}$	25			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu\text{A}$	4			V
V_{BE} Base-emitter voltage	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$	0.71			V
h_{FE} DC current gain Gr. B	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$	110 65		200 220	— —
f_T Transition frequency	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 100\text{ MHz}$		400		MHz
$-C_{re}$ Reverse capacitance	$I_C = 0$ $V_{CE} = 10\text{ V}$ $f = 1\text{ MHz}$		0.5	0.8	pF
NF Noise figure	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $R_g = 100\ \Omega$ $f = 100\text{ MHz}$		3		dB

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

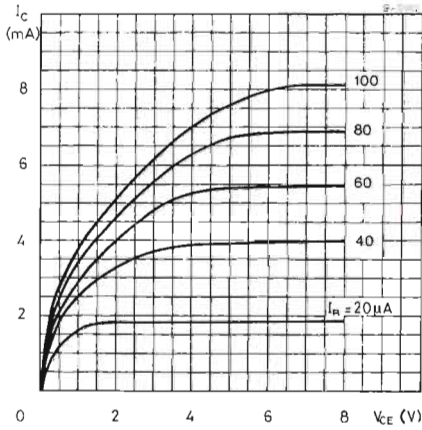
BF 454

ELECTRICAL CHARACTERISTICS (continued)

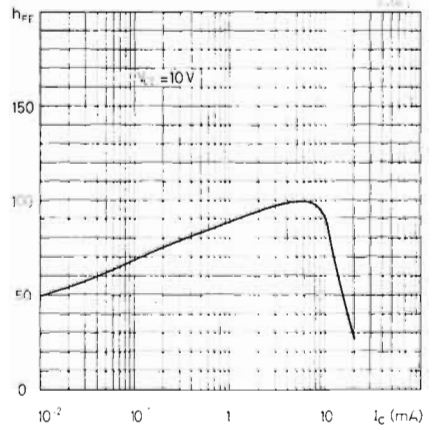
Parameter	Test conditions	Min.	Typ.	Max.	Unit	
g_{ib} Input conductance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		36		mS	
$-b_{ib}$ Input susceptance			3.5		mS	
$ Y_{fb} $ Forward transadmittance				34		mS
ϕ_{fb} Phase angle of the forward transadmittance				160°		—
g_{ob} Output conductance				22		μS
b_{ob} Output susceptance				0.86		mS

BF 454

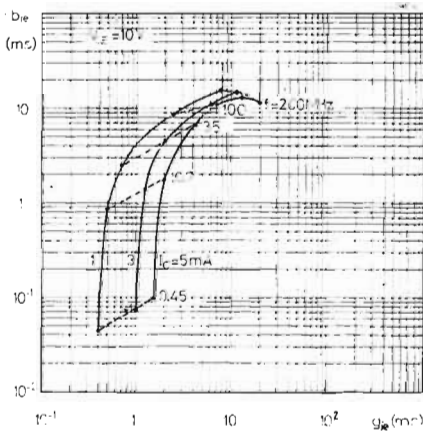
Typical output characteristics



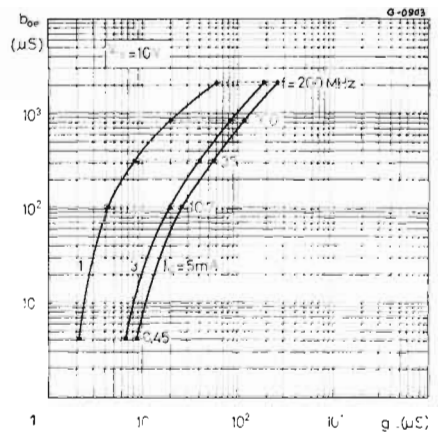
Typical DC current gain



Typical input admittance *



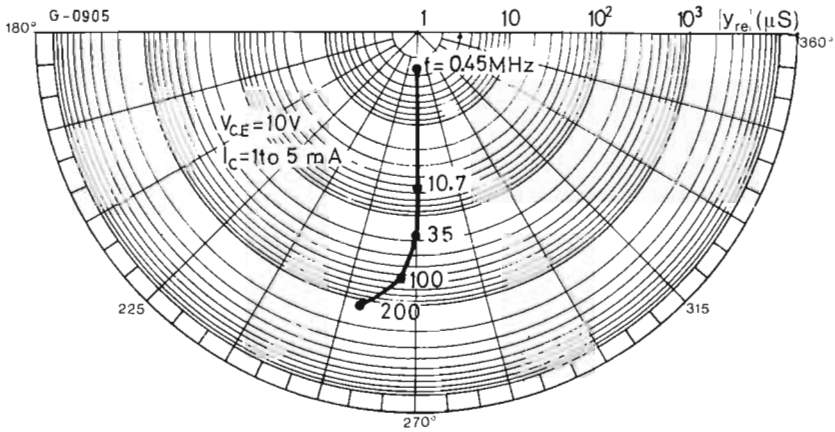
Typical output admittance *



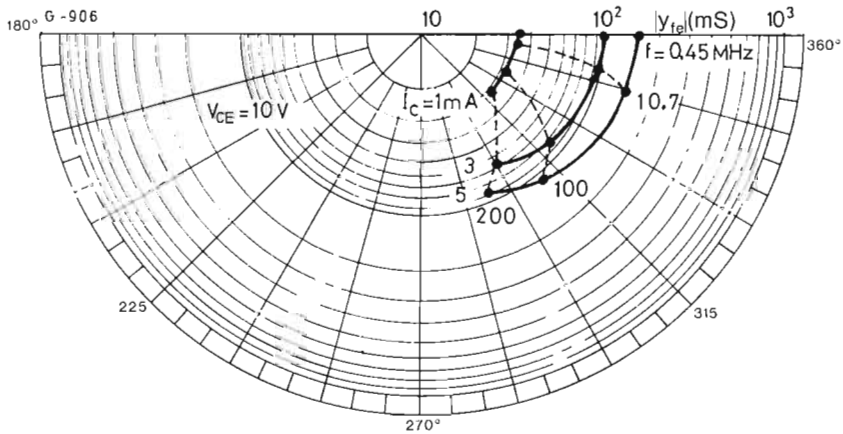
* Lead length = 3 mm.

BF 454

Typical reverse transadmittance *



Typical forward transadmittance *



* Lead length = 3 mm.

BF 455

SILICON PLANAR NPN

PREAMPLIFIER AND AM/FM IF AMPLIFIER

The BF 455 is a silicon planar NPN transistor in TO-18 epoxy package, with low reverse capacitance, very low noise, high output impedance.

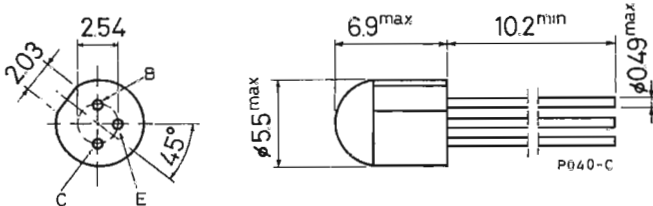
The BF 455 is especially suited for FM tuners, IF amplifiers in AM/FM receivers, AM input stages of car-radios.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	35 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	25 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	4 V
I_C	Collector current	20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200 mW
	at $T_{case} \leq 25^\circ\text{C}$	500 mW
T_{stg}	Storage temperature	-55 to 125 °C
T_j	Junction temperature	125 °C

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BF 455

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W
-----------------	-------------------------------------	-----	-----	------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 10\text{ V}$			200	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = 100\ \mu\text{A}$	35			V
$V_{CEO(sus)}$ * Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = 1\text{ mA}$	25			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = 10\ \mu\text{A}$	4			V
V_{BE} Base-emitter voltage	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$		0.71		V
h_{FE} DC current gain	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$	68		120	—
	Gr. C $I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$	38		75	—
	Gr. D $I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$	35		125	—
f_T Transition frequency	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $f = 100\text{ MHz}$		400		MHz
$-C_{re}$ Reverse capacitance	$I_C = 0$ $V_{CE} = 10\text{ V}$ $f = 1\text{ MHz}$		0.5	0.8	pF
NF Noise figure	$I_C = 1\text{ mA}$ $V_{CE} = 10\text{ V}$ $R_g = 100\ \Omega$ $f = 100\text{ MHz}$		3		dB

* Pulsed: pulse duration = 300 μs , duty factor = 1%.

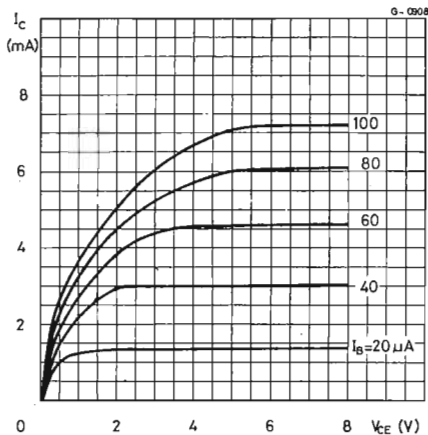
BF 455

ELECTRICAL CHARACTERISTICS (continued)

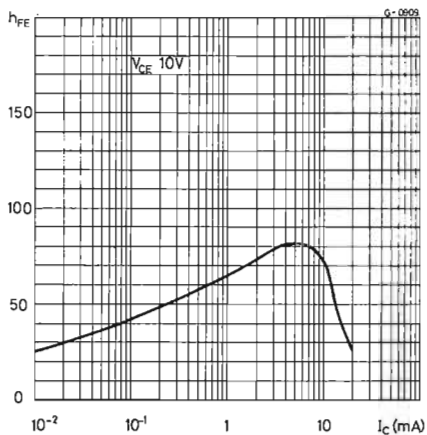
Parameter	Test conditions	Min.	Typ.	Max.	Unit	
g_{ib} Input conductance	$I_C = 1 \text{ mA}$ $V_{CE} = 10 \text{ V}$ $f = 100 \text{ MHz}$		38		mS	
$-b_{ib}$ Input susceptance			2		mS	
$ Y_{fb} $ Forward transadmittance				34		mS
ϕ_{fb} Phase angle of the forward transadmittance				150°		—
g_{ob} Output conductance				13		μS
b_{ob} Output susceptance				0.8		mS

BF 455

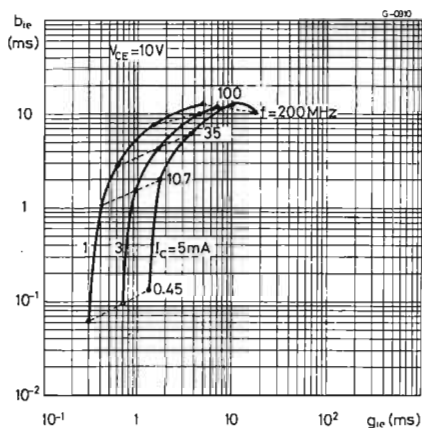
Typical output characteristics



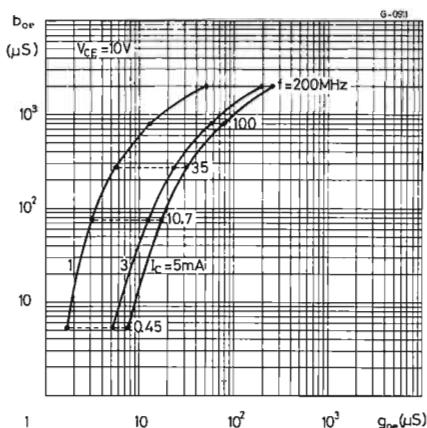
Typical DC current gain



Typical input admittance *



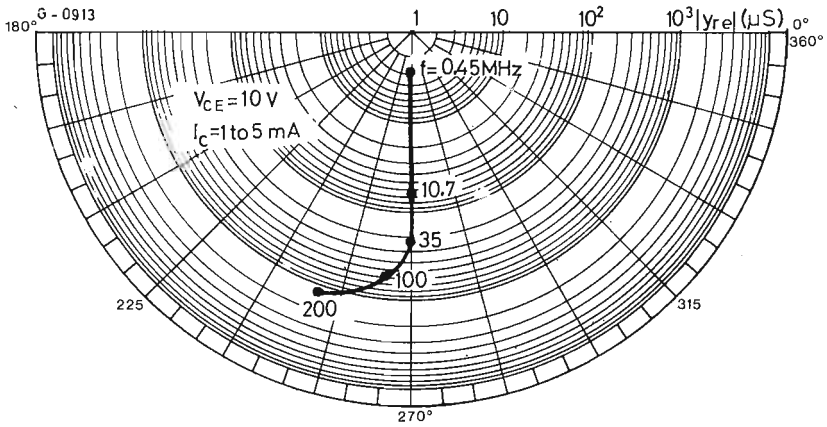
Typical output admittance *



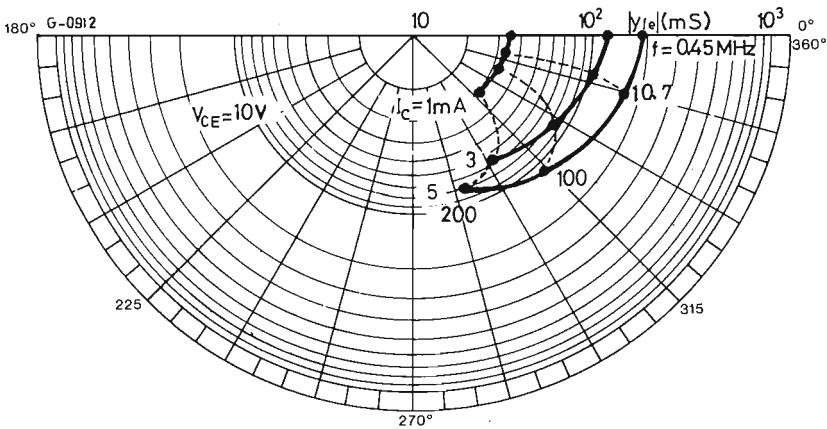
* Lead length = 3 mm.

BF 455

Typical reverse transadmittance *



Typical forward transadmittance *



* Lead length = 3 mm.

BF 479

SILICON PLANAR PNP

PRELIMINARY DATA

LOW-NOISE ULTRA LINEAR UHF-VHF AMPLIFIER

The BF 479 is a PNP silicon planar epitaxial transistor in a T-plastic package mainly intended for high current UHF-VHF stages of TV tuners.

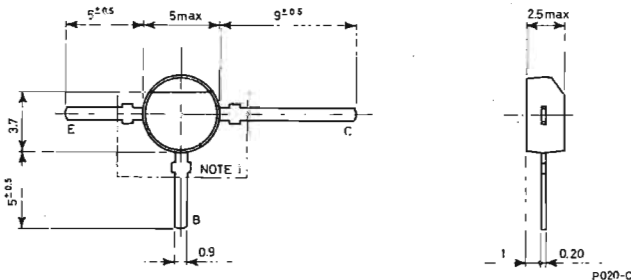
In this application, combined with a PIN diode attenuator circuit, it presents very low noise and very good cross modulation performances up to 900 MHz.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-25 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-3 V
I_C	Collector current	-50 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 45^\circ\text{C}$	170 mW
T_{stg}	Storage temperature	-55 to 150 $^\circ\text{C}$
T_j	Junction temperature	150 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



(1) Within this region the cross section of the leads is uncontrolled

BF 479

THERMAL DATA

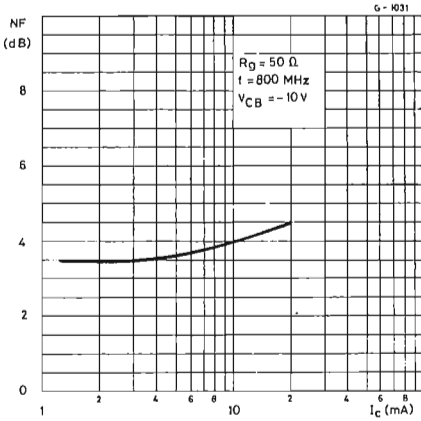
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	600	$^{\circ}\text{C}/\text{W}$
-----------------	-------------------------------------	-----	-----	-----------------------------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25^{\circ}\text{C}$ unless otherwise specified)

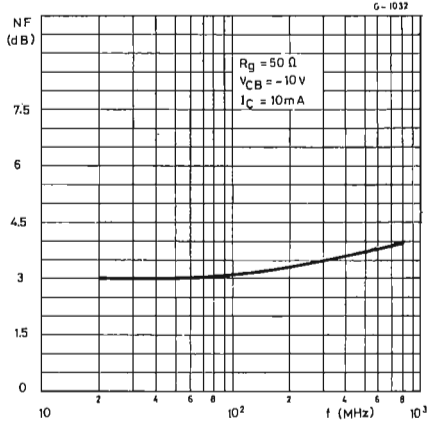
Parameter		Test conditions	Min.	Typ.	Max.	Unit
I_{CBO}	Collector cutoff current ($I_E = 0$)	$V_{CB} = -20\text{ V}$			-100	nA
$V_{(BR)\ CBO}$	Collector-base breakdown voltage ($I_E = 0$)	$I_C = -100\ \mu\text{A}$	-30			V
$V_{(BR)\ CEO}$	Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -5\ \text{mA}$	-25			V
$V_{(BR)\ EBO}$	Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-3			V
h_{FE}	DC current gain	$I_C = -10\ \text{mA}$ $V_{CE} = -10\ \text{V}$	20			—
f_T	Transition frequency	$I_C = -10\ \text{mA}$ $V_{CE} = -10\ \text{V}$ $f = 100\ \text{MHz}$		1.4		GHz
C_{CBO}	Collector-base capacitance	$I_E = 0$ $V_{CB} = -10\ \text{V}$ $f = 1\ \text{MHz}$		0.7		pF
NF	Noise figure	$V_{CB} = -10\ \text{V}$ $R_g = 50\ \Omega$ $I_C = -3\ \text{mA}$ $f = 200\ \text{MHz}$ $I_C = -10\ \text{mA}$ $f = 200\ \text{MHz}$ $I_C = -3\ \text{mA}$ $f = 800\ \text{MHz}$ $I_C = -10\ \text{mA}$ $f = 800\ \text{MHz}$		2.5 3.3 3.5 4	5.5 6	dB dB dB dB
G_{db}	Power gain	$I_C = -10\ \text{mA}$ $V_{CB} = -10\ \text{V}$ $R_L = 2\ \text{k}\Omega$ $f = 800\ \text{MHz}$	15	18		dB

BF 479

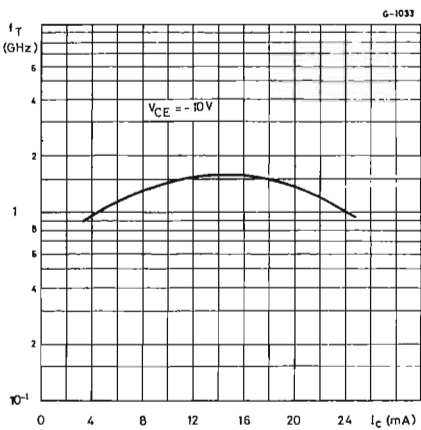
Typical noise figure



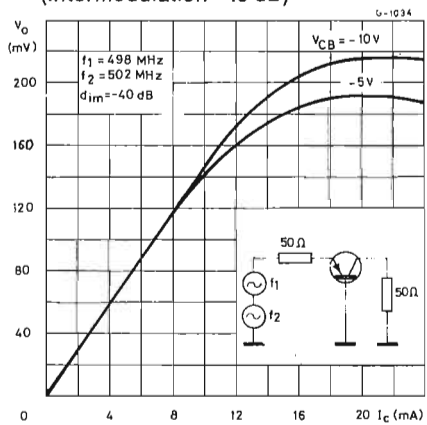
Typical noise figure



Typical transition frequency



Typical output voltage
(intermodulation -40 dB)



SILICON PLANAR PNP

BF 500 BF 500A

PRELIMINARY DATA

VHF PREAMPLIFIERS AND MIXER/OSCILLATORS

The BF 500 and BF 500 A are silicon planar epitaxial PNP transistors in TO-18 epoxy package, designed for use as preamplifiers and mixer/oscillators up to 200 MHz in common base connection.

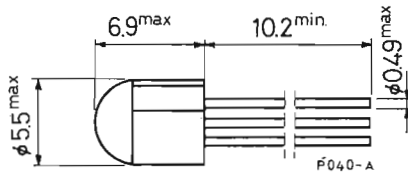
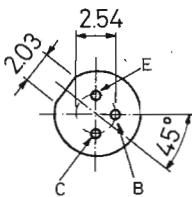
The BF 500 A has a very low guaranteed input noise.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-30 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-30 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-3 V
I_C	Collector current	-20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200 mW
T_{stg}	Storage temperature	-55 to 125 $^\circ\text{C}$
T_j	Junction temperature	125 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TO-18 epoxy

BF 500 BF 500A

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	500	°C/W
-----------------	-------------------------------------	-----	-----	------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -10\text{ V}$			-100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$	-30			V
$V_{CEO(sus)}$ *Collector-emitter sustaining voltage ($I_B = 0$)	$I_C = -1\text{ mA}$	-30			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-4			V
h_{FE} DC current gain	$I_C = -1\text{ mA}$ $V_{CE} = -10\text{ V}$ $I_C = -4\text{ mA}$ $V_{CE} = -10\text{ V}$	30	50	50	— —
f_T Transition frequency	$I_C = -1\text{ mA}$ $V_{CE} = -10\text{ V}$ $f = 100\text{ MHz}$		400		MHz
NF Noise figure (for BF 500 A only)	$I_C = -1\text{ mA}$ $V_{CB} = -6\text{ V}$ $R_g = 100\ \Omega$ $f = 100\text{ MHz}$		2	4	dB
C_{rb} Reverse capacitance	$I_C = 0$ $V_{CB} = -10\text{ V}$ $f = 1\text{ MHz}$		0.3		pF

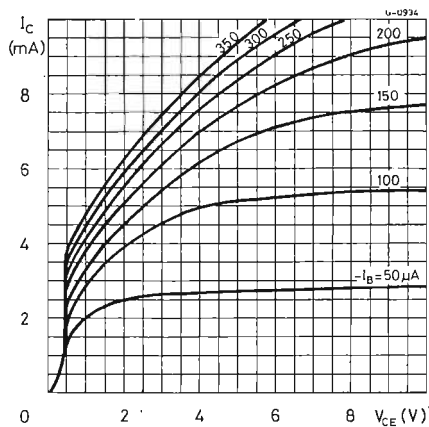
* Pulsed: pulse duration = 300 μs , duty factor = 1%.

BF 500 BF 500A

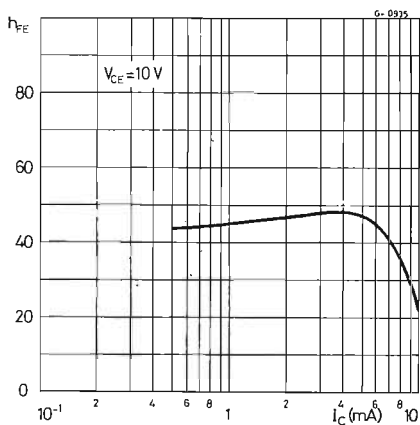
ELECTRICAL CHARACTERISTICS (T_{amb} = 25 °C unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
g_{ib} Input conductance	$I_C = -1 \text{ mA}$ $V_{CB} = -6 \text{ V}$ $f = 100 \text{ MHz}$		36		mS
b_{ib} Input susceptance			4		mS
$ Y_{fb} $ Forward transadmittance			36		mS
ϕ_{fb} Phase angle of the forward transadmittance			167°		—
g_{ob} Output conductance			10		μS
b_{ob} Output susceptance			0.7		mS

Typical output characteristics

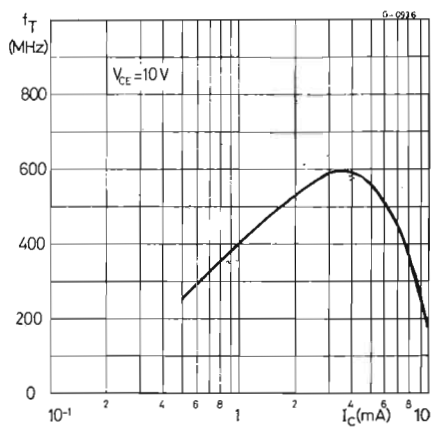


Typical DC current gain



BF 500 BF 500A

Typical transition frequency



BF 516

SILICON PLANAR PNP

UHF-VHF AMPLIFIER

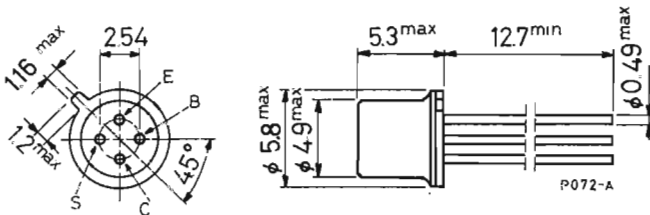
The BF 516 is a silicon planar epitaxial PNP transistor in a TO-72 metal case, intended as general purpose amplifier up to 1 GHz.

ABSOLUTE MAXIMUM RATINGS

V_{CBO}	Collector-base voltage ($I_E = 0$)	-40 V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	-35 V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	-3 V
I_C	Collector current	-20 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	200 mW
T_{stg}	Storage temperature	-55 to 200 $^\circ\text{C}$
T_j	Junction temperature	200 $^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



(sim. to TO-72)

BF 516

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	875 °C/W
-----------------	-------------------------------------	-----	----------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = -20\text{ V}$			-100	nA
$V_{(BR)\ CBO}$ Collector-base breakdown voltage ($I_E = 0$)	$I_C = -10\ \mu\text{A}$	-40			V
$V_{(BR)\ CEO}$ Collector-emitter breakdown voltage ($I_B = 0$)	$I_C = -3\text{ mA}$	-35			V
$V_{(BR)\ EBO}$ Emitter-base breakdown voltage ($I_C = 0$)	$I_E = -10\ \mu\text{A}$	-3			V
V_{BE} Base-emitter voltage	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$	-0.75			V
h_{FE} DC current gain	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$	25	50		—
f_T Transition frequency	$I_C = -3\text{ mA}$ $V_{CE} = -10\text{ V}$	700	850		MHz
$-C_{re}$ Feedback capacitance	$I_C = 0$ $f = 1\text{ MHz}$ $V_{CB} = -10\text{ V}$		0.3		pF
C_{rb} Feedback capacitance	$I_C = 0$ $f = 1\text{ MHz}$ $V_{CE} = -10\text{ V}$		0.05	0.09	pF
NF Noise figure	$I_C = -3\text{ mA}$ $V_{CB} = -12\text{ V}$ $R_g = 50\ \Omega$ $f = 800\text{ MHz}$ $I_C = -3\text{ mA}$ $V_{CB} = -12\text{ V}$ $R_g = 50\ \Omega$ $f = 200\text{ MHz}$		3.5	6	dB
G_{pb} Power gain	$I_C = -3\text{ mA}$ $V_{CB} = -12\text{ V}$ $R_L = 2\text{ k}\Omega$ $f = 800\text{ MHz}$ $I_C = -3\text{ mA}$ $V_{CB} = -12\text{ V}$ $R_L = 2\text{ k}\Omega$ $f = 200\text{ MHz}$		11	14	dB
				19	dB

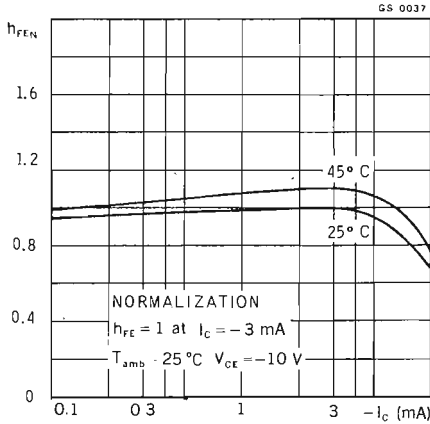
BF 516

ELECTRICAL CHARACTERISTICS (continued)

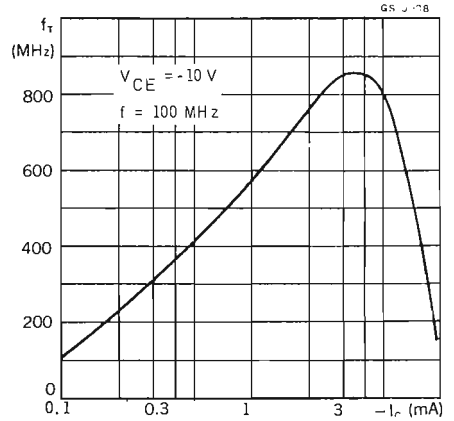
Parameter		Test conditions	Min. Typ. Max.	Unit
g_{ib}	Input conductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$	7	mS
		$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 200 \text{ MHz}$	60	mS
b_{ib}	Input susceptance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$	-26	mS
		$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 200 \text{ MHz}$	-36	mS
g_{ob}	Output conductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$	0.77	mS
		$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 200 \text{ MHz}$	0.10	mS
b_{ob}	Output susceptance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$	5	mS
		$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 200 \text{ MHz}$	1.3	mS
g_{fb}	Forward transconductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$	11	mS
		$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 200 \text{ MHz}$	-51	mS
b_{fb}	Forward transusceptance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$	23	mS
		$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 200 \text{ MHz}$	45	mS
g_{rb}	Reverse transconductance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$	-0.1	mS
		$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 200 \text{ MHz}$	-0.02	mS
b_{rb}	Reverse transusceptance	$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 800 \text{ MHz}$	-0.35	mS
		$I_C = -3 \text{ mA}$ $V_{CB} = -10 \text{ V}$ $f = 200 \text{ MHz}$	-0.1	mS

BF 516

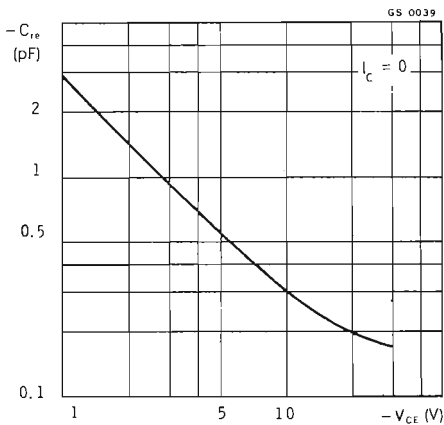
Normalized DC current gain



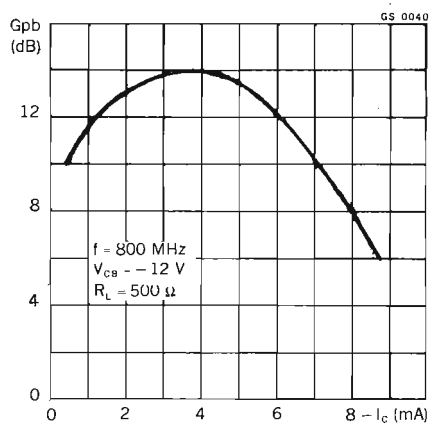
Transition frequency



Feedback capacitance

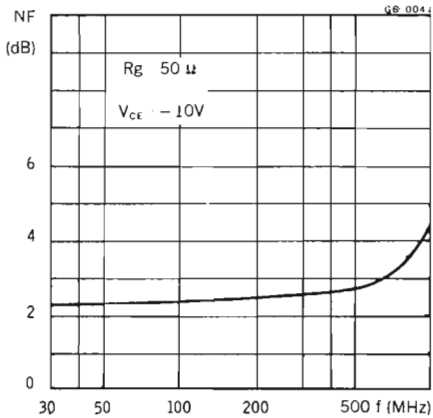


Power gain

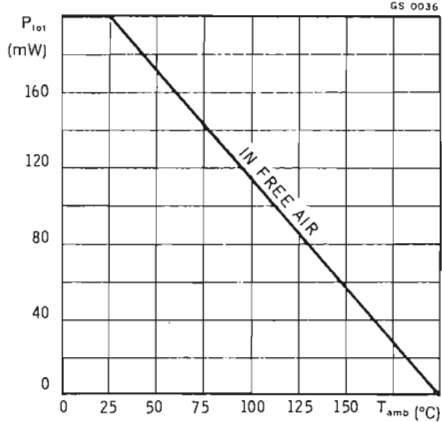


BF 516

Noise figure

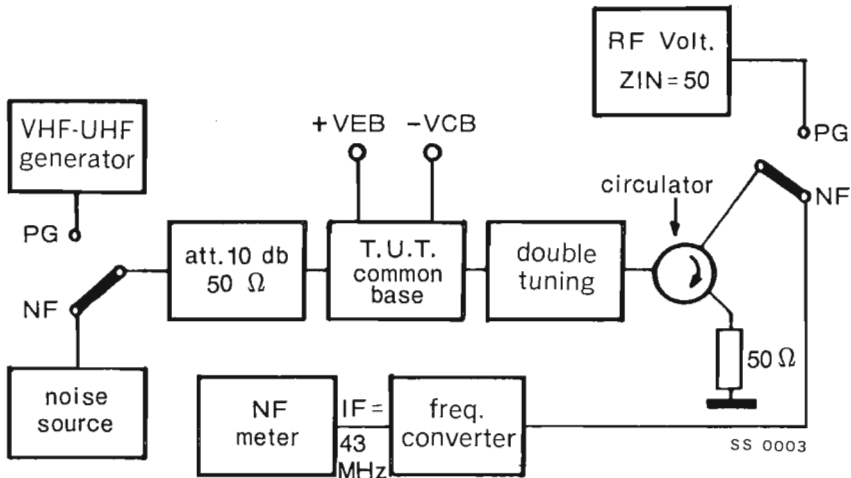


Power rating chart



TEST CIRCUIT

Power gain, AGC and noise figure ($f = 200$ to 800 MHz)



INTEGRATED CIRCUITS



LINEAR INTEGRATED CIRCUIT

TAA 550 TBA 271

VOLTAGE STABILIZER

- LOW TEMPERATURE COEFFICIENT
- LOW ZENER RESISTANCE

The TAA 550/TBA 271 is a monolithic integrated voltage stabilizer in a TO-18 two pins metal case. It is especially designed as voltage supplier for varicap diodes in television tuners.

The TAA 550/TBA 271 is supplied in 3 groups of stabilized voltage identified by a letter after the code, as shown in the "ORDERING NUMBERS"

ABSOLUTE MAXIMUM RATINGS

I_z	Zener current at $T_{case} \leq 70^\circ\text{C}$	15 mA
T_{stg}	Storage temperature	-20 to 150 °C
T_{op}	Operating temperature	*

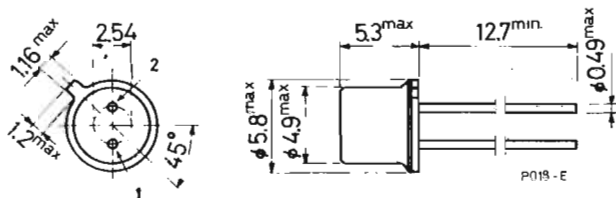
* Refer to "Power rating chart" (Fig. 1).

ORDERING NUMBERS: TAA 550 A or TBA 271 A (for V_s range : 30-32 V)
TAA 550 B or TBA 271 B (for V_s range : 32-34 V)
TAA 550 C or TBA 271 C (for V_s range : 34-36 V)

MECHANICAL DATA

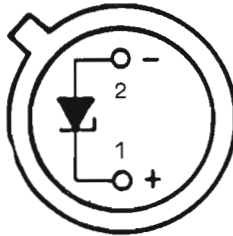
Dimensions in mm

Lead 1 connected to case



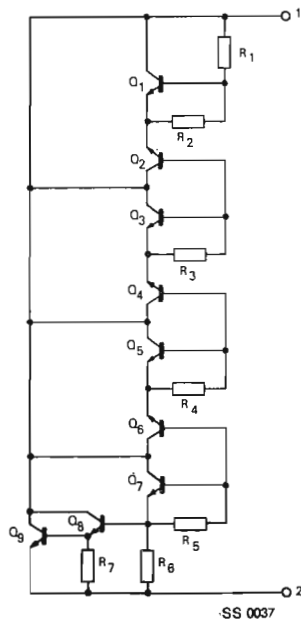
TAA 550 TBA 271

CONNECTION DIAGRAM (bottom view)



SS 0036

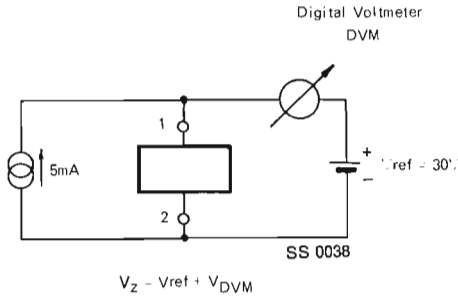
SCHEMATIC DIAGRAM



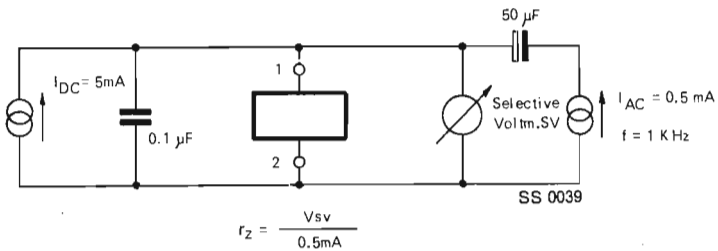
SS 0037

TEST CIRCUITS

Circuit No. 1 (for V_z measurement)



Circuit No. 2 (for r_z measurement)



TAA 550 TBA 271

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	150	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	400	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_z Zener voltage	$I_z = 5\text{ mA}$ (circuit No. 1) for TAA 550 A/TBA 271 A for TAA 550 B/TBA 271 B for TAA 550 C/TBA 271 C	30 32 34	31 33 35	32.2 34.2 36	V V V
r_z Zener dynamic resistance	$I_z = 5\text{ mA}$ $I_{AC} = 0.5\text{ mA}$ $f = 1\text{ kHz}$ (circuit No. 2)		10	25	Ω
$\frac{\Delta V_z}{\Delta T_{amb}}$ Temperature coefficient	$I_z = 5\text{ mA}$ $\Delta T_{amb} = 0\text{ to }50\text{ °C}$	-3.2		+1.6	mV/°C

TAA 550 TBA 271

Fig. 1 - Power rating chart

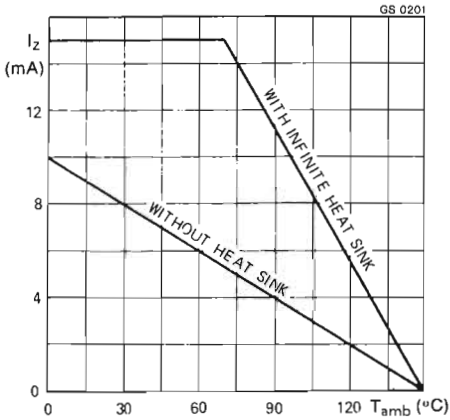


Fig. 2 - Typical zener dynamic resistance vs zener current

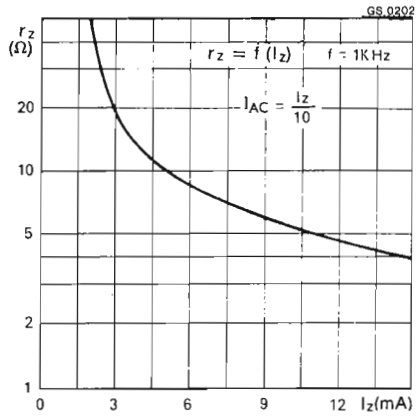


Fig. 3 - Typical temperature coefficient

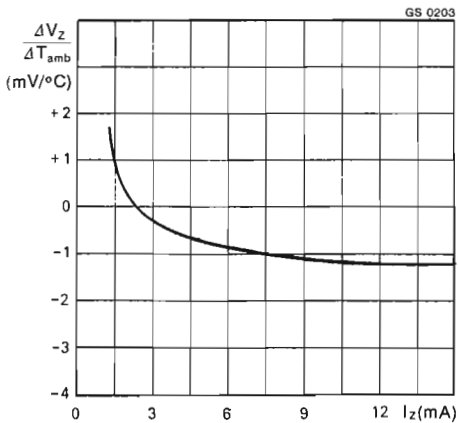
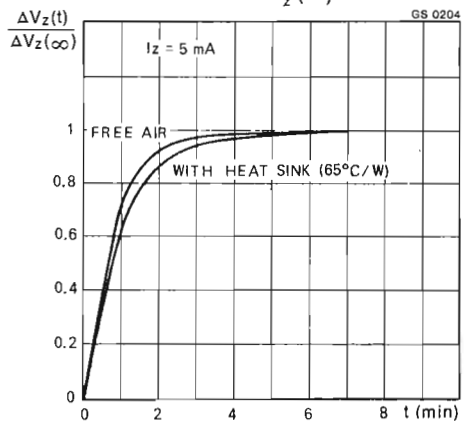
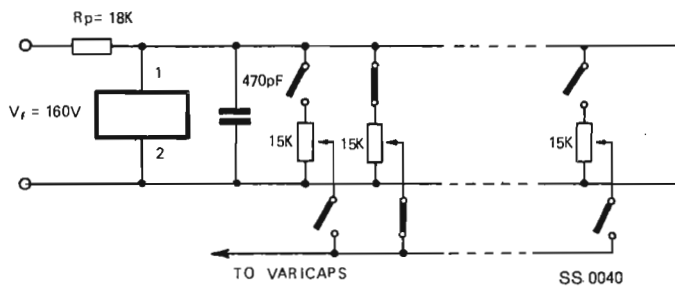


Fig. 4 - Typical $\frac{\Delta V_z(t)}{\Delta V_z(\infty)}$ vs time



TAA 550 TBA 271

TYPICAL APPLICATION



LINEAR INTEGRATED CIRCUIT

TAA 611A

AUDIO AMPLIFIER

- OUTPUT POWER 1.8 W (9 V - 4 Ω)
- LOW DISTORTION
- LOW QUIESCENT CURRENT
- HIGH INPUT IMPEDANCE

The TAA 611 A is a monolithic integrated circuit in a 14-lead quad in-line plastic package or in a TO-96 metal case.

It is particularly designed for use in radio receivers and record-players as audio amplifier. The usable range of supply voltage varies from 6 V to 10 V and the circuit requires a minimum number of external components.

ABSOLUTE MAXIMUM RATINGS

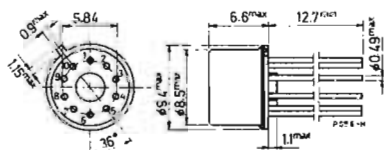
ABSOLUTE MAXIMUM RATINGS		TAA 611 A12	TAA 611 A55
V_s	Supply voltage	12 V	
V_i^*	Input voltage	-0.5 to 12 V	
I_o	Output peak current	1 A	
P_{tot}	Power dissipation at $T_{amb} \leq 25^\circ\text{C}$ at $T_{case} \leq 70^\circ\text{C}$	1.35 W	0.57 W
		—	1.15 W
T_{stg}	Storage temperature	-55 to 125 °C	-55 to 150 °C
T_j	Junction temperature	150 °C	

* For $V_s < 12\text{ V}$, $V_{i\max} = V_s$

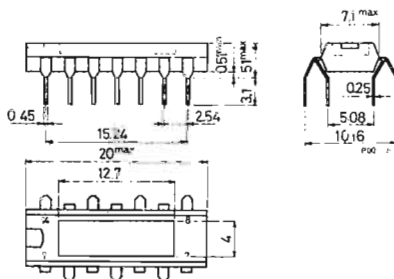
ORDERING NUMBERS: TAA 611 A55 (for TO-100 metal case)
TAA 611 A12 (for quad in-line plastic package)

MECHANICAL DATA

Dimensions in mm



TAA 611 A55

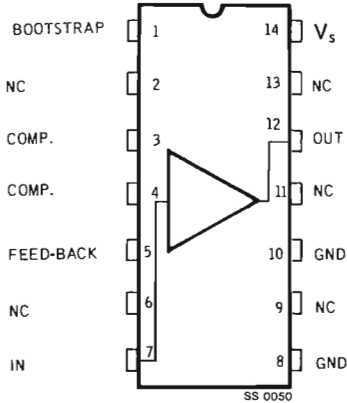


TAA 611 A12

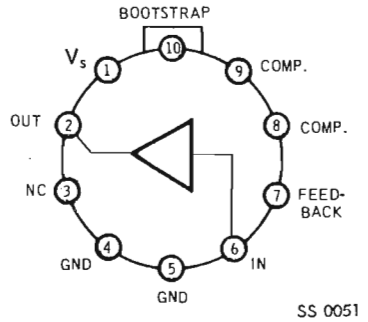
TAA 611A

CONNECTION DIAGRAMS (top views)

For TAA 611 A12

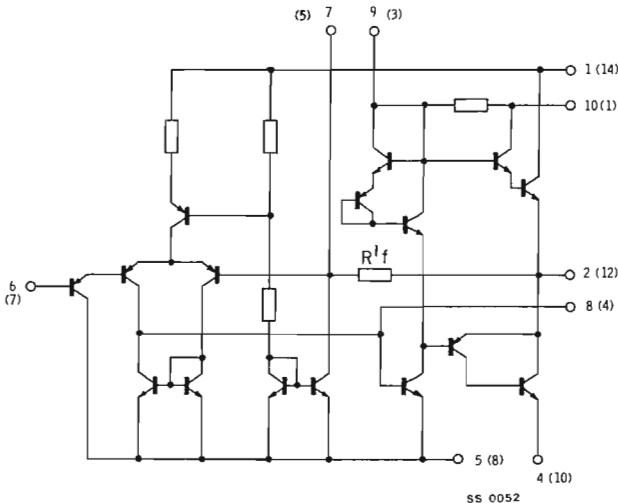


For TAA 611 A55



SS 0051

SCHEMATIC DIAGRAM



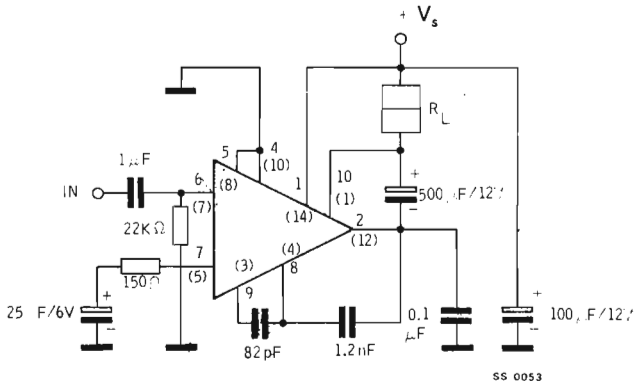
SS 0052

The pin numbers in brackets refer to the TAA 611 A12 and those without brackets refer to the TAA 611 A55.

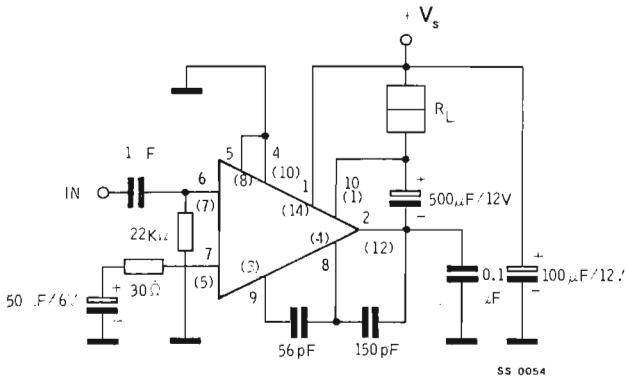
TAA 611A

TEST CIRCUITS

Circuit No. 1 ($G_v = 50$)



Circuit No. 2 ($G_v = 250$)



TAA 611A

THERMAL DATA (maximum values)

		TAA 611 A12	TAA 611 A55
$R_{th\ j-case}$	Thermal resistance junction-case	—	50 °C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	93 °C/W	220 °C/W

ELECTRICAL CHARACTERISTICS

($T_{amb} = 25\text{ °C}$, $V_s = 9\text{ V}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit		
V_o	Quiescent output voltage		4.8		V		
I_d	Total quiescent drain current		3		mA		
I_d	Quiescent drain current of output transistors		1		mA		
I_d	Drain current	$P_o = 1.15\text{ W}$		$R_L = 8\ \Omega$	170	mA	
I_b	Input bias current		60		nA		
P_o^*	Output power	$d = 2\%$	$V_s = 6\text{ V}$	$R_L = 4\ \Omega$	0.50	W	
			$V_s = 6\text{ V}$	$R_L = 8\ \Omega$	0.35	W	
			$V_s = 9\text{ V}$	$R_L = 4\ \Omega$	1.4	W	
			$V_s = 9\text{ V}$	$R_L = 8\ \Omega$	0.9	W	
			$d = 10\%$	$V_s = 6\text{ V}$	$R_L = 4\ \Omega$	0.65	W
				$V_s = 6\text{ V}$	$R_L = 8\ \Omega$	0.45	W
				$V_s = 9\text{ V}$	$R_L = 4\ \Omega$	1.8	W
				$V_s = 9\text{ V}$	$R_L = 8\ \Omega$	0.85	1.15
R_f'	Internal feedback resistance (see schematic diagram)		7.5		k Ω		
Z_i	Input impedance (open loop)		0.75		M Ω		

TAA 611A

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
d Distortion	Test circuit 1				
	$P_o = 50 \text{ mW}$ $V_s = 9 \text{ V}$				
	$R_L = 8 \Omega$ $f = 1 \text{ kHz}$		0.4		%
	$P_o = 0.5 \text{ W}$ $V_s = 9 \text{ V}$				
G_v Voltage gain (open loop)	Test circuit 2				
	$P_o = 50 \text{ mW}$ $V_s = 9 \text{ V}$				
	$R_L = 8 \Omega$ $f = 1 \text{ kHz}$		1.7		%
	$P_o = 0.5 \text{ W}$ $V_s = 9 \text{ V}$				
$R_L = 8 \Omega$ $f = 1 \text{ kHz}$		1.2		%	
G_v Voltage gain (open loop)	$R_L = 8 \Omega$		68		dB

* External heat-sink not required except for TAA 611 A55 at $V_s = 9 \text{ V}$, $R_L = 4 \Omega$.

Fig. 1 - Typical output power vs load resistance

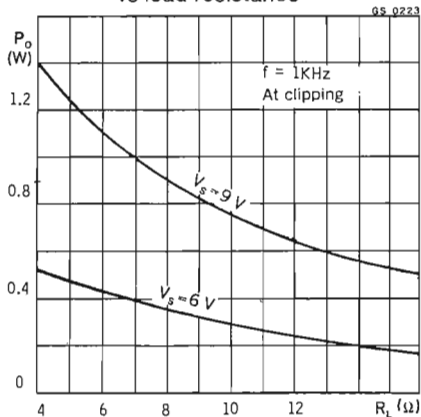
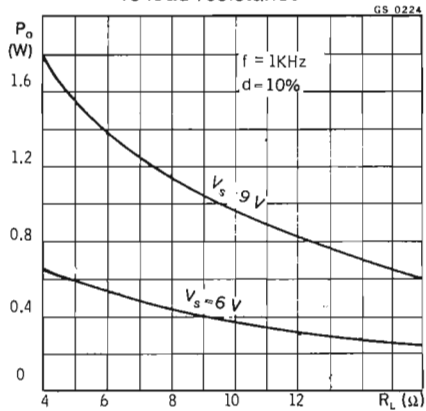


Fig. 2 - Typical output power vs load resistance



TAA 611A

Fig. 3 - Typical distortion vs output power

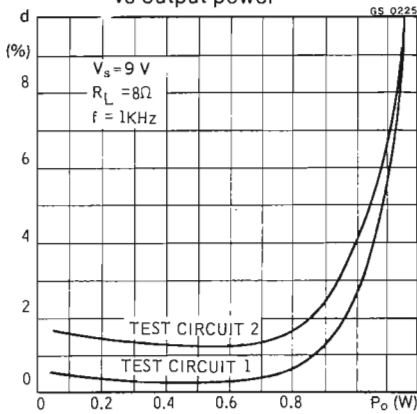


Fig. 4 - Typical distortion vs output power

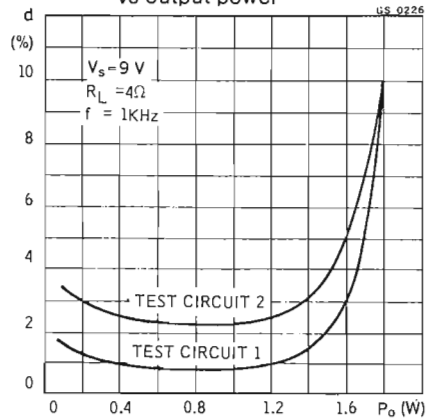


Fig. 5 - Typical relative frequency response

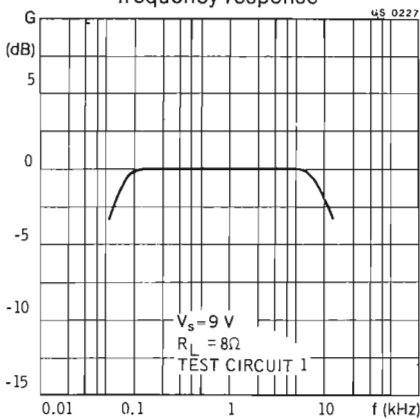
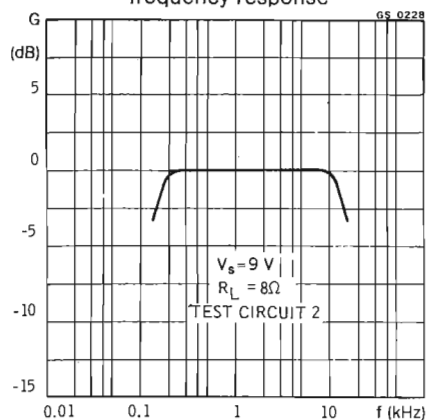
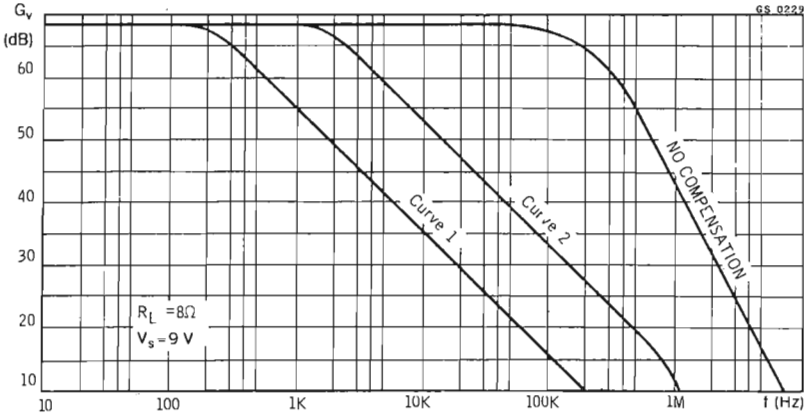


Fig. 6 - Typical relative frequency response



TAA 611A

Fig. 7 - Typical voltage gain (open loop) vs frequency



Curve 1: TAA611 A 55, $C_{9-8} = 82\text{pF}$ $C_{8-2} = 1.2\text{ nF}$ $C_{10-1} = 0.1\mu\text{F}$
 TAA611 A 12, $C_{3-4} = 82\text{pF}$ $C_{4-12} = 1.2\text{ nF}$ $C_{1-14} = 0.1\mu\text{F}$
 Curve 2: TAA611 A 55, $C_{9-8} = 56\text{pF}$ $C_{8-2} = 150\text{ pF}$ $C_{10-1} = 0.1\mu\text{F}$
 TAA611 A 12, $C_{3-4} = 56\text{pF}$ $C_{4-12} = 150\text{pF}$ $C_{1-14} = 0.1\mu\text{F}$

Fig. 8 - Typical output power vs input voltage

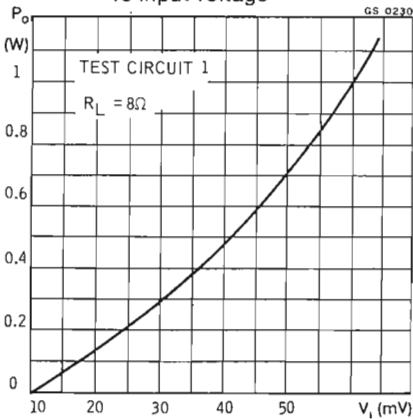
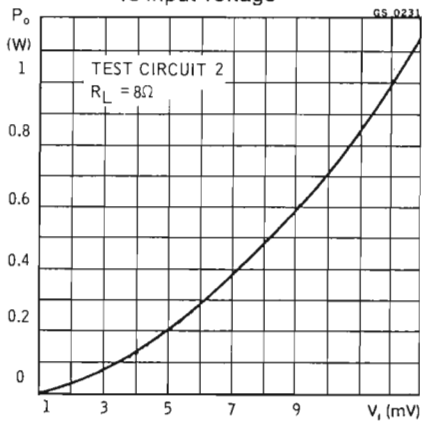


Fig. 9 - Typical output power vs input voltage



TAA 611A

Fig. 10 - Typical power dissipation and efficiency vs output power

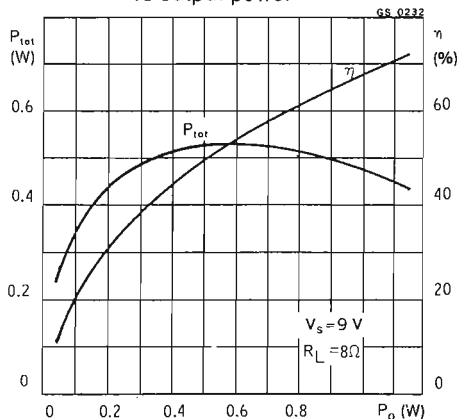


Fig. 11 - Typical power dissipation and efficiency vs output power

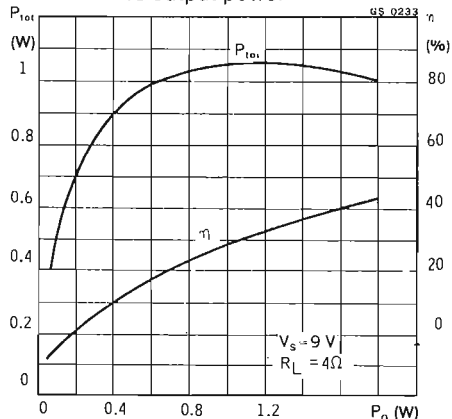


Fig. 12 - Typical power dissipation and efficiency vs output power

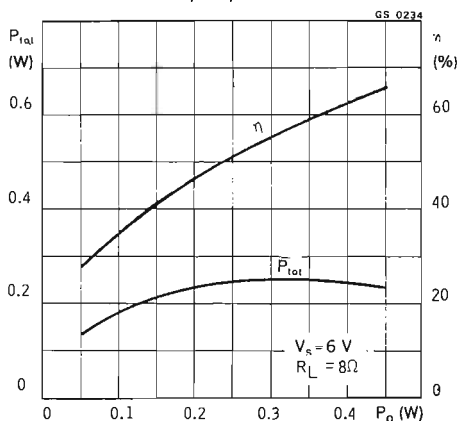
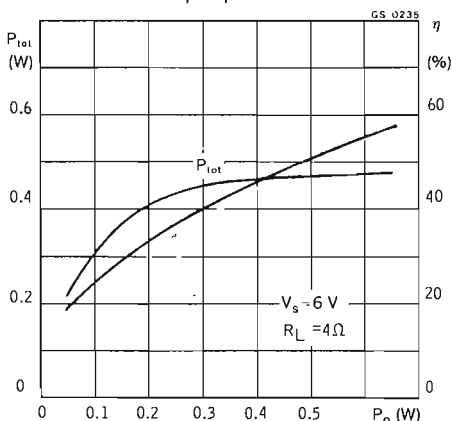


Fig. 13 - Typical power dissipation and efficiency vs output power



TAA 611A

Fig. 14 - Typical drain current vs output power

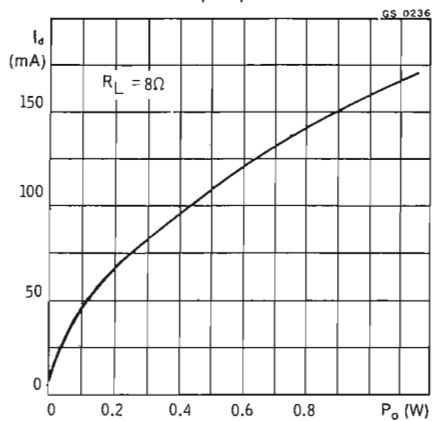


Fig. 15 - Maximum power dissipation vs load resistance

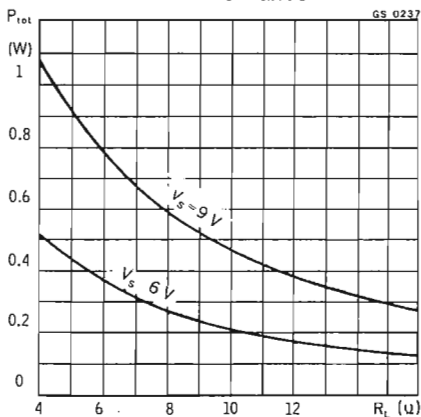


Fig. 16 - Power rating chart (TAA 611 A55)

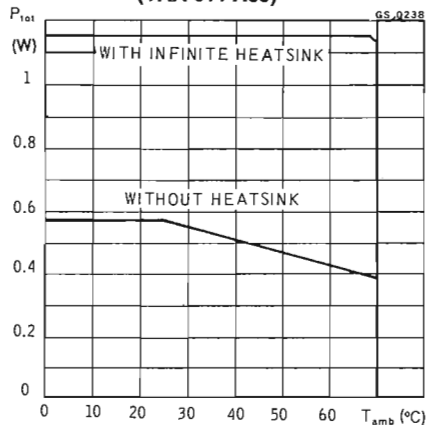
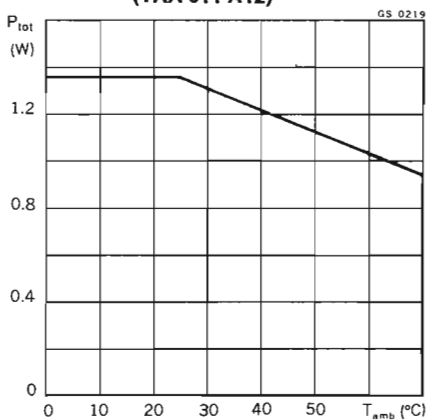


Fig. 17 - Power rating chart (TAA 611 A12)



TAA 611A

Fig. 18 - Typical quiescent drain current vs supply voltage

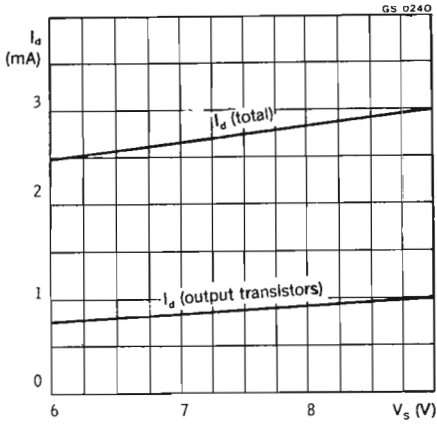


Fig. 19 - Typical quiescent drain current vs ambient temperature

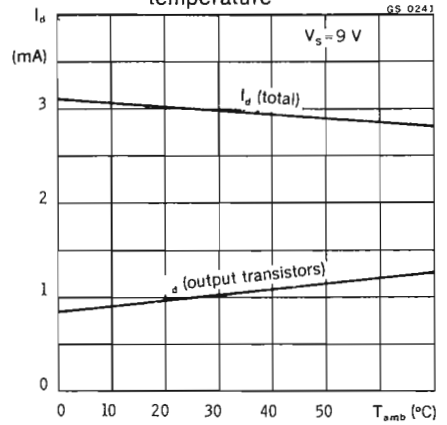
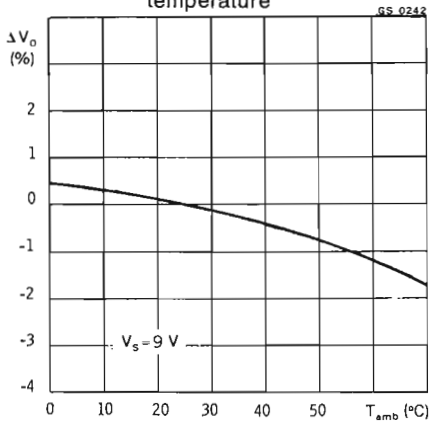


Fig. 20 - Typical quiescent output voltage vs ambient temperature



TAA 611A

TYPICAL APPLICATIONS

Fig. 21 - Audio amplifier for record-player.

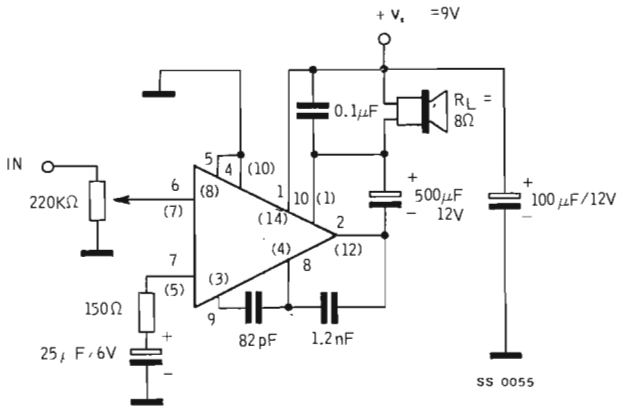
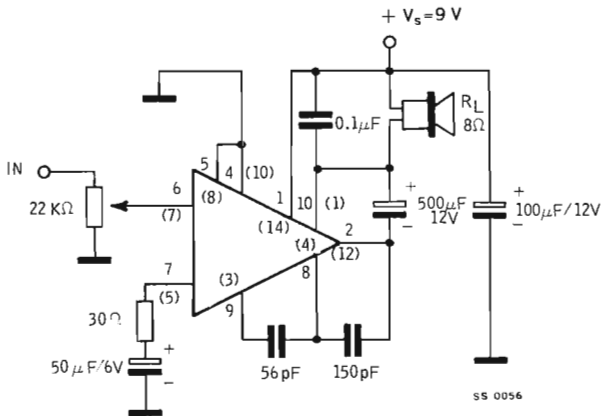


Fig. 22 - Audio amplifier for radio.



The pin numbers in brackets refer to the TAA 611 A12 and those without brackets refer to the TAA 611 A55.

TAA 611B

LINEAR INTEGRATED CIRCUIT

AUDIO AMPLIFIER

- OUTPUT POWER 2.1 W (12 V - 8 Ω)
- LOW DISTORTION
- LOW QUIESCENT CURRENT
- HIGH INPUT IMPEDANCE

The TAA 611 B is a monolithic integrated circuit in a 14-lead quad in-line plastic package.

It is particularly designed for use in radio receivers and record-players as audio amplifier. The usable range of supply voltage varies from 6 V to 15 V and the circuit requires a minimum number of external components.

ABSOLUTE MAXIMUM RATINGS

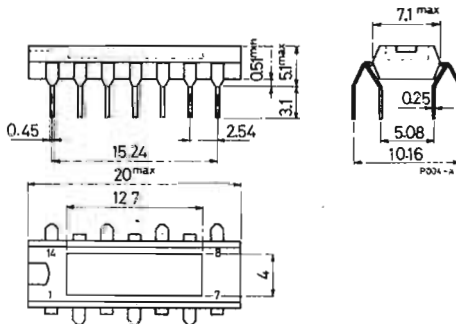
V_s	Supply voltage	15	V
V_i^*	Input voltage	-0.5 to 15	V
I_o	Output peak current	1	A
P_{tot}	Power dissipation at $T_{amb} \leq 25$ °C	1.35	W
T_{stg}	Storage temperature	-55 to 125	°C
T_j	Junction temperature	150	°C

* For $V_s < 15$ V, $V_{i\ max} = V_s$

ORDERING NUMBER: TAA 611 B12

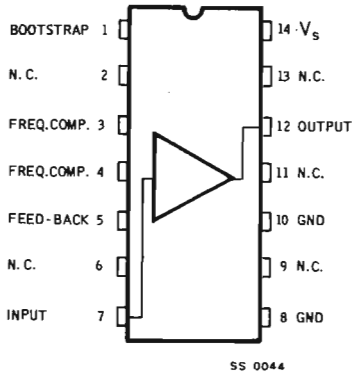
MECHANICAL DATA

Dimensions in mm

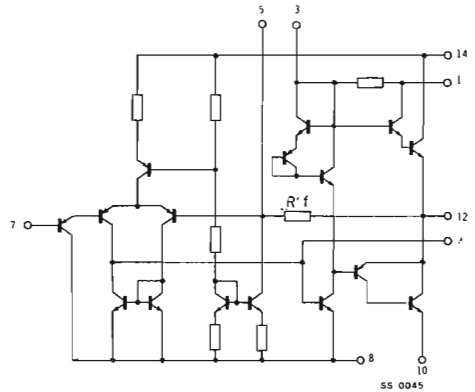


TAA 611B

CONNECTION DIAGRAM (top view)

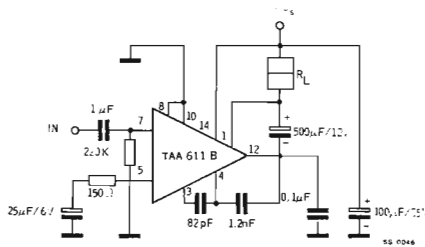


SCHEMATIC DIAGRAM

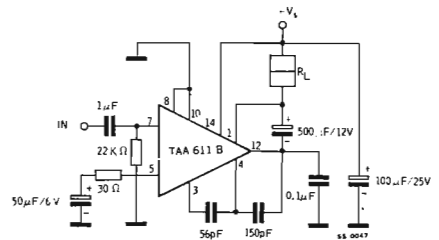


TEST CIRCUITS

Circuit No. 1 ($G_v = 50$)



Circuit No. 2 ($G_v = 250$)



TAA 611B

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	93 °C/W
-----------------	-------------------------------------	-----	---------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
V_o	Quiescent output voltage	$V_s = 9\text{ V}$ $V_s = 12\text{ V}$		4.8 6.3		V V
I_d	Total quiescent drain current	$V_s = 9\text{ V}$ $V_s = 12\text{ V}$		3 3.5		mA mA
I_d	Quiescent drain current of output transistors	$V_s = 9\text{ V}$ $V_s = 12\text{ V}$		1 1.2		mA mA
I_d	Drain current	$R_L = 8\ \Omega$ $P_o = 1.15\text{ W}$ $V_s = 9\text{ V}$ $P_o = 2.1\text{ W}$ $V_s = 12\text{ V}$		170 235		mA mA
I_b	Input bias current	$V_s = 9\text{ V}$ $V_s = 12\text{ V}$		60 75		nA nA
P_o	Output power	$d = 2\%$ $R_L = 8\ \Omega$ $V_s = 9\text{ V}$ $V_s = 12\text{ V}$ $d = 10\%$ $R_L = 8\ \Omega$ $V_s = 9\text{ V}$ $V_s = 12\text{ V}$		0.9 1.7 1.15 2.1		W W W W
R_f'	Internal feedback resistance (see schematic diagram)			7.5		k Ω
Z_i	Input impedance	open loop		0.75		M Ω

TAA 611B

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
d Distortion	Test circuit 1				
	$R_L = 8 \Omega$ $f = 1 \text{ kHz}$				
	$P_o = 50 \text{ mW}$ $V_s = 9 \text{ V}$		0.4		%
	$P_o = 50 \text{ mW}$ $V_s = 12 \text{ V}$		0.3		%
	$P_o = 0.5 \text{ W}$ $V_s = 9 \text{ V}$		0.3		%
	$P_o = 1 \text{ W}$ $V_s = 12 \text{ V}$		0.2		%
	Test circuit 2				
	$R_L = 8 \Omega$ $f = 1 \text{ kHz}$				
	$P_o = 50 \text{ mW}$ $V_s = 9 \text{ V}$		1.7		%
	$P_o = 50 \text{ mW}$ $V_s = 12 \text{ V}$		1.5		%
$P_o = 0.5 \text{ W}$ $V_s = 9 \text{ V}$		1.2		%	
$P_o = 1 \text{ W}$ $V_s = 12 \text{ V}$		1		%	
G_v Voltage gain (open loop)	$R_L = 8 \Omega$ $V_s = 9 \text{ V}$		68		dB
	$R_L = 8 \Omega$ $V_s = 12 \text{ V}$		70		dB

TAA 611B

Fig. 1 - Typical output power vs load resistance

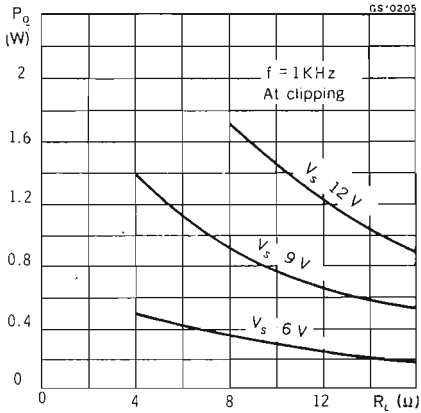


Fig. 2 - Typical output power vs load resistance

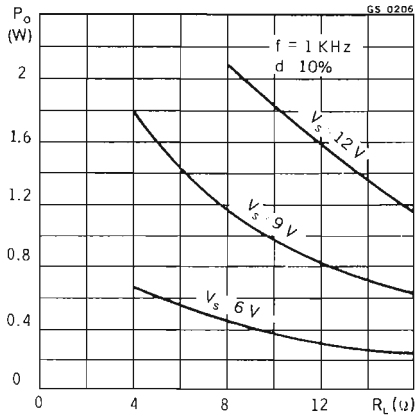


Fig. 3 - Typical distortion vs output power

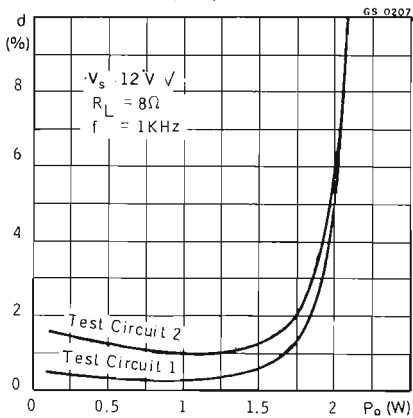
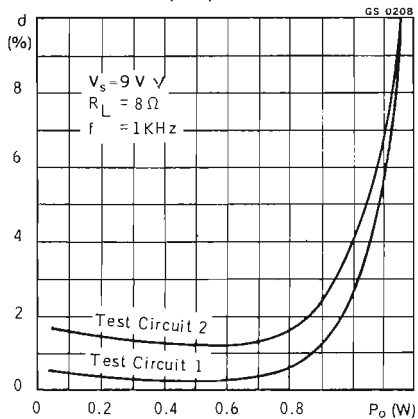


Fig. 4 - Typical distortion vs output power



TAA 611B

Fig. 7 - Typical voltage gain (open loop) vs frequency

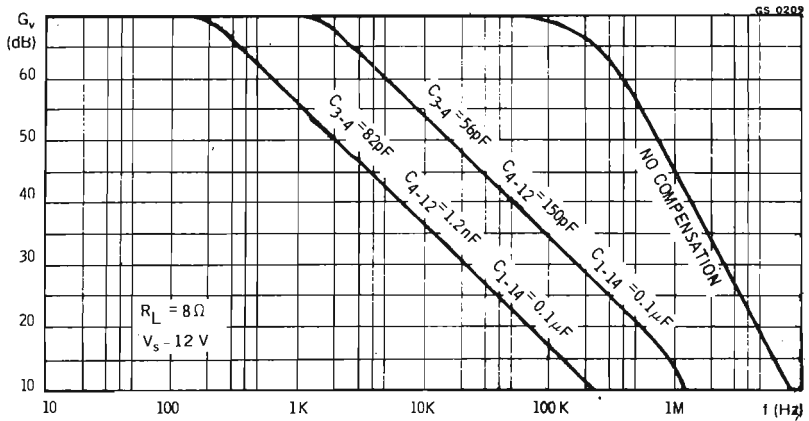


Fig. 6 - Typical relative frequency response

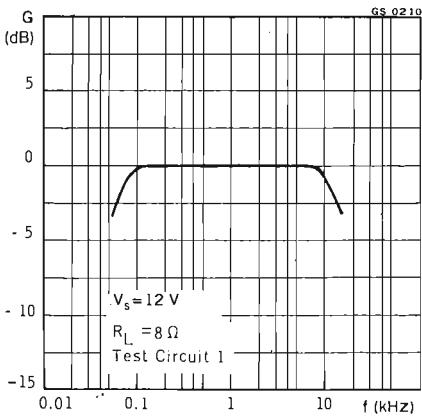
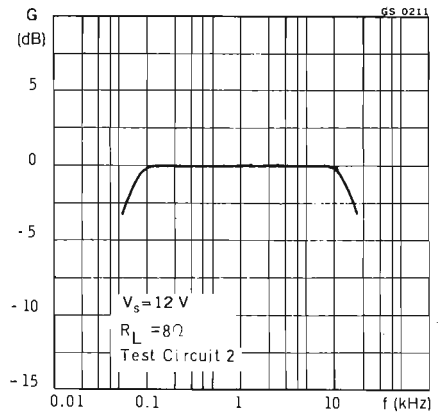


Fig. 7 - Typical relative frequency response



TAA 611B

Fig. 8 - Typical output power vs input voltage

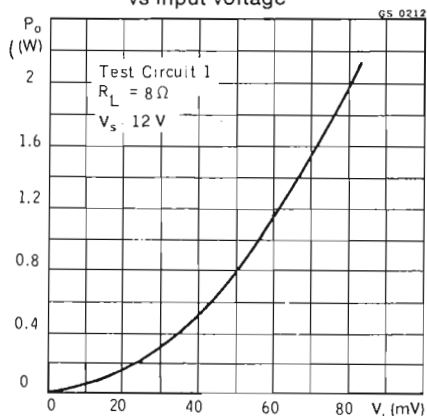


Fig. 9 - Typical output power vs input voltage

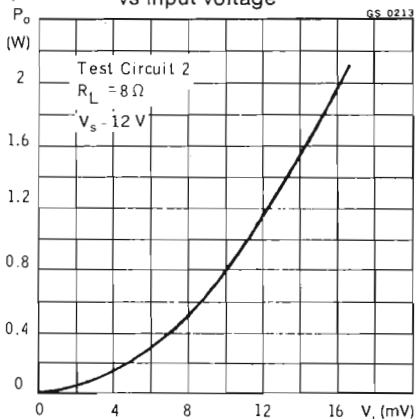


Fig. 10 - Typical power dissipation and efficiency vs output power

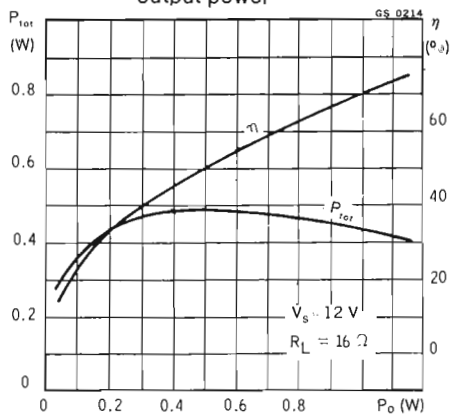
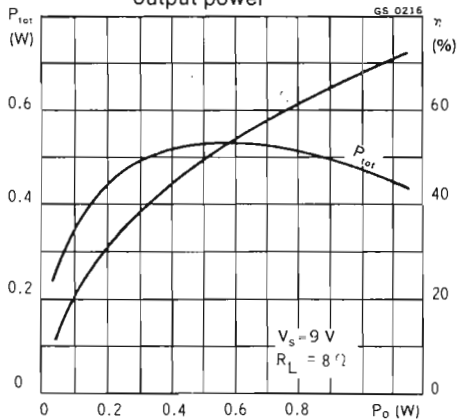


Fig. 11 - Typical power dissipation and efficiency vs output power



TAA 611B

Fig. 12 - Typical power dissipation and efficiency vs output power

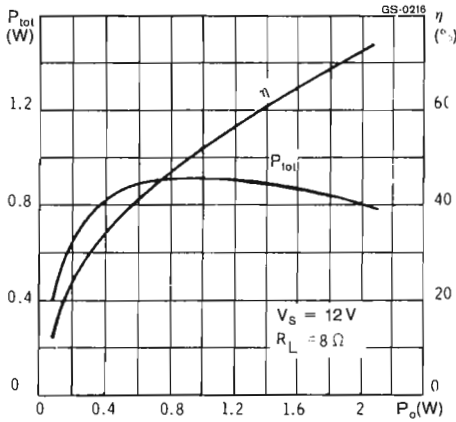


Fig. 13 - Typical drain current vs output power

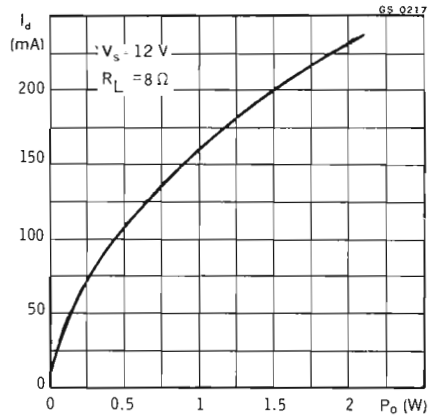


Fig. 14 - Maximum power dissipation vs load resistance

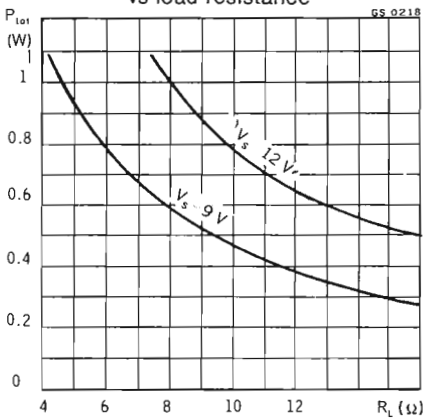
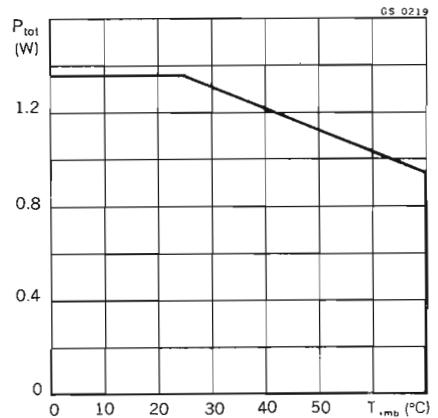


Fig. 15 - Power rating chart



TAA 611B

Fig. 16 - Typical quiescent drain current vs supply voltage

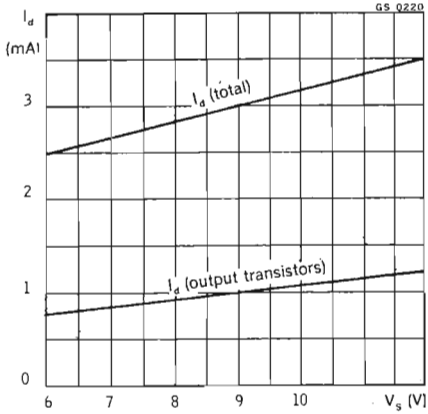


Fig. 17 - Typical quiescent drain current vs ambient temperature

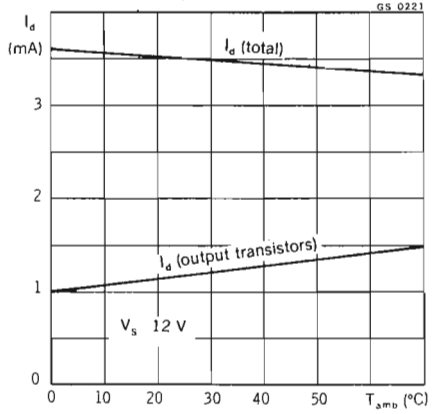
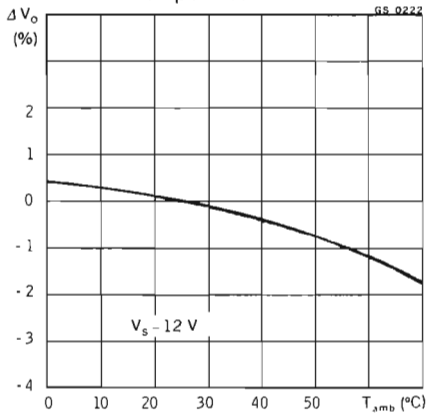


Fig. 18 - Quiescent output voltage variation vs ambient temperature



TAA 611B

TYPICAL APPLICATIONS

Fig. 19 - Audio amplifier for radio.

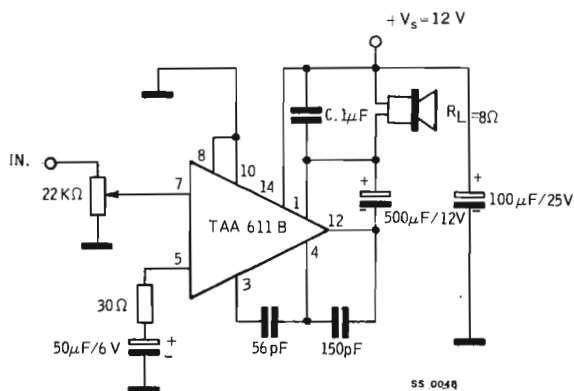
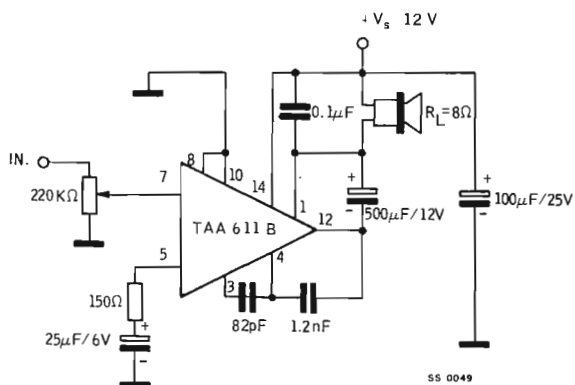


Fig. 20 - Audio amplifier for record-player.



TAA 611C

LINEAR INTEGRATED CIRCUIT

AUDIO AMPLIFIER

- OUTPUT POWER 3.3 W (15 V - 8 Ω)
- LOW DISTORTION
- LOW QUIESCENT CURRENT
- SELF CENTERING BIAS
- HIGH IMPEDANCE

The TAA 611C is a monolithic integrated circuit in a 14-lead quad in-line plastic package with external heat-sink.

It is particularly designed for use as audio amplifier in radio receivers, record players and portable TV sets. The usable range of supply voltage varies from 6 to 16 V, and the circuit requires a minimum number of external components.

The package has very low thermal resistance. To decrease the thermal resistance further an external heat-sink can easily be mounted by means of ordinary hardware.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage (no signal)	22	V
V_s	Operating supply voltage	18	V
V_i^*	Input voltage	-0.5 to 20	V
I_o	Output peak current	1	A
P_{tot}	Power dissipation at $T_{amb} \leq 25^\circ\text{C}$	2	W
	at $T_{case} \leq 70^\circ\text{C}$ (with ∞ h.s.)	3	W
T_{stg}	Storage temperature	-55 to 125	$^\circ\text{C}$
T_j	Junction temperature	150	$^\circ\text{C}$

* For $V_s < 20$ V, $V_{i\ max} = V_s$.

ORDERING NUMBERS:

TAA 611 C72 (for quad in-line plastic package with spacer)

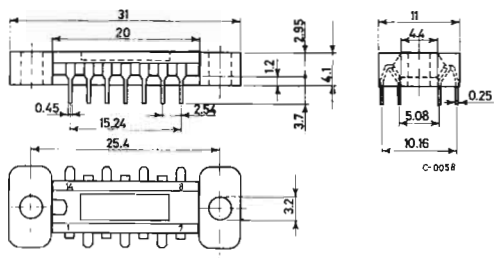
TAA 611 CX1 (for quad in-line plastic package with external bar)

TAA 611 C11 (for quad in-line plastic package with inverted external bar)

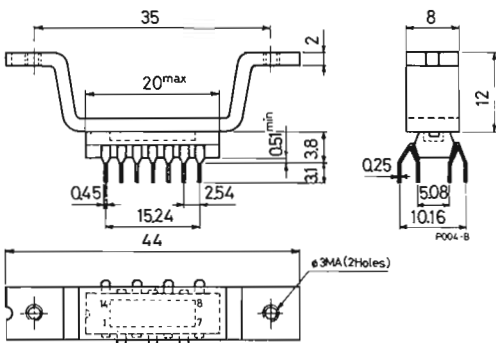
TAA 611C

MECHANICAL DATA (Dimensions in mm)

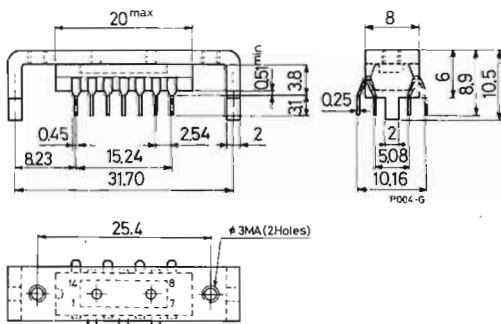
Quad in-line plastic package with spacer for TAA 611 C72 (see also "MOUNTING INSTRUCTIONS")



Quad in-line plastic package with external bar for TAA 611 CX1

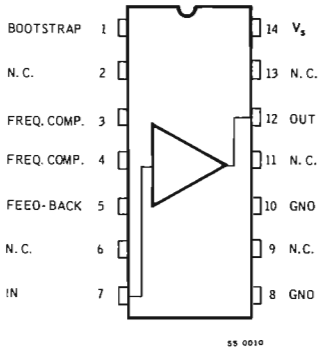


Quad in-line plastic package with inverted external bar for TAA 611 C11

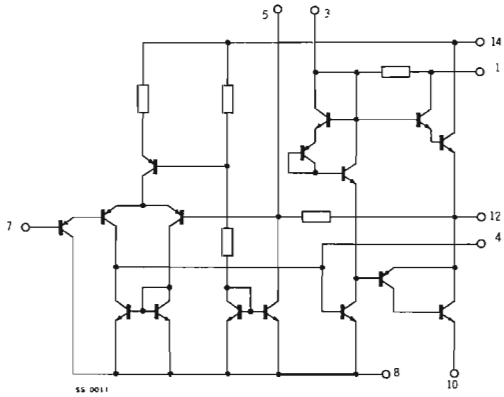


TAA 611C

CONNECTION DIAGRAM (top view)

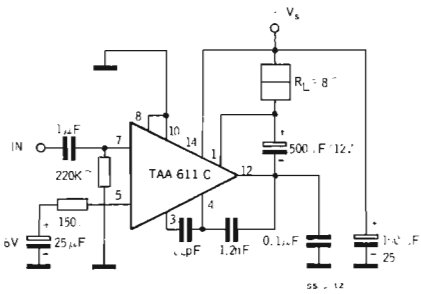


SCHEMATIC DIAGRAM

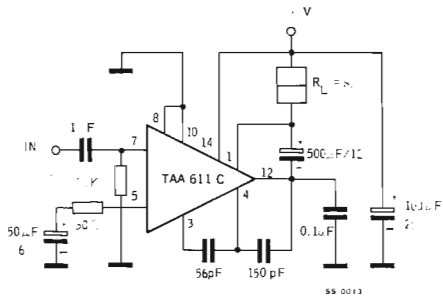


TEST CIRCUITS

Circuit No. 1 ($G_v = 50$)



Circuit No. 2 ($G_v = 250$)



TAA 611C

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	17	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	63	°C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_o Quiescent output voltage	$V_s = 12\text{ V}$		6.3		V
	$V_s = 15\text{ V}$		7.9		V
I_d Total quiescent drain current	$V_s = 12\text{ V}$		3.5		mA
	$V_s = 15\text{ V}$		4		mA
I_d Quiescent drain current of output transistors	$V_s = 12\text{ V}$		1.2		mA
	$V_s = 15\text{ V}$		1.8		mA
I_d Drain current	$V_s = 12\text{ V}$ $P_o = 2.1\text{ W}$		235		mA
	$R_L = 8\ \Omega$ $P_o = 3.3\text{ W}$		300		mA
I_b Input bias current	$V_s = 12\text{ V}$		75		nA
	$V_s = 15\text{ V}$		95		nA
P_o^* Output power	$d = 2\%$				
	$V_s = 9\text{ V}$ $R_L = 4\ \Omega$		1.4		W
	$V_s = 9\text{ V}$ $R_L = 8\ \Omega$		0.9		W
	$V_s = 12\text{ V}$ $R_L = 8\ \Omega$		1.7		W
	$V_s = 15\text{ V}$ $R_L = 8\ \Omega$		2.8		W
	$V_s = 15\text{ V}$ $R_L = 16\ \Omega$		1.6		W
	$d = 10\%$				
	$V_s = 9\text{ V}$ $R_L = 4\ \Omega$		1.8		W
	$V_s = 9\text{ V}$ $R_L = 8\ \Omega$		1.15		W
	$V_s = 12\text{ V}$ $R_L = 8\ \Omega$		2.1		W
	$V_s = 15\text{ V}$ $R_L = 8\ \Omega$	2.5	3.3		W
	$V_s = 15\text{ V}$ $R_L = 16\ \Omega$		1.9		W

* External heat-sink not required except for the conditions $V_s = 15\text{ V}$, $R_L = 8\ \Omega$.

TAA 611C

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min. Typ. Max.	Unit
R'_f	Internal feedback resistance (see schematic diagram)		7.5	Ω
Z_i	Input impedance	open loop	0.75	M Ω
d	Distortion	Circuit No. 1 $R_L = 8 \Omega$ $f = 1 \text{ kHz}$ $V_s = 12 \text{ V}$ $P_o = 50 \text{ mW}$ $V_s = 15 \text{ V}$ $P_o = 50 \text{ mW}$ $V_s = 12 \text{ V}$ $P_o = 1 \text{ W}$ $V_s = 15 \text{ V}$ $P_o = 1 \text{ W}$ Circuit No. 2 $R_L = 8 \Omega$ $f = 1 \text{ kHz}$ $V_s = 12 \text{ V}$ $P_o = 50 \text{ mW}$ $V_s = 15 \text{ V}$ $P_o = 50 \text{ mW}$ $V_s = 12 \text{ V}$ $P_o = 1 \text{ W}$ $V_s = 15 \text{ V}$ $P_o = 1 \text{ W}$	0.3 0.3 0.2 0.2 1.5 1.5 1 1	% % % % % % % %
G_v	Voltage gain (open loop)	$V_s = 12 \text{ V}$ $R_L = 8 \Omega$ $V_s = 15 \text{ V}$ $R_L = 8 \Omega$	70 72	dB dB

TAA 611C

Fig. 1 - Typical distortion vs output power

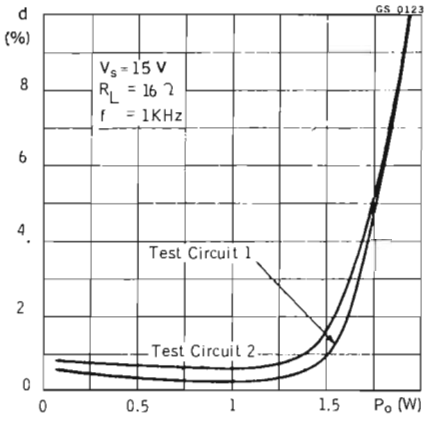


Fig. 2 - Typical distortion vs output power

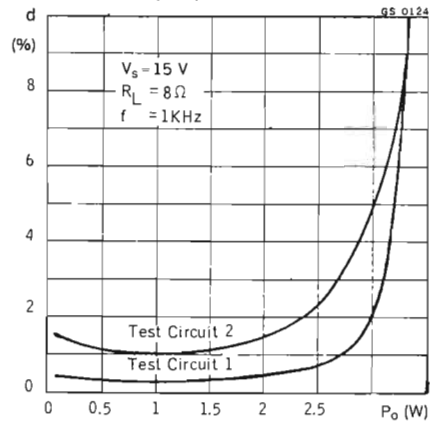


Fig. 3 - Typical distortion vs output power

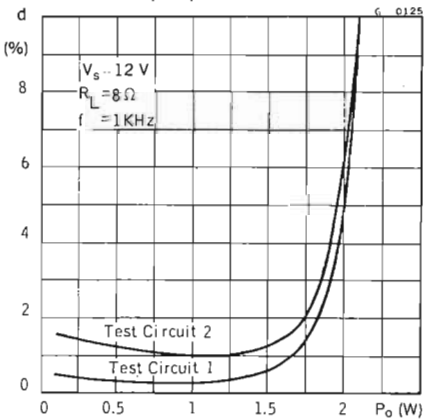
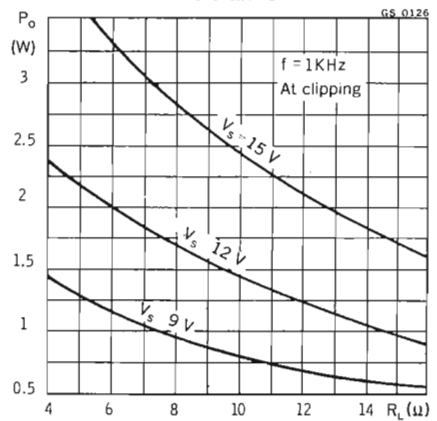


Fig. 4 - Typical output power vs load resistance



TAA 611C

Fig. 5 - Typical output power vs load resistance

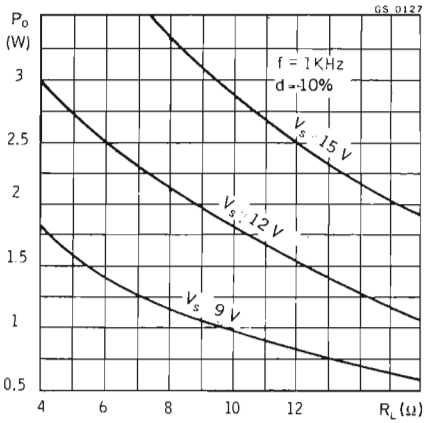


Fig. 6 - Maximum power dissipation vs load resistance

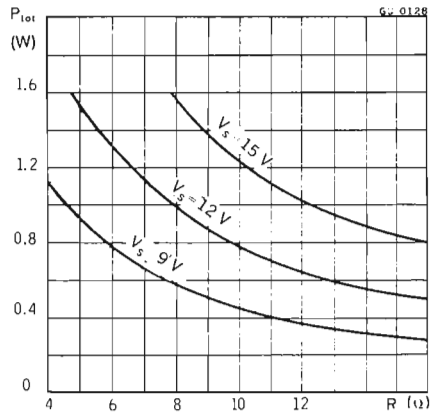
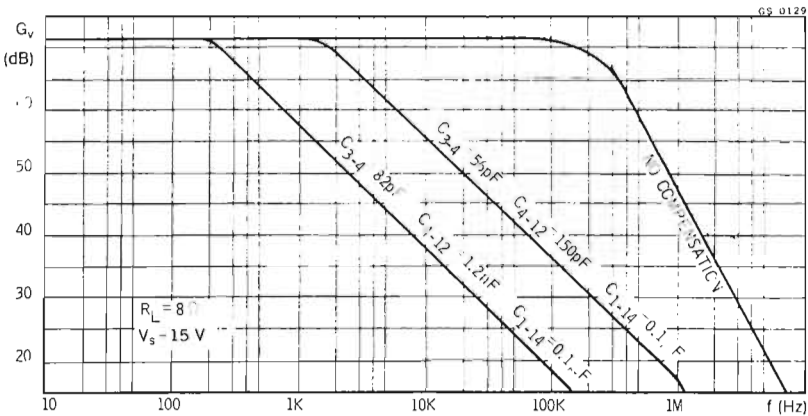


Fig. 5 - Typical voltage gain (open loop) vs frequency



TAA 611C

Fig. 8 - Typical relative frequency response

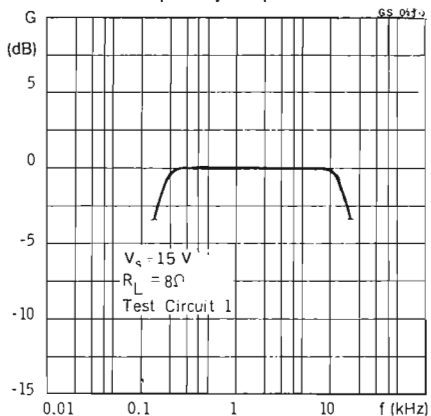


Fig. 9 - Typical relative frequency response

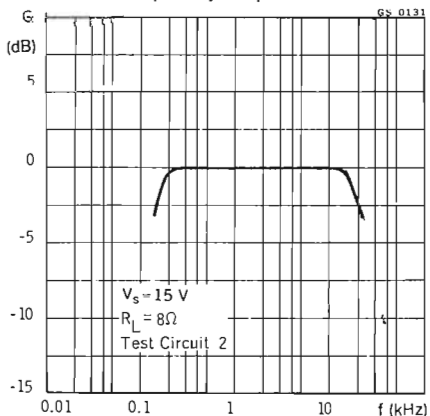


Fig. 10 - Typical output power vs input voltage

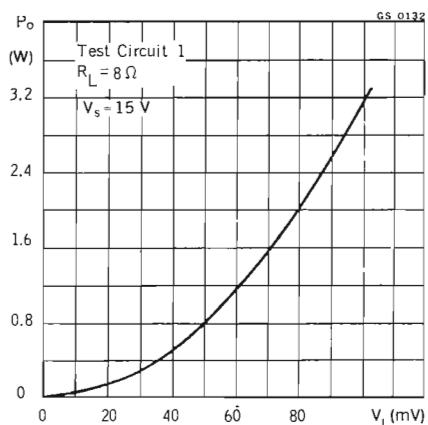
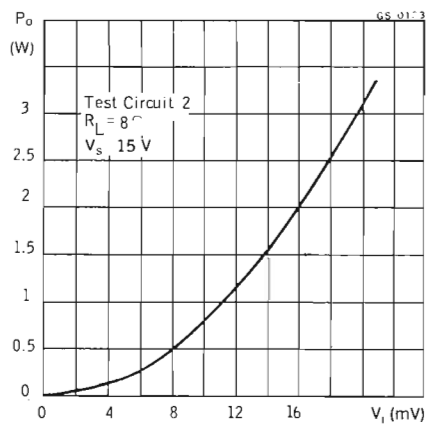


Fig. 11 - Typical output power vs input voltage



TAA 611C

Fig. 12 - Typical power dissipation and efficiency vs output power

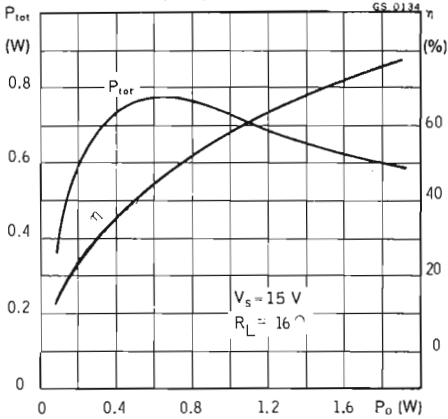


Fig. 13 - Typical power dissipation and efficiency vs output power

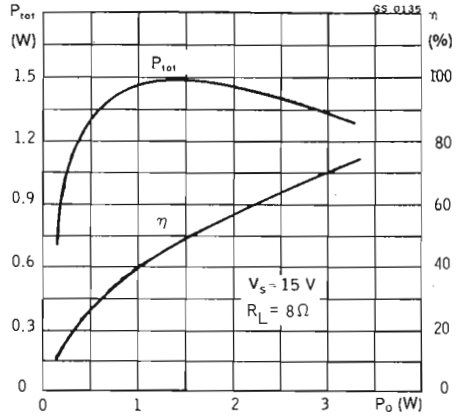


Fig. 14 - Typical power dissipation and efficiency vs output power

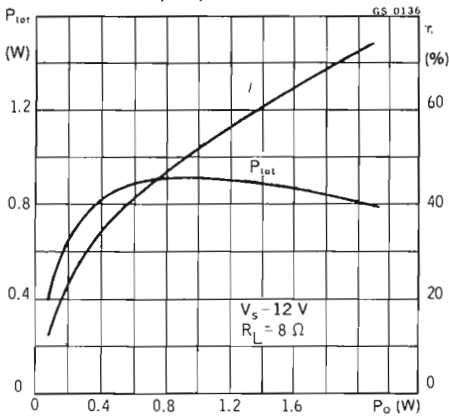
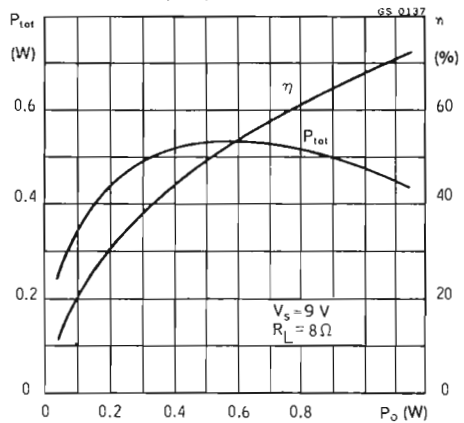


Fig. 15 - Typical power dissipation and efficiency vs output power



TAA 611C

Fig. 16 - Typical drain current vs output power

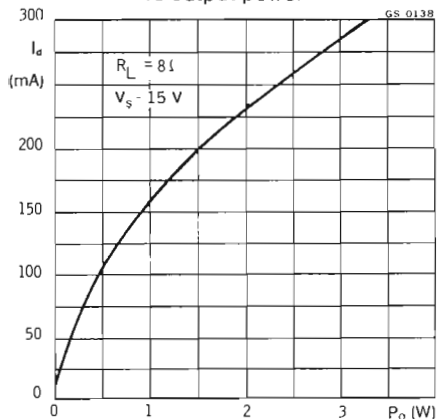


Fig. 17 - Typical quiescent drain current vs supply voltage

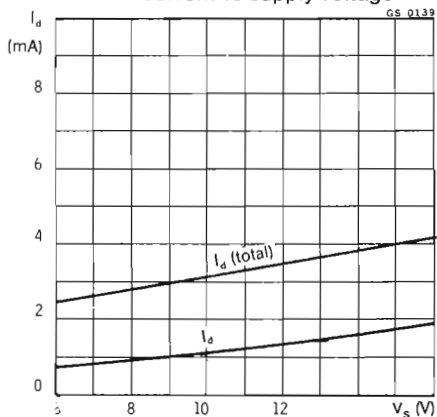


Fig. 18 - Typical total quiescent drain current vs ambient temperature

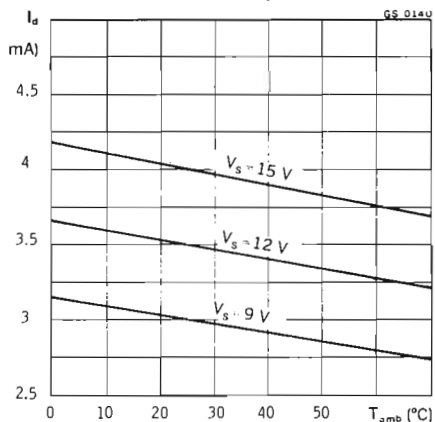
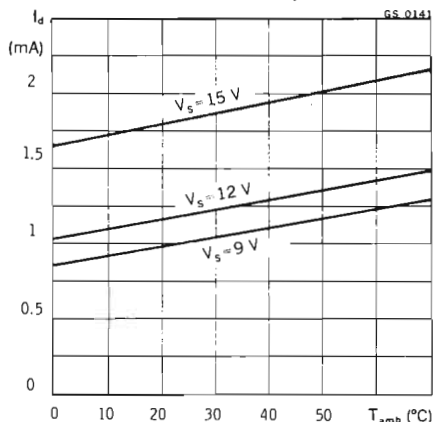


Fig. 19 - Typical quiescent drain current of output transistors vs ambient temperature



TAA 611C

Fig. 20 - Typical output voltage variation vs ambient temperature

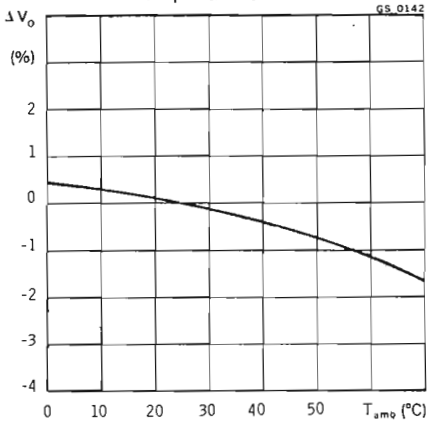
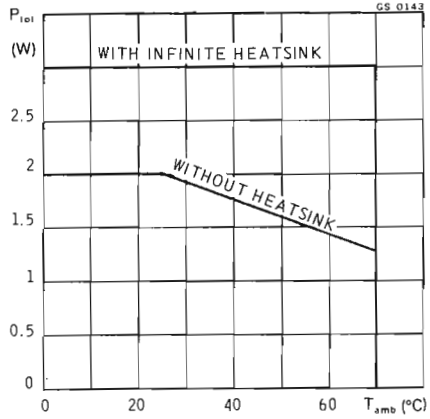


Fig. 21 - Power rating chart



TYPICAL APPLICATIONS

Fig. 22 - Audio amplifier for radio

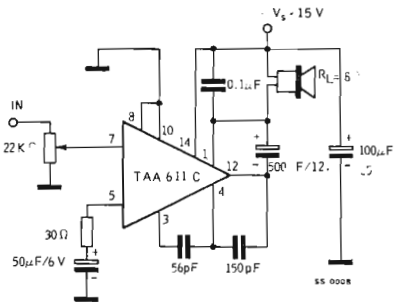
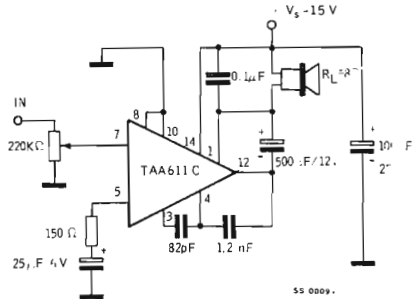


Fig. 23 - Audio amplifier for record-player



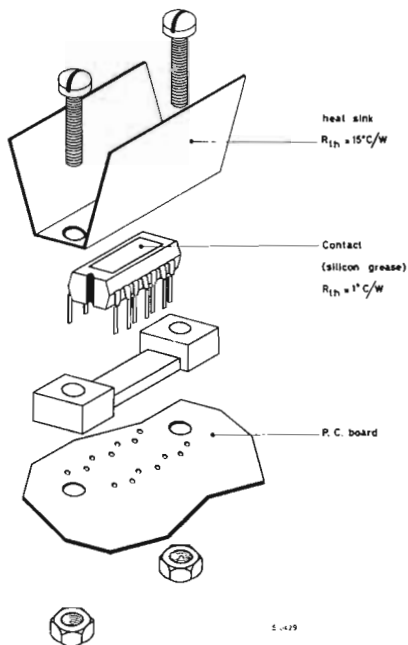
TAA 611C

MOUNTING INSTRUCTIONS

Heat-sinking with spacer.

Fig. 24 shows a method of mounting the TAA 611C with the spacer, satisfactory both mechanically and from the point of view of heat dissipation. Better thermal contact between package and heat-sink can be obtained by using a small quantity of silicon grease. For heat dissipation the desired thermal resistance is obtained by fixing the elements shown to a heat-sink of suitable dimensions.

Fig. 24



TAA 611C

MOUNTING INSTRUCTIONS (continued)

Heat-sinking with external bar.

Power dissipation can be achieved by means of an additional external heat-sink fixed with two screws (both packages) or by soldering the pins of the external bar to suitable copper areas on the p.c. board (TAA 611 C11).

A. In the former case, the thermal resistance case-ambient of the added heat-sink can be calculated as follows:

$$R_{th} = \frac{(T_{jmax} - T_{amb}) - P_{tot} \cdot R_{th j-case}}{P_{tot}}$$

where:

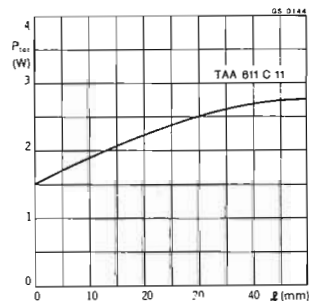
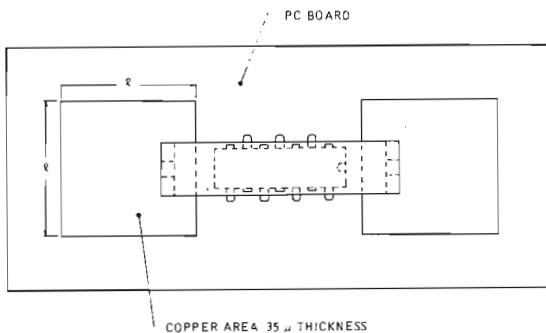
T_{jmax} = Max junction temperature

T_{amb} = Ambient temperature

P_{tot} = Power dissipation

$R_{th j-case}$ = Thermal resistance junction-case

B. If copper areas on the p.c. board are used (TAA 611 C11) the diagrams enclosed give the maximum power dissipation as a function of copper area, with copper thickness 35μ and ambient temperature 55°C .



TAA 621

LINEAR INTEGRATED CIRCUIT

AUDIO AMPLIFIER

- OUTPUT POWER 4 W (24 V - 16 Ω)
- SELF CENTERING BIAS
- LOW QUIESCENT OUTPUT CURRENT
- NO CROSS OVER DISTORTION
- HIGH EFFICIENCY

The TAA 621 is an integrated monolithic circuit in a 14-lead quad in-line plastic package with external heat-sink. It is particularly designed for use in television sets as audio amplifier.

Special features of the circuit include:

- Self centering bias for any supply voltage from 6 to 24 V.
- Direct coupled input, high input impedance and high supply voltage rejection ratio.
- Minimum number of external components.

The package has very low thermal resistance. To decrease the thermal resistance further, an external heat-sink can easily be mounted by means of ordinary hardware.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	27	V
V_i^*	Input voltage	0.5 to 27	V
I_o	Output peak current	1	A
P_{tot}	Power dissipation at $T_{amb} = 25^\circ\text{C}$ at $T_{case} = 70^\circ\text{C}$	2 4.5	W W
T_{stg}, T_j	Storage and junction temperature	-55 to 150	$^\circ\text{C}$

* For $V_s < 27\text{ V}$, $V_{i\max} = V_s$.

ORDERING NUMBERS:

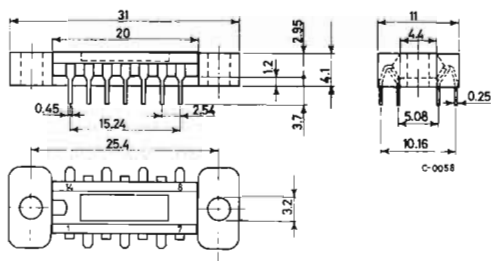
TAA 621 A72 (for quad in-line plastic package with spacer)

TAA 621 AX1 (for quad in-line plastic package with external bar)

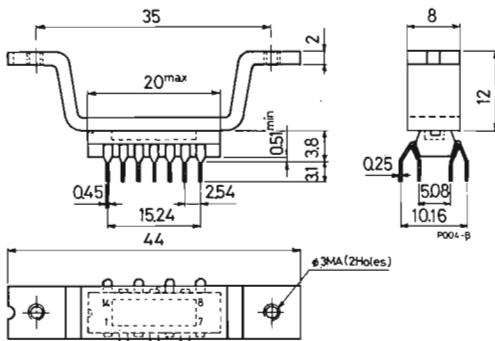
TAA 621 A11 (for quad in-line plastic package with inverted external bar)

TAA 621

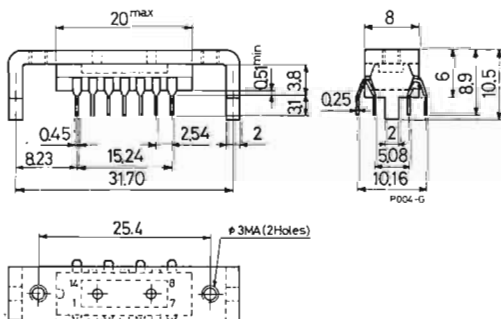
Quad in-line plastic package with spacer for TAA 621 A72 (see also "MOUNTING INSTRUCTIONS")



Quad in-line plastic package with external bar for TAA 621 AX1



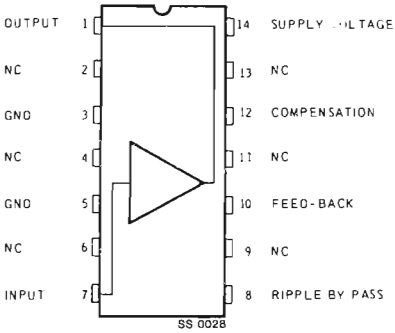
Quad in-line plastic package with inverted external bar for TAA 621 A11



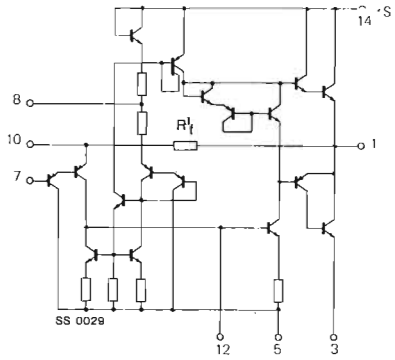
TAA 621

CONNECTION DIAGRAM

(top view)

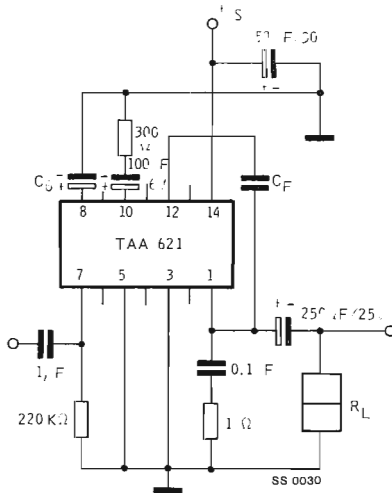


SCHEMATIC DIAGRAM



The heat-sink is connected to the substrate: (pin 5)

TEST CIRCUIT



TAA 621

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	17 °C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	63 °C/W

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_d	Total quiescent drain current $V_s = 18\text{ V}$ $V_s = 24\text{ V}$		6.2 7.5		mA mA
I_d	Quiescent drain current of output transistors $V_s = 18\text{ V}$ $V_s = 24\text{ V}$		2.5 3		mA mA
I_d	Drain current $d = 10\%$ $R_L = 16\ \Omega$ $P_o = 2.2\text{ W}$ $V_s = 18\text{ V}$ $P_o = 4\text{ W}$ $V_s = 24\text{ V}$		175 220		mA mA
I_b	Input bias current $V_s = 18\text{ V}$ $V_s = 24\text{ V}$		180 250		nA nA
P_o^*	Output power $d = 3\%$ $R_L = 16\ \Omega$ $V_s = 18\text{ V}$ $R_L = 16\ \Omega$ $V_s = 24\text{ V}$ $R_L = 16\ \Omega$ $d = 10\%$ $R_L = 16\ \Omega$ $V_s = 18\text{ V}$ $R_L = 16\ \Omega$ $V_s = 24\text{ V}$ $R_L = 16\ \Omega$		1.7 2.7 2.2 3	4	W W W W
R'_f	Internal feedback resistance (see schematic diagram)		15		k Ω
Z_i	Input impedance $V_s = 18\text{ V}$ $V_s = 24\text{ V}$		150 110		k Ω k Ω
d	Distortion $P_o = 50\text{ mW}$ $f = 1\text{ kHz}$ $R_L = 16\ \Omega$ $V_s = 18\text{ V}$ $V_s = 24\text{ V}$		0.1 0.1		% %
G_v	Voltage gain open loop $R_L = 16\ \Omega$ $V_s = 18\text{ V}$ $V_s = 24\text{ V}$		72 74		dB dB

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
SVR Supply voltage rejection	$R_L = 16 \Omega$				
	$f(\text{ripple}) = 100 \text{ Hz}$				
	$C_o = 100 \mu\text{f}$ (see application circuit diagrams)				
	$V_s = 18 \text{ V}$		52		dB
	$V_s = 24 \text{ V}$		52		dB
	$C_o = 50 \mu\text{F}$				
	$V_s = 18 \text{ V}$		46		dB
	$V_s = 24 \text{ V}$		46		dB

* External heat-sink not required except for the conditions $V_s = 24 \text{ V}$, $R_L = 16 \Omega$.

Fig. 1 - Typical output power vs supply voltage

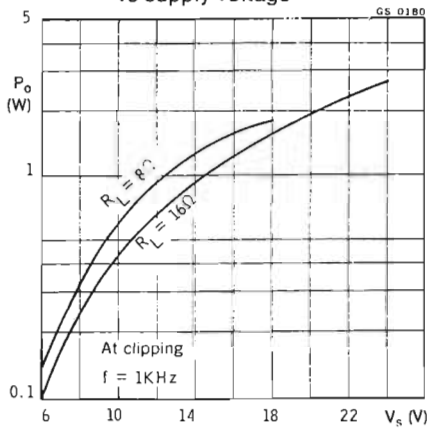
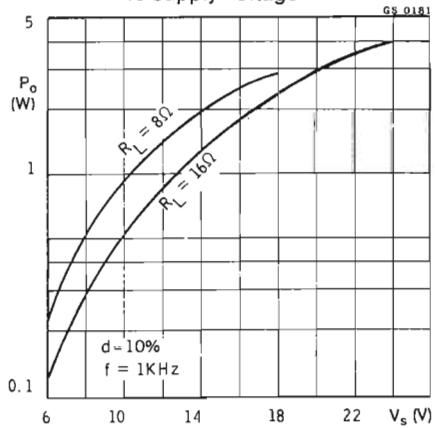


Fig. 2 - Typical output power vs supply voltage



TAA 621

Fig. 3 - Typical distortion vs output power

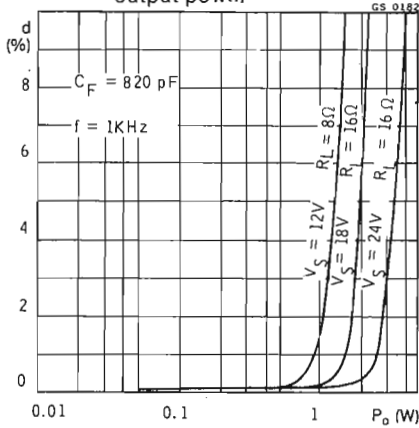


Fig. 4 - Typical relative frequency response

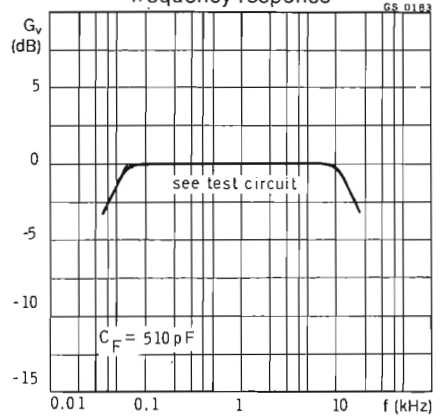


Fig. 5 - Typical relative frequency response

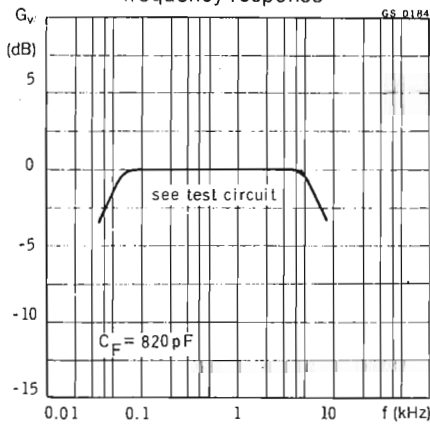


Fig. 6 - Typical open loop voltage gain vs frequency

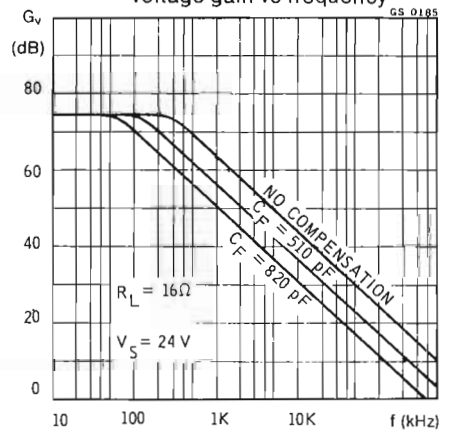


Fig. 7 - Typical output power vs input voltage

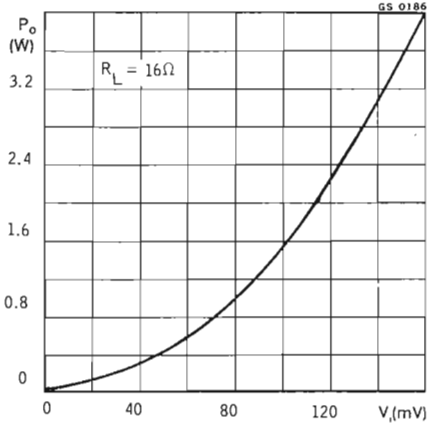


Fig. 8 - Typical output power vs input voltage

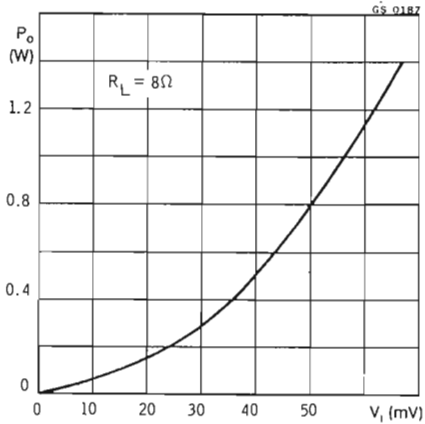


Fig. 9 - Typical power dissipation and efficiency vs output power

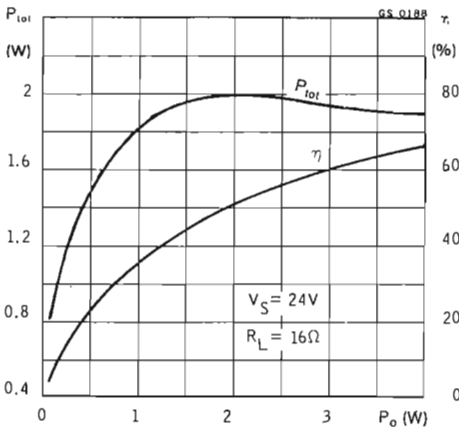
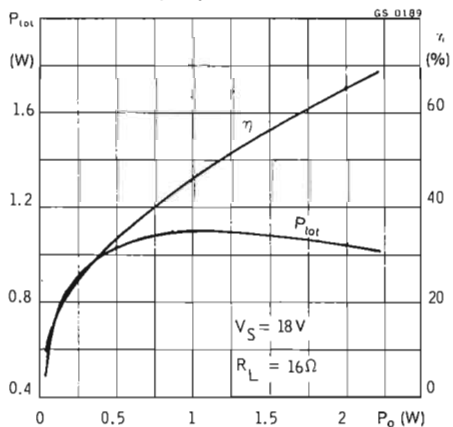


Fig. 10 - Typical power dissipation and efficiency vs output power



TAA 621

Fig. 11 - Typical power dissipation and efficiency vs output power

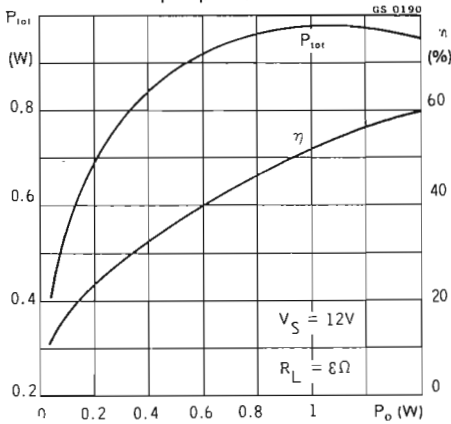


Fig. 12 - Typical drain current vs output power

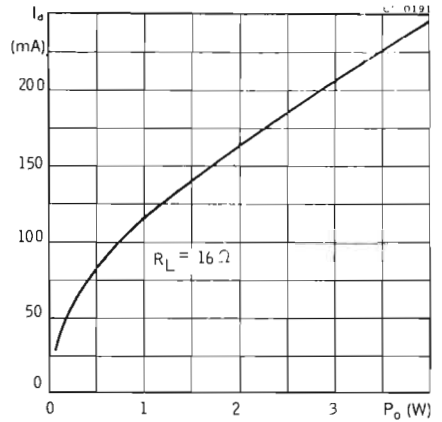


Fig. 13 - Typical drain current vs output power

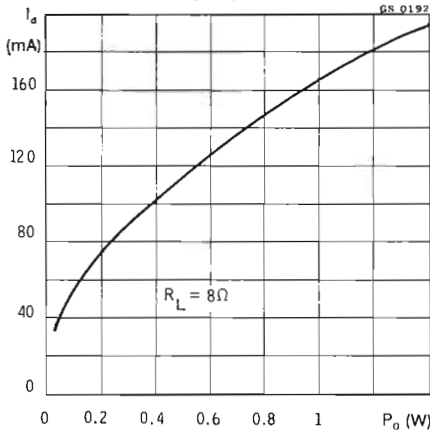
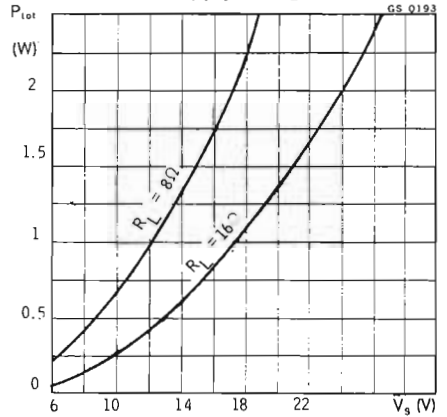


Fig. 14 - Maximum power dissipation vs supply voltage



TAA 621

Fig. 15 - Typical quiescent drain current vs supply voltage

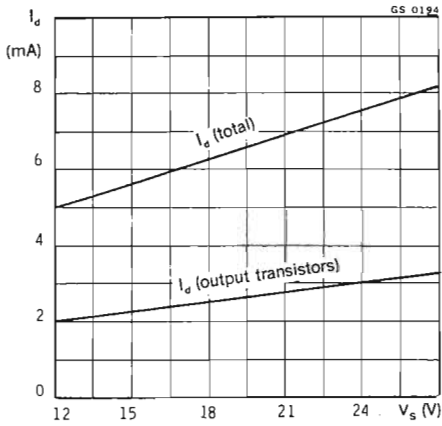


Fig. 16 - Typical total quiescent drain current vs ambient temperature

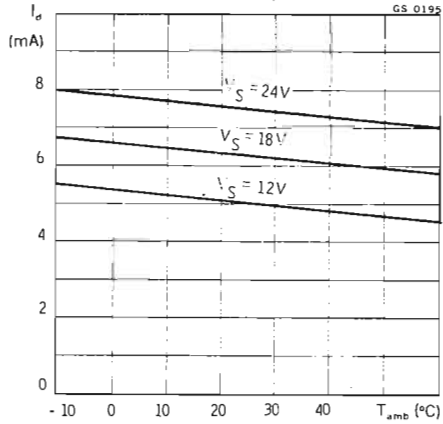


Fig. 17 - Typical quiescent drain current of output transistors vs ambient temperature

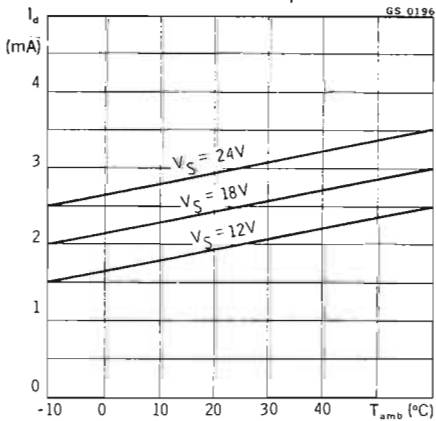
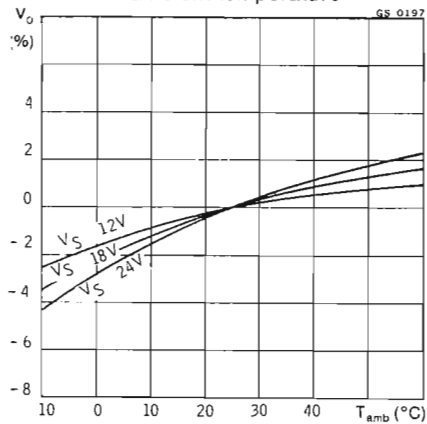
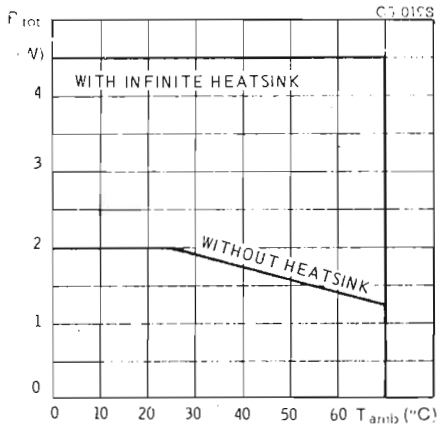


Fig. 18 - Typical relative DC output level vs ambient temperature



TAA 621

Fig. 19 - Power rating chart



TYPICAL APPLICATIONS

Fig. 20 - Record player

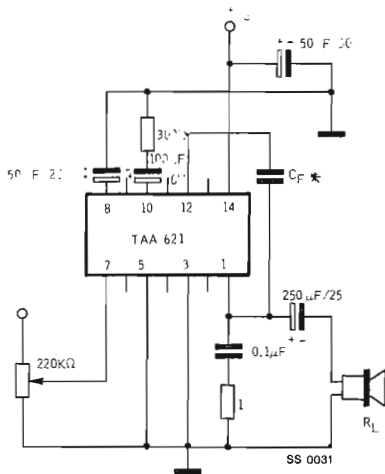
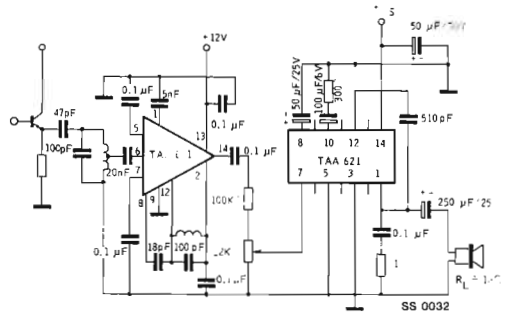


Fig. 21 - Complete TV sound section



* C_F see figs. 4 and 5.

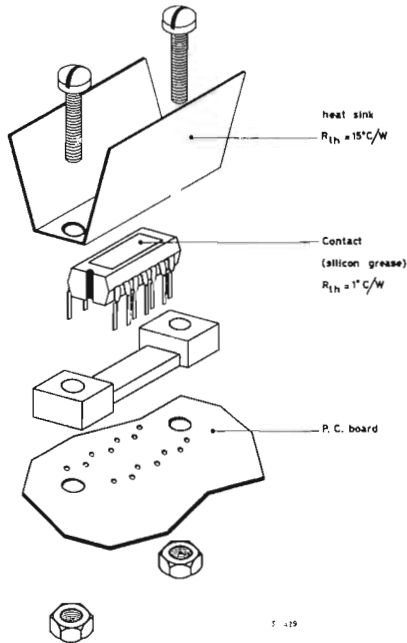
TAA 621

MOUNTING INSTRUCTIONS

Heat-sinking with spacer.

Fig. 22 shows a method of mounting the TAA 621 with the spacer, satisfactory both mechanically and from the point of view of heat dissipation. Better thermal contact between package and heat-sink can be obtained by using a small quantity of silicon grease. For heat dissipation the desired thermal resistance is obtained by fixing the elements shown to a heat-sink of suitable dimensions.

Fig. 22



TAA 621

MOUNTING INSTRUCTIONS (continued)

Heat-sinking with external bar

Power dissipation can be achieved by means of an additional external heat-sink fixed with two screws (both packages) or by soldering the pins of the external bar to suitable copper areas on the p.c. board (TAA 621 A11).

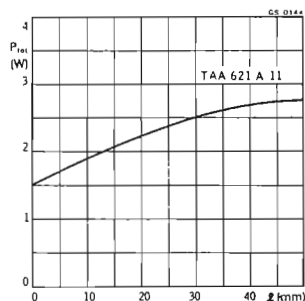
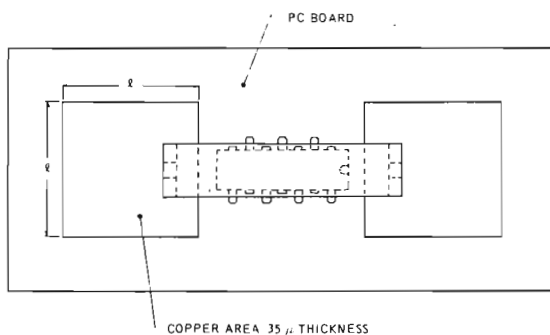
- A. In the former case, the thermal resistance case-ambient of the added heat-sink can be calculated as follows:

$$R_{th} = \frac{(T_{jmax} - T_{amb}) - P_{tot} \cdot R_{th\ j-case}}{P_{tot}}$$

where:

- T_{jmax} = Max junction temperature
- T_{amb} = Ambient temperature
- P_{tot} = Power dissipation
- $R_{th\ j-case}$ = Thermal resistance junction-case

- B. If copper areas on the p.c. board are used (TAA 621 A11) the diagrams enclosed give the maximum power dissipation as a function of copper area, with copper thickness $35\ \mu$ and ambient temperature $55\ ^\circ\text{C}$.



TAA 630S

LINEAR INTEGRATED CIRCUIT

PRELIMINARY DATA

SYNCHRONOUS DEMODULATOR FOR PAL COLOUR TV SETS

The TAA 630 S is a silicon monolithic integrated circuit in a 16-lead dual in-line plastic package. It incorporates the following functions:

- active synchronous demodulators for F (B-Y) and $\pm F$ (R-Y) signals
- matrix for G-Y signal [G-Y = -0.51 (R-Y) -0.19 (B-Y)]
- flip-flop
- PAL switch and colour killer.

It is intended for PAL colour television receivers employing colour difference output stages with clamping circuits.

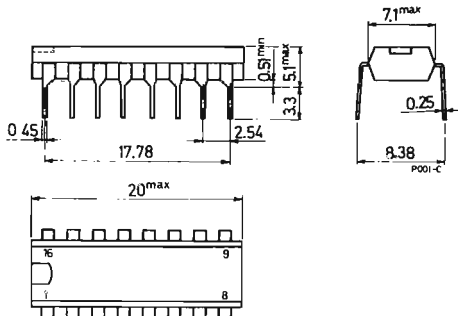
ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage (between pins 6 and 16 - see note)	13.2 V
V_1	Reverse identification input voltage	-5 V
I_1	Identification input current	1 mA
I_o	Output current (from pins 4, 5 and 7)	5 mA
P_{tot}	Total power dissipation: at $T_{amb} \leq 50^\circ\text{C}$ (see note)	550 mW
T_{stg}	Storage temperature	-20 to 125 °C
T_{op}	Operating temperature	-20 to 60 °C

NOTE: $V_s = 16\text{ V}$ and $P_{tot} = 800\text{ mW}$ (at $T_{amb} \leq 50^\circ\text{C}$) are permissible during warm up time of tubes in mixed sets.

MECHANICAL DATA

Dimensions in mm



TAA 630S

ELECTRICAL CHARACTERISTICS

(measured using the test circuit of fig. 3 at $T_{amb} = 25^{\circ}\text{C}$)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
-----------	-----------------	------	------	------	------

STATIC (DC) CHARACTERISTICS

I_1	Input current for identification circuit ON	$V_{10} \cong 0.9 \text{ V}$	80		μA
V_1	Input voltage for identification circuit ON		0.75		V
V_1	Input voltage for identification circuit OFF			0.4	V
V_4^*	DC voltage at (R-Y) output			see note	V
V_5^*	DC voltage at (G-Y) output			see note	V
V_7	DC voltage at (B-Y) output			7.3	V
V_{10}	Killer input voltage for colour ON			0.9	V
V_{10}	Killer input voltage for colour OFF			0.3	V

DYNAMIC CHARACTERISTICS

V_1	Peak to peak identification input voltage	$V_{10} \cong 0.9 \text{ V}$ $f = 7.8 \text{ kHz}$	4		V
V_3	Peak to peak flip-flop output voltage			2.5	V
V_4	R-Y output voltage swing	$V_{10} \cong 0.9 \text{ V}$ $f = 4.4 \text{ MHz}$ Linearity $m \cong 0.7$		3.2	V
V_5	G-Y output voltage swing			1.8	V
V_7	B-Y output voltage swing			4	V

TAA 630S

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_2^{**} R-Y reference input voltage	$V_{10} \geq 0.9 \text{ V}$ $f = 4.4 \text{ MHz}$		1		V
V_8^{**} B-Y reference input voltage			1		V
V_{14} Peak flip-flop input voltage	$V_{10} \geq 0.9 \text{ V}$ $f = 15.6 \text{ kHz}$	-2.5		-5	V
V_{15} Peak flip-flop input voltage		-2.5		-5	V
V_4/V_{13}^{***} R-Y demodulator gain	$V_{10} \geq 0.9 \text{ V}$ $f = 4.4 \text{ MHz}$ V_i (peak to peak) = 50 mV		7		—
$V_7/V_9 \cdot V_{13}/V_4$ B-Y demodulator gain to R-Y demodulator gain ratio			1.78		—
R_9 Parallel input resistance at pin 9	$V_{10} \geq 0.9 \text{ V}$ $f = 4.4 \text{ MHz}$ $V_i = 20 \text{ mV}$	800			Ω
C_9 Parallel input capacitance at pin 9				10	pF
R_{13} Parallel input resistance at pin 13		800			Ω
C_{13} Parallel input capacitance at pin 13				10	pF
$ Z_4 $ R-Y output impedance	$V_{10} \geq 0.9 \text{ V}$			100	Ω
$ Z_5 $ G-Y output impedance				100	Ω
$ Z_7 $ B-Y output impedance				100	Ω
$ Z_2 $ Parallel input impedance at pin 2	$V_{10} \geq 0.9 \text{ V}$ $f = 4.4 \text{ MHz}$ $V_i = 400 \text{ mV}$		900		Ω
$ Z_8 $ Parallel input impedance at pin 8			900		Ω

NOTES: * Adjustable to the same level of V_7 by variable resistors, or by variable voltages $\leq 1.2 \text{ V}$, connected between pins 11 and 16 for V_4 and between pins 12 and 16 for V_5 .

** Maximum permissible range : 0.5 to 2 V (peak to peak).

*** Peak to peak output voltage to peak to peak input voltage ratio.

TAA 630S

Fig. 1 - Schematic diagram

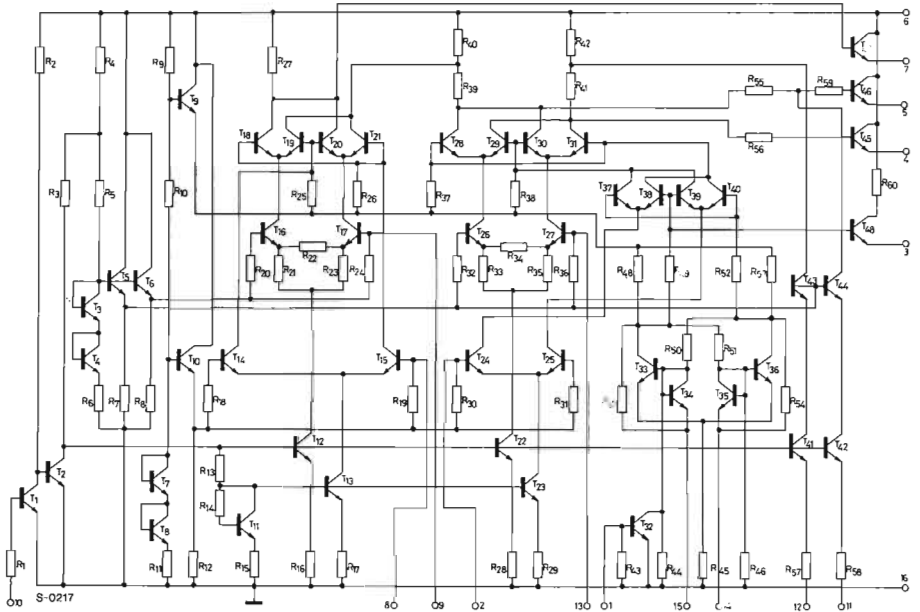
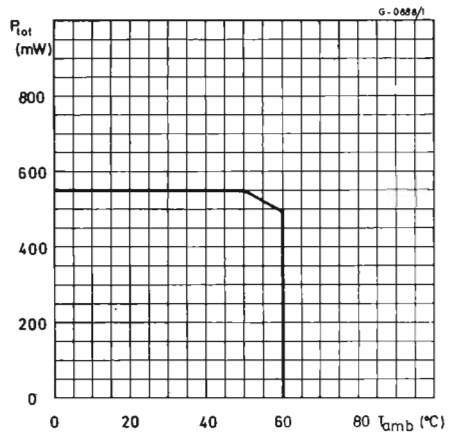


Fig. 2 - Power rating chart



TAA 630S

Fig. 3 - Test circuit

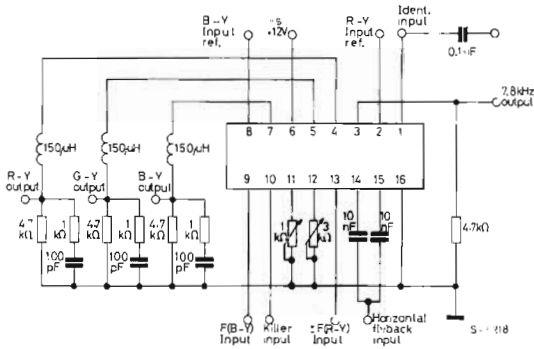
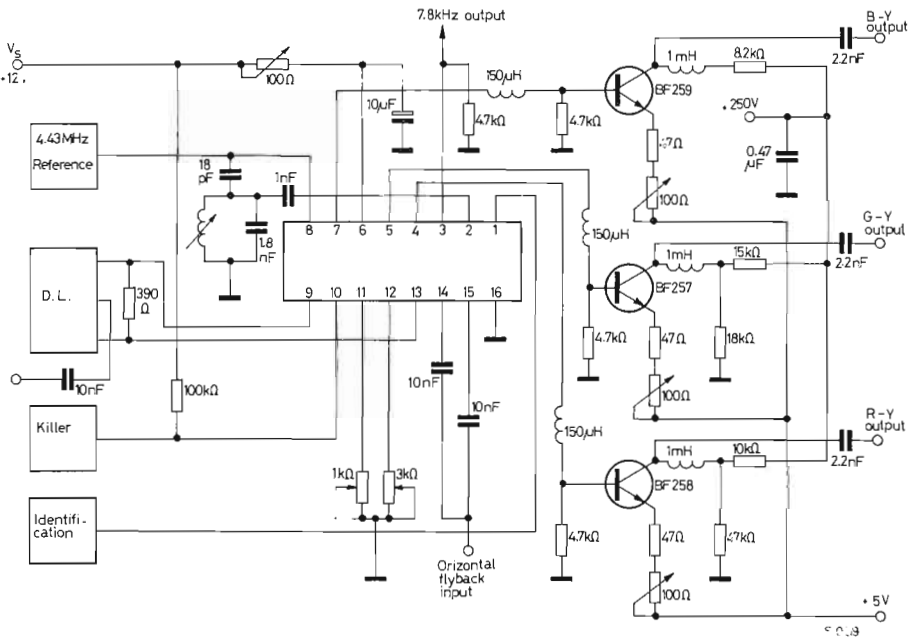


Fig. 4 - Typical application circuit



TAA 661

LINEAR INTEGRATED CIRCUIT

FM IF AMPLIFIER-LIMITER AND DETECTOR

- HIGH GAIN
- FREQUENCY RANGE 5 kHz to 60 MHz
- THRESHOLD LIMITING VOLTAGE 100 μ V (5.5 MHz)
- COINCIDENCE GATE DETECTOR
- AUDIO OUTPUT VOLTAGE 1.4 Vrms (d = 1%)

The TAA 661 is a monolithic integrated circuit in a 14-lead quad in-line plastic package or in a Jedec TO-100 metal case. Particularly designed for use in TV sound IF or FM IF amplifiers, it includes: a limiter amplifier, a coincidence detector and a voltage regulator. By using the TAA 661 the ratio detector transformer is eliminated and the audio signal is capable of driving an output amplifier directly. Detector alignment is obtained by adjusting a single coil which provides the quadrature signal to the coincidence gate detector.

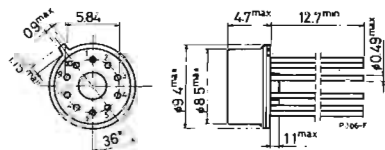
ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	15 V
P_{tot}	Power dissipation at $T_{amb} \leq 70^\circ\text{C}$	350 mW
	for TAA 661 A55	500 mW
	for TAA 661 BX2	
T_{stg}	Storage temperature	-25 to 125 $^\circ\text{C}$
T_{op}	Operating temperature	0 to 70 $^\circ\text{C}$

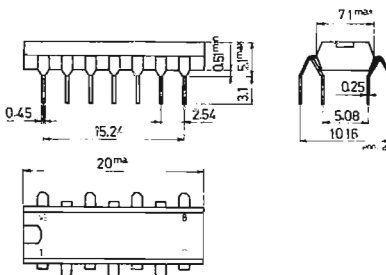
ORDERING NUMBERS: TAA 661 A55 (for TO-100 metal case)
TAA 661 BX2 (for 14-lead quad in-line plastic package)

MECHANICAL DATA

Dimensions in mm



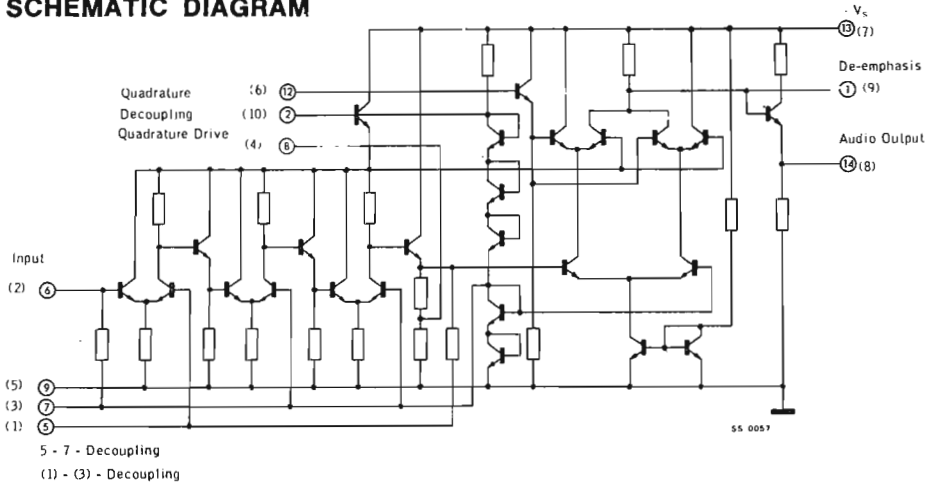
TAA 661 A55



TAA 661 BX2

TAA 661

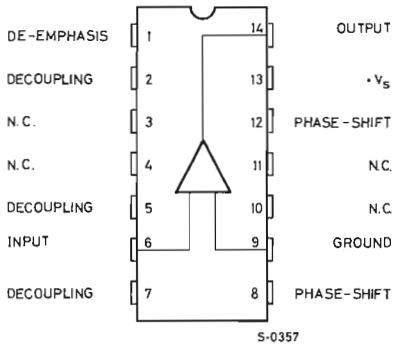
SCHEMATIC DIAGRAM



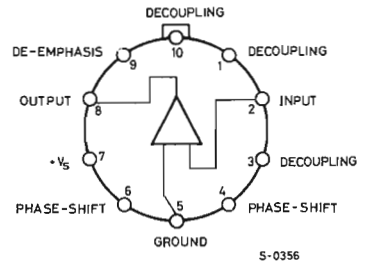
NOTE: the number in brackets refers to the TO-100 package.

CONNECTION DIAGRAMS (top views)

For TAA 661 BX2

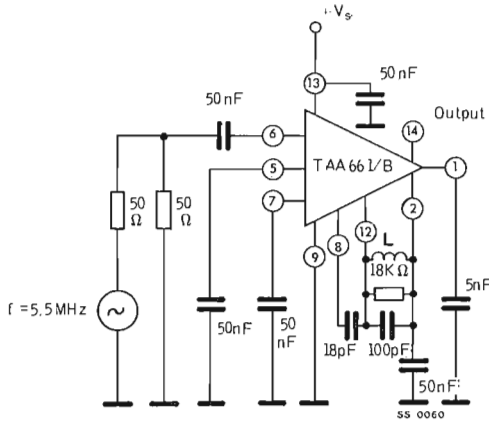


For TAA 661 A55



TAA 661

TEST CIRCUIT



L = 35 turns of 0.16 mm nylon covered copper wire.

ELECTRICAL CHARACTERISTICS (T_{amb} = 25 °C unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	
I _d Quiescent drain current	V _s = 6 V V _s = 9 V V _s = 12 V	9	11	14	17 20	mA
V _{i(threshold)} Input limiting voltage	f = 5.5 MHz f = 10.7 MHz		100	230	μV μV	
V _o Recovered output voltage	V _i = 10 mV f = 5.5 MHz f _m = 1 kHz Δf = ±50 kHz V _s = 6 V V _s = 9 V V _s = 12 V		0.5 0.75 1.4		V _{rms} V _{rms} V _{rms}	
d Distortion	V _s = 12 V V _i = 10 mV f = 5.5 MHz f _m = 1 kHz Δf = ±25 kHz		1		%	

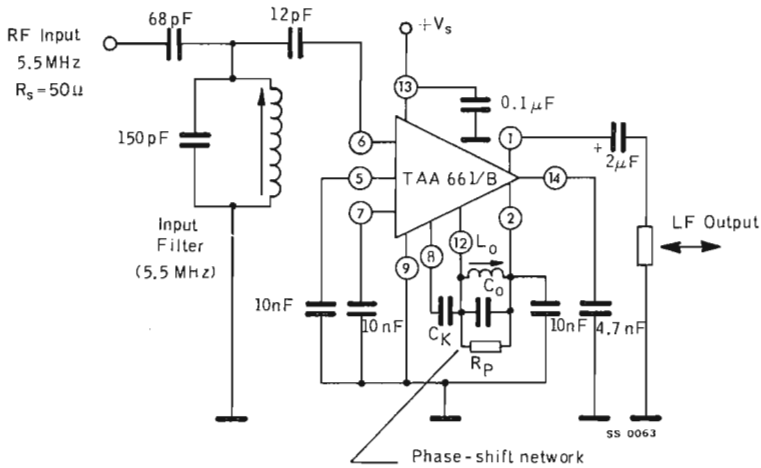
TAA 661

ELECTRICAL CHARACTERISTICS (continued)

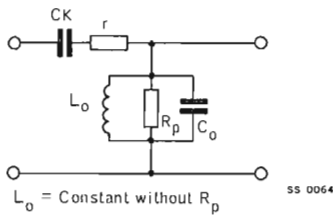
Parameter		Test conditions	Min.	Typ.	Max.	Unit
AMR	Amplitude modulation rejection	$V_s = 12\text{ V}$ $V_i = 10\text{ mV}$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 50\text{ kHz}$ $m = 0.3$		45		dB
R_i	Input resistance	$f = 5.5\text{ MHz}$ $f = 10.7\text{ MHz}$		2.5 2		$k\Omega$ $k\Omega$
C_i	Input capacitance	$V_s = 9\text{ V}$ $f = 5.5\text{ MHz}$		2.5		pF
Z_o	Output impedance	$V_s = 6\text{ V}$ $V_s = 9\text{ V}$ $V_s = 12\text{ V}$		200 150 100		Ω Ω Ω
R_L	Min. load impedance without clipping	$V_s = 6\text{ V}$ $V_s = 9\text{ V}$ $V_s = 12\text{ V}$		10 4 2		$k\Omega$ $k\Omega$ $k\Omega$
R_{5-6}	Resistance between pins 5 and 6 of the TAA 661 A55	$f = 5.5\text{ MHz}$ $V_s = 6\text{ V}$ $V_s = 9\text{ V}$ $V_s = 12\text{ V}$		50 50 50		$k\Omega$ $k\Omega$ $k\Omega$
C_{5-6}	Capacitance between pins 5 and 6 of the TAA 661 A55	$f = 5.5\text{ MHz}$		3		pF

TAA 661

TAA 661 AS TV SOUND IF AMPLIFIER (outputs referred to the TAA 661 BX2)



PHASE SHIFT NETWORK



	a	b	c	d	e	f
Co/pF	120	100	56	33	15	—

TAA 661

Fig. 1 - Typical distortion

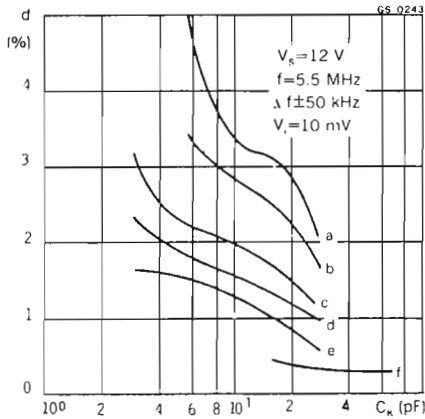


Fig. 2 - Typical amplitude modulation rejection

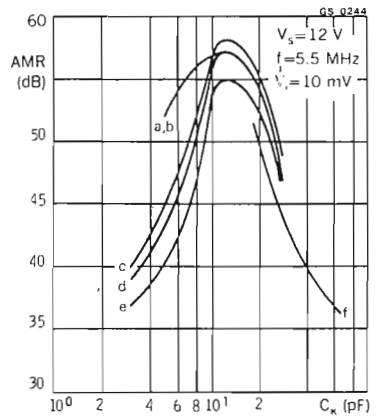
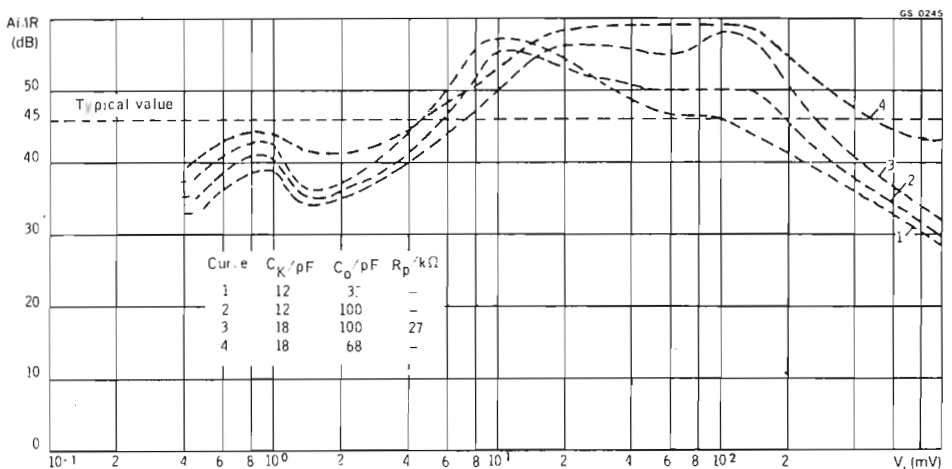


Fig. 3 - Typical amplitude modulation rejection vs input voltage



TAA 661

Fig. 4 - Typical recovered output voltage

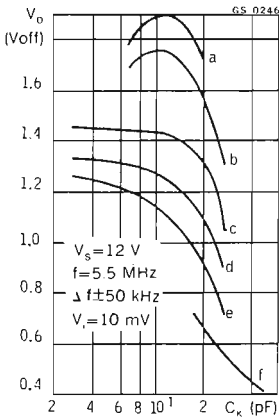
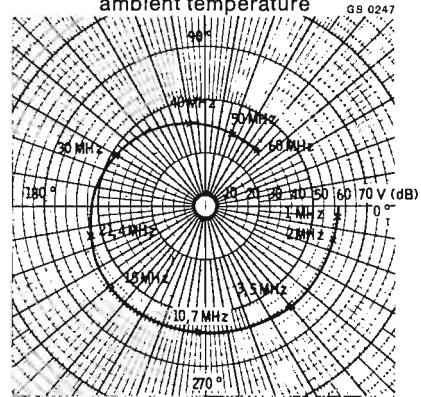
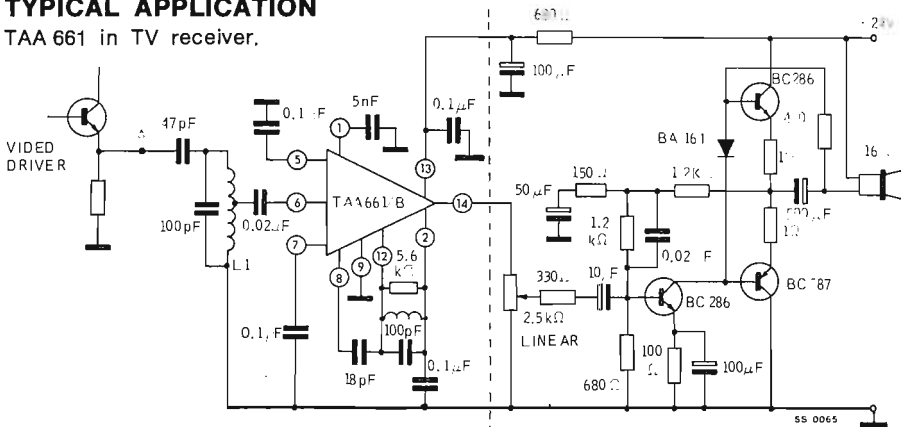


Fig. 5 - Phase response of the TAA 661 wide band amplifier measured at 25 °C ambient temperature



TYPICAL APPLICATION

TAA 661 in TV receiver.



Notes:

- Pin numbers shown are for the TAA 661 BX2.
- $L_1 = 24$ turns of 0.16 mm nylon covered copper wired with tapping at turn 12 from ground.
- $L_2 = 35$ turns of 0.16 mm nylon covered copper wired.
- Neosid former K4/21.5/0.5 - Neosid core GW4 x 0.5 x 10FE10($Q_o = 80$).

TAA 691

LINEAR INTEGRATED CIRCUIT

WIDE-BAND AMPLIFIER, FM DETECTOR, AUDIO PREAMPLIFIER/DRIVER

The TAA 691 provides, in a single monolithic silicon chip, a major subsystem for the sound section of TV receivers in a 14-lead quad in-line plastic package.

As shown in the schematic diagram the TAA 691 contains a multistage wide-band IF amplifier/limiter section, an FM-detector stage, a Zener-diode-regulated power-supply section and an audio-amplifier section specifically designed to drive directly any type of valve or transistor output stage.

In FM receivers, the TAA 691 can be used to provide IF amplification and limiting, FM detection and AF preamplification.

In the TAA 691, the demodulation is effected by a single tuning discriminator coils as well as a ratio detector.

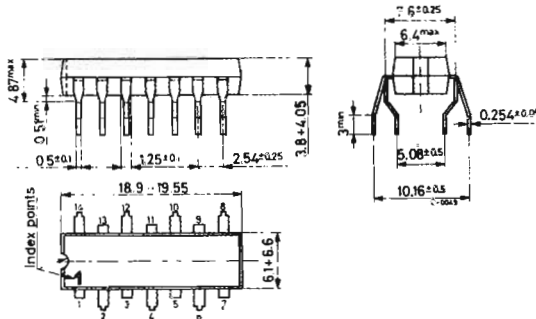
The TAA 691 provides exceptional versatility of circuit design because the IF amplifier/limiter section, FM-detector section and audio-preamplifier/driver section can be used independently of each other.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage (at pin 6)	20	V
V_i	Input-signal voltage (between terminals 1 and 3)	± 3	V
I_s	Supply current (at pin 14)	50	mA
I_o	Output current (from pin 5)	80	mA
P_{tot}	Total power dissipation at $T_{amb} \leq 25^\circ\text{C}$	850	mW
T_{stg}	Storage temperature	-25 to 150	$^\circ\text{C}$
T_{op}	Operating temperature	0 to 85	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm



TAA 691

THERMAL DATA

$R_{th j-amb}$	Thermal resistance junction-ambient	typ.	150	$^{\circ}\text{C}/\text{W}$
----------------	-------------------------------------	------	-----	-----------------------------

ELECTRICAL CHARACTERISTICS* ($T_{amb} = 25^{\circ}\text{C}$)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
I_{11}	Bias current of third amplifier	0.25	0.63	1	mA	3
I_{14}	Supply current	$V_s = 6.2\text{ V}$ (applied direct. to pin 14)			mA	3
V_{14}	Internal reference voltage	6.9	7.4	8.1	V	3
$V_{i(lim)}$	Input limiting voltage (pin 13)	$f = 5.5\text{ MHz}$			μV	5-6
V_o	Recovered audio voltage (pin 13)	$V_i = 10\text{ mV}$ $R_L = 50\text{ k}\Omega$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 50\text{ kHz}$			mV	5
V_o	Audio output voltage (pin 5)	$V_i = 10\text{ mV}$ $R_L = 1\text{ k}\Omega$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 50\text{ kHz}$			V	10
d	Distortion (pin 13)	$V_i = 100\text{ mV}$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 50\text{ kHz}$			%	10
V_i	Input voltage (pin 7)	$P_o = 1.5\text{ W}$ $f = 1\text{ kHz}$			mV	2

* DC supply voltage, V_s , of +16 V applied to terminal 14 through a resistance of 100Ω , unless otherwise indicated.

TAA 691

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
d	Distortion (on R_L) $V_s = 10.8\text{ V}$ $P_o = 1\text{ W}$ $f = 1\text{ kHz}$ Input at pin 7		1		%	2
R_i	Input resistance (pin 1) $f = 5.5\text{ MHz}$		11		$k\Omega$	4
R_i	Input resistance (pin 7) $f = 1\text{ kHz}$		100		$k\Omega$	—
R_o	Output resistance (pin 11) $f = 5.5\text{ MHz}$		100		$k\Omega$	—
R_o	Output resistance (pin 5) $f = 1\text{ kHz}$		250		Ω	—
R_o	Output resistance (pin 13) $f = 1\text{ kHz}$		10		$k\Omega$	—
C_i	Input capacitance (pin 1) $f = 5.5\text{ MHz}$		5		pF	4
C_o	Output capacitance (pin 11) $f = 5.5\text{ MHz}$		4		pF	—
G_v	Voltage gain $f = 5.5\text{ MHz}$		67		dB	8
P_{tot}	Total power dissipation	245	265	280	mW	3
AMR	Amplitude modulation rejection $f = 5.5\text{ MHz}$	35	48		dB	5-7

TAA 691

SCHEMATIC DIAGRAM

Fig. 1

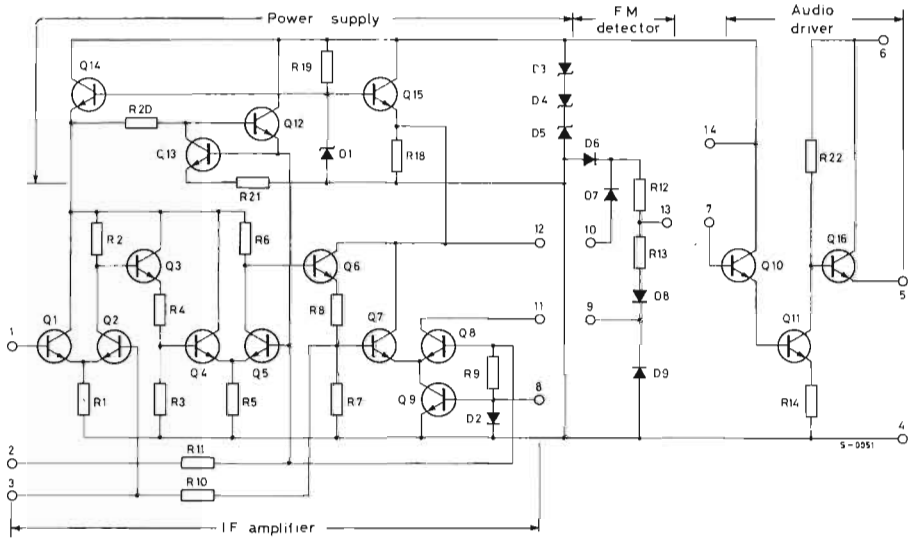
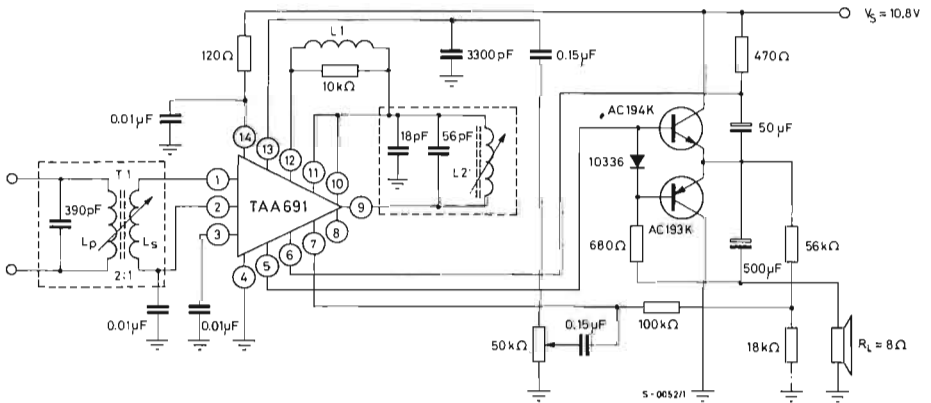


Fig. 2 - Typical circuit utilizing the TAA 691 and SGS-ATES AC 193K and AC 194K transistors



$T_1 = 5.5 \text{ MHz TRANSFORMER:}$

$L_p = 5.5 \mu\text{H}; Q_0 = 80; 19 \text{ turns } \varnothing 0.15 \text{ mm silk-covered copper wire.}$

$L_s = 9 \text{ turns } \varnothing 0.15 \text{ mm.}$

TAA 691

$L_1 = 31 \mu\text{H}$ (150 turns \varnothing 0.04 mm wound on 1/2 W - 5.6 k Ω resistor).

$L_2 = 18 \mu\text{H}$; $Q_o = 75$; (double-layer winding, 45 turns \varnothing 0.08 with powdered-iron core).

Fig. 3 - Test setup for measurement of total device dissipation, quiescent current into pin 11 and drain current from 6.2 Volt

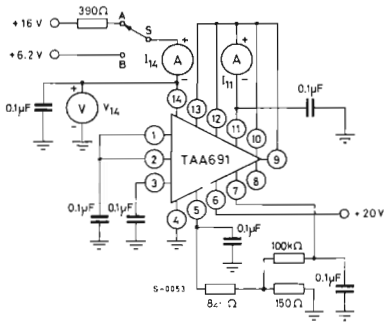


Fig. 4 - Test setup for measurement of input-impedance

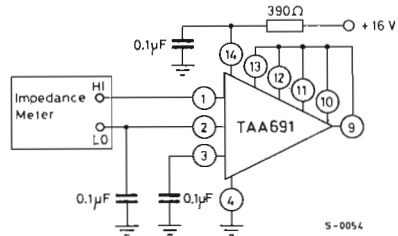
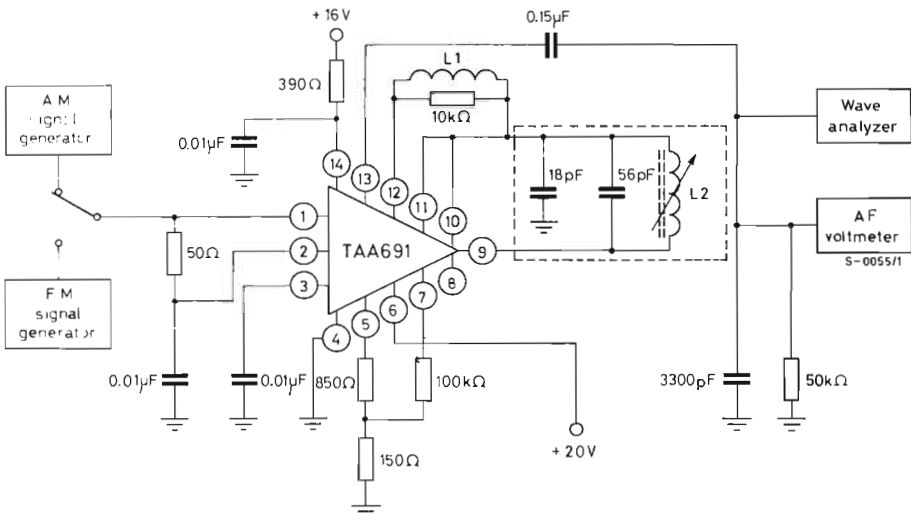


Fig. 5 - Test setup for measurement of AM rejection, input limiting voltage, FM-detector output voltage and distortion



TAA 691

Fig. 6 - Typical FM-detector output voltage versus input voltage

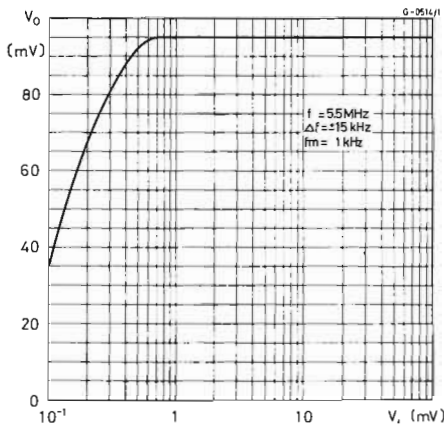


Fig. 7 - Typical amplitude-modulation rejection versus input voltage

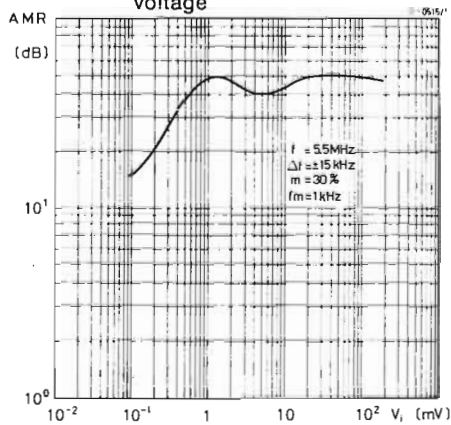


Fig. 8 - Test setup for measurement of IF amplifier voltage gain

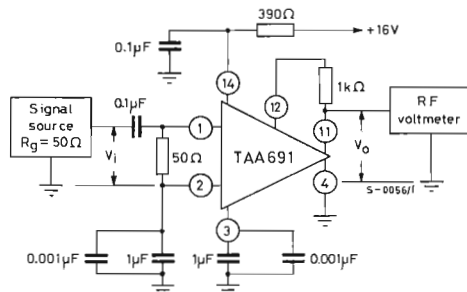
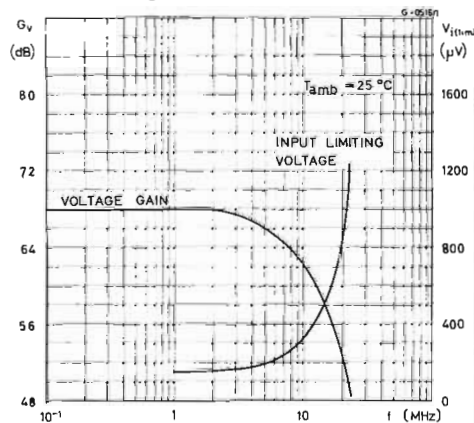
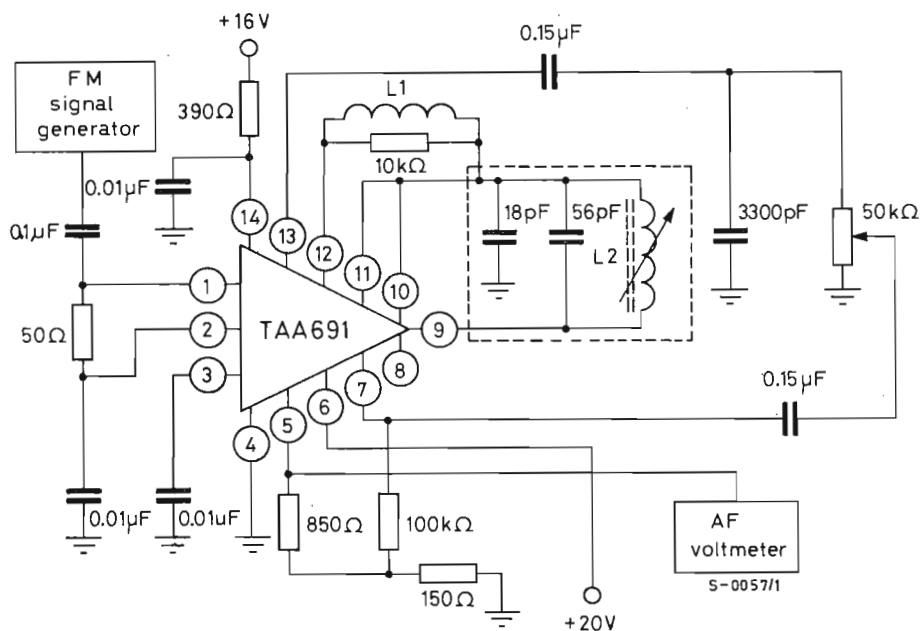


Fig. 9 - Typical IF amplifier voltage gain and input limiting voltage characteristics



TAA 691

Fig. 10 - Test setup for measurement of audio output voltage



TBA 231

LINEAR INTEGRATED CIRCUIT

DUAL LOW NOISE OPERATIONAL AMPLIFIER

- SINGLE or DUAL SUPPLY OPERATION
- LOW NOISE FIGURE
- HIGH GAIN
- LARGE INPUT VOLTAGE RANGE
- EXCELLENT GAIN STABILITY VERSUS SUPPLY VOLTAGE
- NO LATCH UP
- OUTPUT SHORT CIRCUIT PROTECTED

The TBA 231 is a monolithic integrated dual operational amplifier in a 14-lead dual in-line plastic package.

These low-noise, high-gain amplifiers show extremely stable operating characteristics over a wide range of supply voltage and temperatures.

The device is intended for a variety of applications requiring two high performance operational amplifiers, such as phono and tape stereo preamplifier, TV remote control receiver, etc.

ABSOLUTE MAXIMUM RATINGS

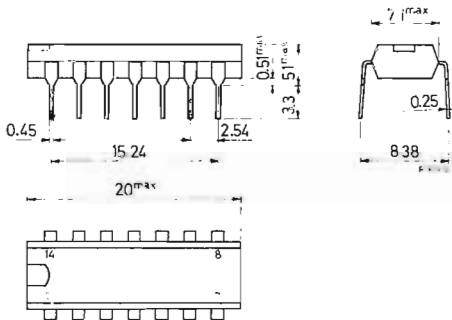
V_s	Supply voltage	± 18 V
	Differential input voltage	± 5 V
	Common mode input voltage	± 15 V
P_{tot}	Power dissipation at $T_{amb} \leq 60^\circ\text{C}$	500 mW
T_{stg}	Storage temperature	-55 to 125 °C
T_{op}	Operating temperature	0 to 70 °C

* For $V_s \leq \pm 15$ V, $V_i \text{ max} = V_s$

ORDERING NUMBER: TBA 231

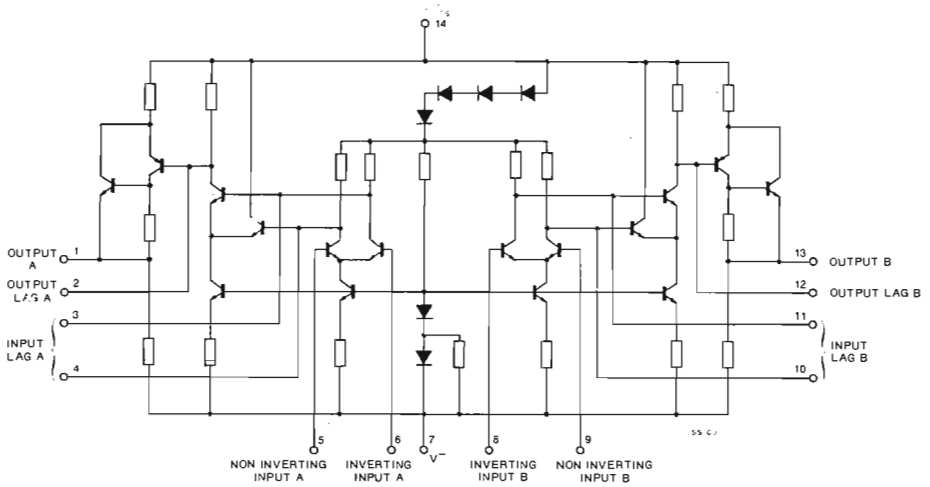
MECHANICAL DATA

Dimensions in mm



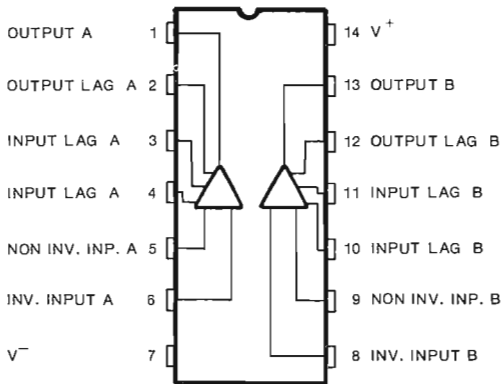
TBA 231

SCHEMATIC DIAGRAM



CONNECTION DIAGRAM

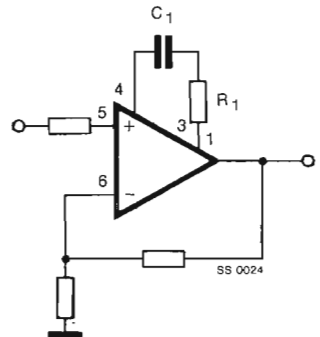
(top view)



SS 0022

TEST CIRCUIT

Frequency response



SS 0024

TBA 231

ELECTRICAL CHARACTERISTICS

($T_{amb} = 25\text{ }^{\circ}\text{C}$, $R_L = 50\text{ k}\Omega$ to pin 7 unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
-----------	-----------------	------	------	------	------

$V_s = \pm 15\text{ V}$

I_d	Quiescent drain current	$V_o = 0$	9	14	mA
$ V_{BE1} - V_{BE2} $	Input offset voltage	$R_s = 200\ \Omega$	1	6	mV
$ I_{B1} - I_{B2} $	Input offset current		50	1000	nA
I_b	Input bias current		250	2000	nA
	Common mode input voltage range		± 10	± 11	V
R_i	Input resistance	$f = 1\text{ kHz}$	37	150	$\text{k}\Omega$
G_v	Voltage gain	$V_o = \pm 5\text{ V}$	6500	20.000	—
V_o	Positive output voltage swing		+12	+13	V
V_o	Negative output voltage swing		-14	-15	V
R_o	Output resistance	$f = 1\text{ kHz}$	5		$\text{k}\Omega$
CMRR	Common mode rejection ratio	$R_s = 200\ \Omega$	70	90	dB
SVR	Supply voltage rejection	$R_s = 200\ \Omega$	50		$\mu\text{V}/\text{V}$
SR	Slew rate	Unity gain $C_1 = 0.1\ \mu\text{F}$ $R_1 = 4.7\ \Omega$ see frequency response test circuit	1		$\text{V}/\mu\text{s}$
	Channel separation	$R_s = 10\text{ k}\Omega$ $f = 10\text{ kHz}$	140		dB
NF	Noise figure	$R_s = 10\text{ k}\Omega$ $B = 10\text{ Hz to } 10\text{ kHz}$	1.5		dB

$V_s = \pm 4\text{ V}$

I_d	Quiescent drain current	$V_o = 0$	2.5		mA
$ V_{BE1} - V_{BE2} $	Input offset voltage	$R_s = 200\ \Omega$	1	6	mV

TBA 231

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
$ I_{B1}-I_{B2} $ Input offset current			50	1000	nA
I_b Input bias current			250		nA
G_v Voltage gain	$V_o = \pm 1 V$	2500	15.000		—
V_o Positive output voltage swing		+2.5	+2.8		V
V_o Negative output voltage swing		-3.6	-4		V

Fig. 1 - Power rating chart

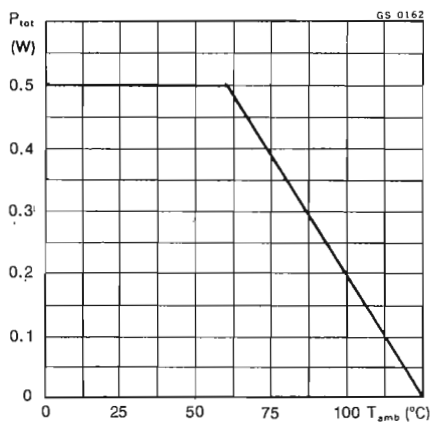
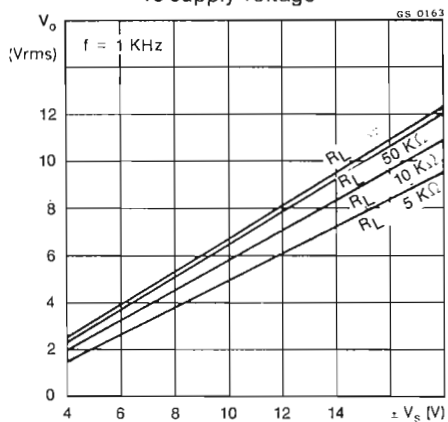


Fig. 2 - Typical output capability vs supply voltage



TBA 231

Fig. 3 - Typical quiescent drain current vs supply voltage

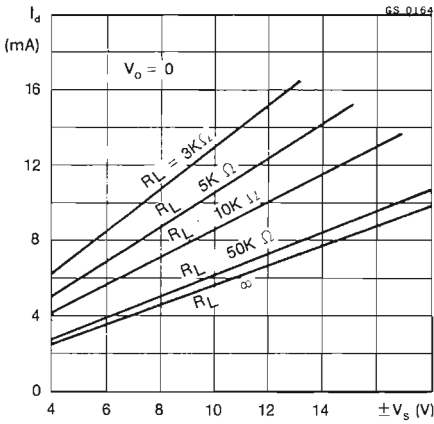


Fig. 4 - Typical open loop voltage gain vs supply voltage

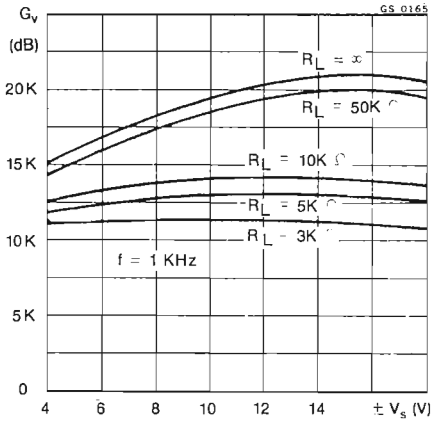


Fig. 5 - Typical open loop frequency response using recommended compensation networks

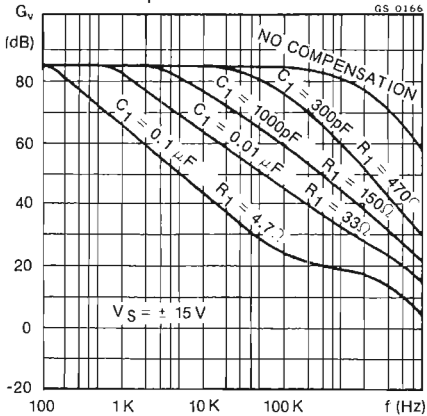
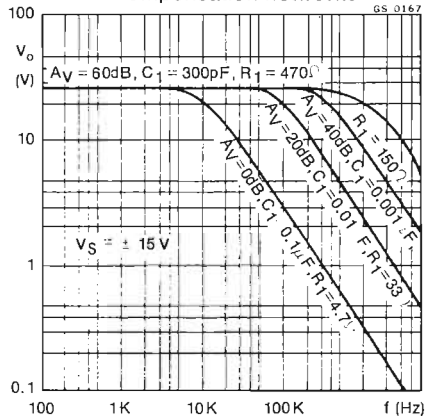


Fig. 6 - Output voltage swing vs frequency for various compensation networks



TBA 231

Fig. 7 - Typical input noise voltage vs frequency

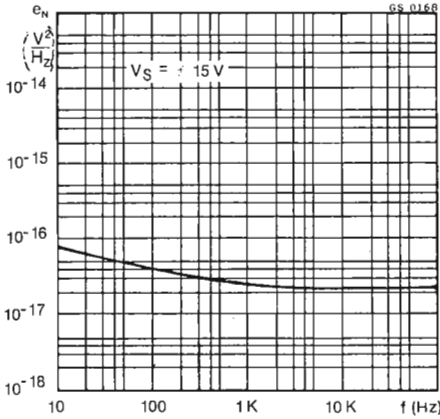


Fig. 8 - Typical input noise current vs frequency

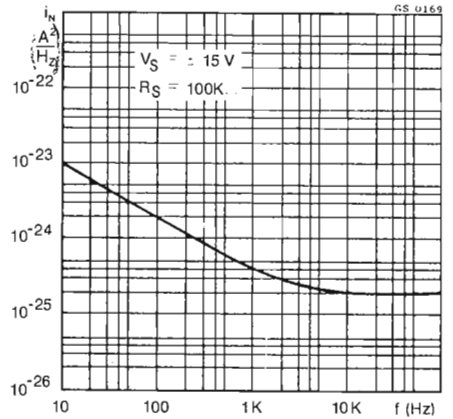


Fig. 9 - Typical closed loop gain vs frequency

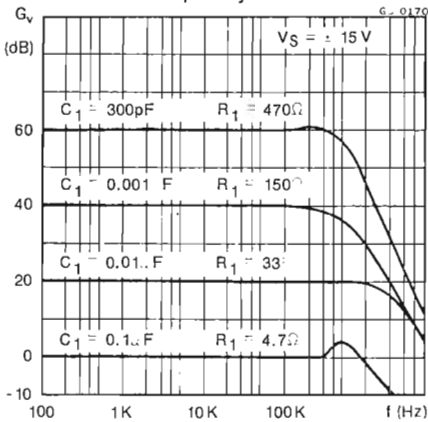
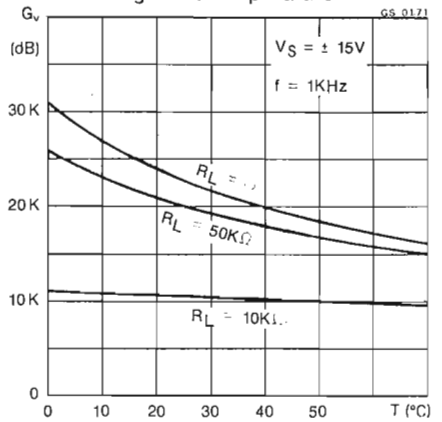


Fig. 10 - Typical open loop voltage gain vs temperature



TBA 261

LINEAR INTEGRATED CIRCUIT

FM IF AMPLIFIER-LIMITER, DETECTOR, DC VOLUME CONTROL

- AUDIO OUTPUT VOLTAGE $0.9 V_{rms}$
- REMOTE CONTROL RANGE 70 dB
- INPUT LIMITING VOLTAGE $100 \mu V$

The TBA 261 is a monolithic integrated circuit in a 14-lead quad in-line or dual in-line plastic package. It is particularly designed for use in TV sound IF or FM IF amplifiers; it includes: a three stages FM limiter amplifier, a gated coincidence detector and a remote control stage.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	15 V
P_{tot}	Power dissipation at $T_{amb} \leq 70^\circ C$	500 mW
T_{stg}	Storage temperature	-55 to 125 °C
T_{op}	Operating temperature	0 to 70 °C

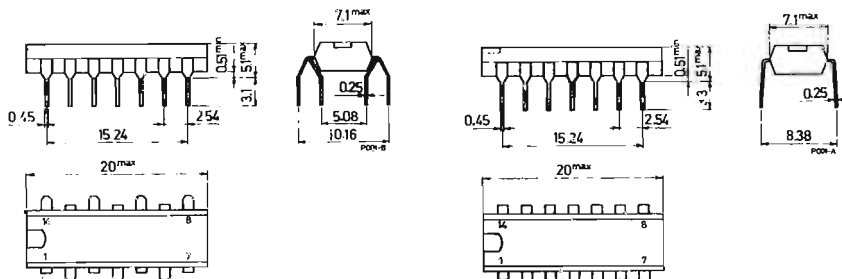
ORDERING NUMBERS:

TBA 261 AX2 (for 14-lead quad in-line plastic package)

TBA 261 AX7 (for 14-lead dual in-line plastic package)

MECHANICAL DATA

Dimensions in mm



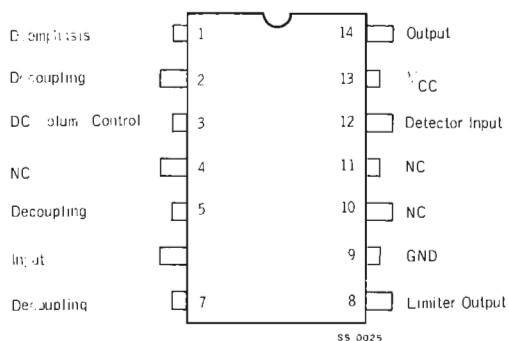
TBA 261 AX2

TBA 261 AX7

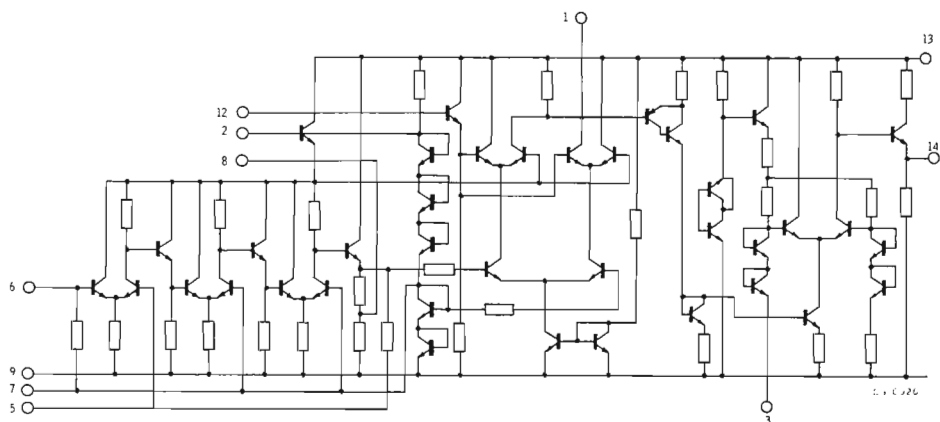
TBA 261

CONNECTION DIAGRAM

(top view)



SCHEMATIC DIAGRAM



TBA 261

ELECTRICAL CHARACTERISTICS

($T_{amb} = 25\text{ }^{\circ}\text{C}$, $V_s = 12\text{ V}$ unless otherwise noted)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_d Quiescent drain current	$V_3 = 0$		20		mA
$V_{i(\text{threshold})}$ Input limiting voltage	$R_s = 50\ \Omega$ $f = 5.5\text{ MHz}$ $\Delta f = \pm 25\text{ kHz}$		100		μV
V_o Recovered output voltage	$V_i = 10\text{ mV}$ $R_s = 50\ \Omega$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 25\text{ kHz}$		0.9		V_{rms}
Remote control range			70		dB
AMR Amplitude modulation rejection	$V_i = 10\text{ mV}$ $R_s = 50\ \Omega$ $f = 5.5\text{ MHz}$ $m = 0.3$ $\Delta f = \pm 50\text{ kHz}$ $\Delta f = \pm 25\text{ kHz}$		50 45		dB dB
d Distortion	$V_i = 10\text{ mV}$ $R_s = 50\ \Omega$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 25\text{ kHz}$		1		%
R_i Input resistance	$V_i = 10\text{ mV}$ $f = 5.5\text{ MHz}$		5		$\text{k}\Omega$
C_i Input capacitance	$V_i = 10\text{ mV}$ $f = 5.5\text{ MHz}$		3		pF

TBA 261

Fig. 1 - Typical relative audio output voltage vs input voltage

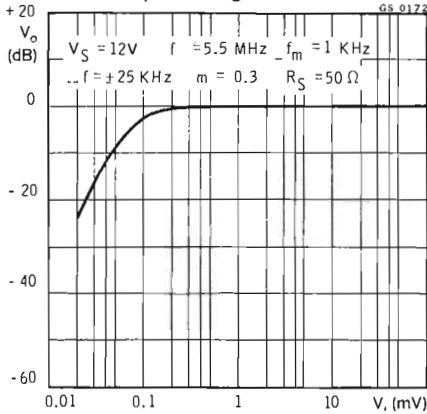


Fig. 2 - Typical relative audio output voltage vs volume control resistance

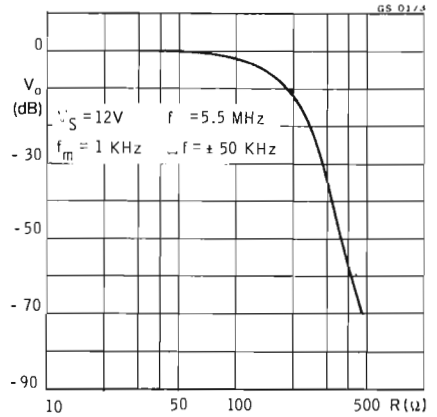


Fig. 3 - Maximum output voltage swing vs load resistance

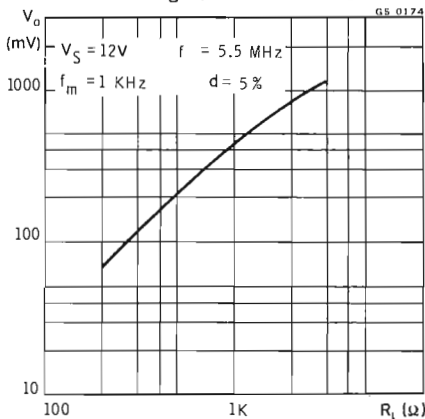
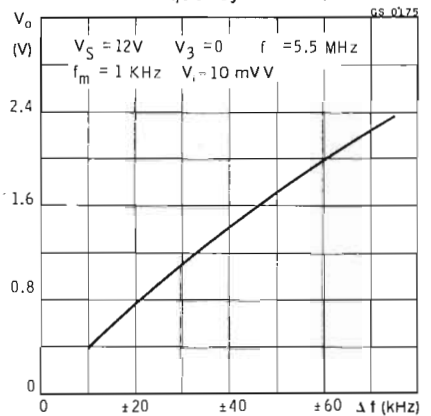


Fig. 4 - Typical audio output voltage vs frequency variation



TBA 261

Fig. 5 - Typical relative audio output voltage vs modulating frequency

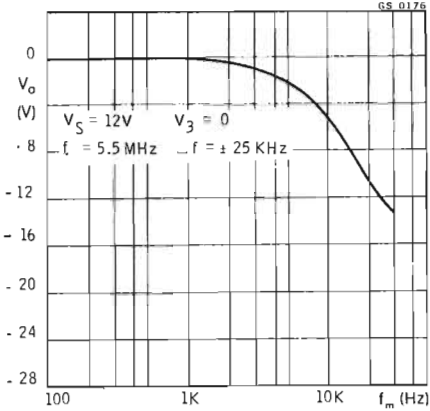


Fig. 6 - Typical distortion vs modulating frequency

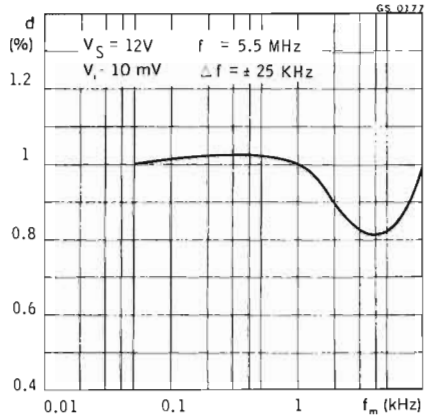


Fig. 7 - Typical distortion vs frequency deviation

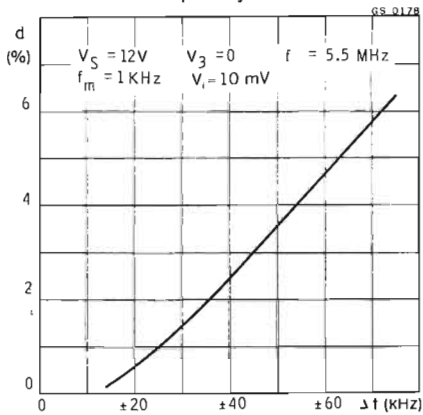
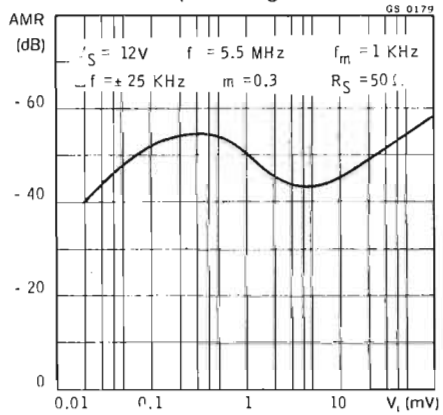


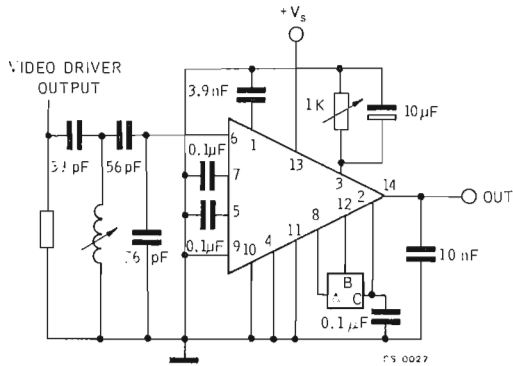
Fig. 8 - Typical AM rejection vs input voltage



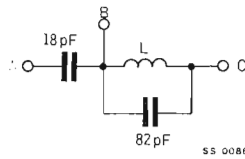
TBA 261

TYPICAL APPLICATION

TV sound IF amplifier (5.5 MHz)



Frequency shift network



$L = 55 t \varnothing = 0.2 \text{ mm.}$

$Q_o = 86$ at 5.5 MHz with tuning capacitor and without shield.

$Q_o = 57$ at 5.5 MHz with tuning capacitor and with shield connected to ground.

LINEAR INTEGRATED CIRCUIT

TBA 311

TV SIGNAL PROCESSING CIRCUIT

The TBA 311 is a monolithic integrated circuit in a 16-lead dual in-line or quad in-line plastic package. It is intended for use as signal processing circuit for black and white and colour television sets.

The circuit is designed for receivers equipped with tubes or transistors in the deflection and video output stages, and with PNP or NPN transistors in the tuner and NPN in the IF amplifier.

Only signals with the negative modulation can be handled by the circuit. The circuit is protected against short circuit between video output and GND. The TBA 311 includes:

- VIDEO PREAMPLIFIER with EMITTER FOLLOWER OUTPUT
- GATED AGC for VIDEO IF AMPLIFIER and TUNER
- NOISE INVERTER CIRCUIT for GATING AGC and SYNC. PULSE SEPARATOR
- HORIZONTAL SYNC. PULSE SEPARATOR
- VERTICAL SYNC. PULSE SEPARATOR
- BLANKING FACILITY for the VIDEO AMPLIFIER

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	16 V
P_{tot}	Power dissipation at $T_{amb} \leq 70^\circ\text{C}$	500 mW
T_{stg}	Storage temperature	-55 to 125 °C
T_{op}	Operating temperature	-25 to 70 °C

ORDERING NUMBERS:

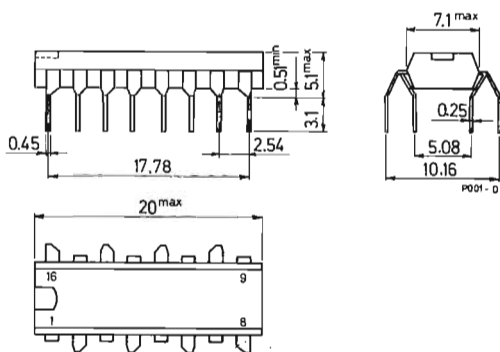
TBA 311 A22 (for 16-lead quad in-line plastic package)

TBA 311 A17 (for 16-lead dual in-line plastic package)

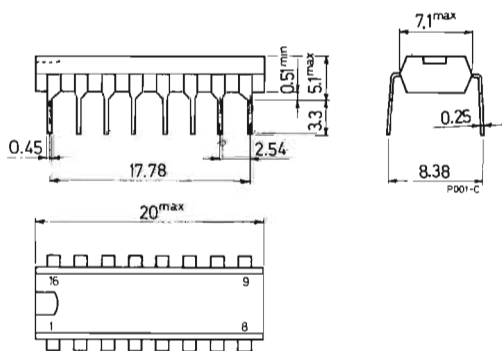
TBA 311

MECHANICAL DATA (Dimensions in mm)

Quad in-line plastic package
for TBA 311 A22



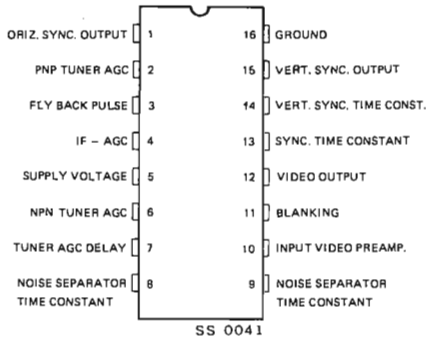
Dual in-line plastic package
for TBA 311 A17



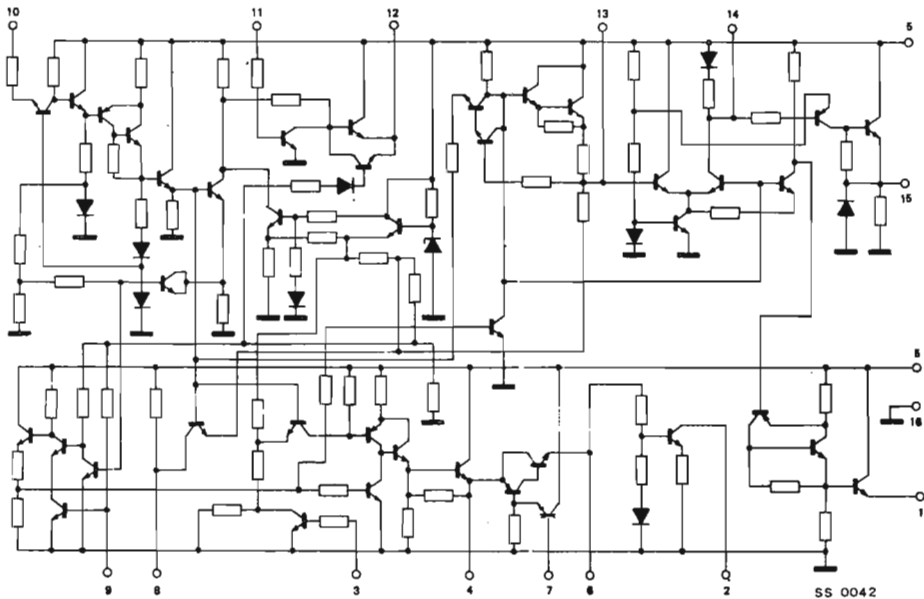
TBA 311

CONNECTION DIAGRAM

(top view)

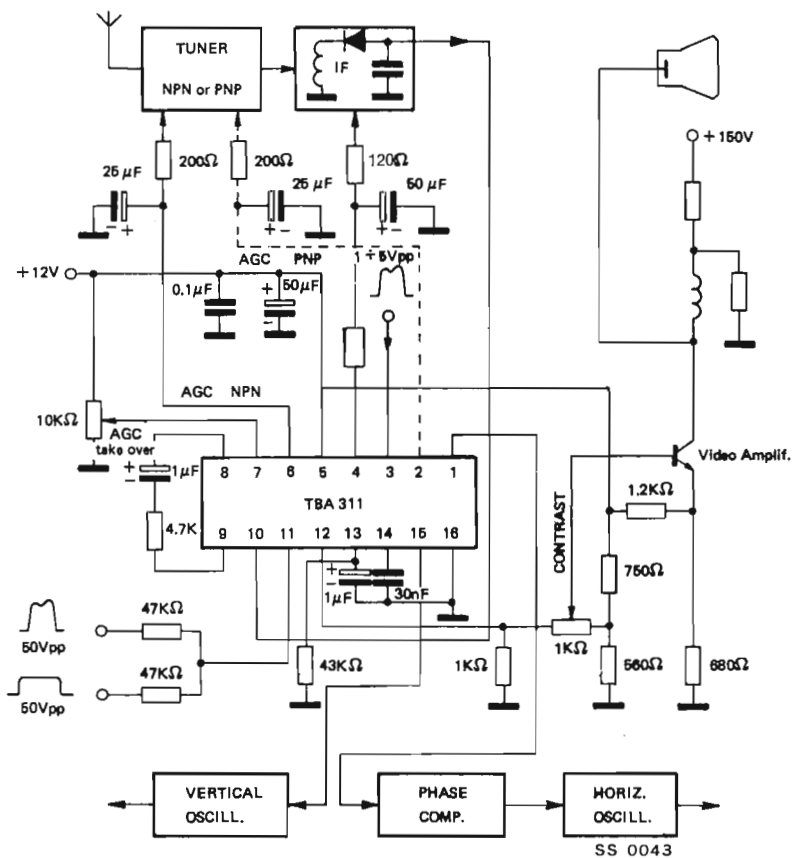


SCHEMATIC DIAGRAM



TBA 311

TEST CIRCUIT



TBA 311

ELECTRICAL CHARACTERISTICS

($T_{amb} = 25^{\circ}\text{C}$, $V_s = 12\text{ V}$ unless otherwise specified, see also test circuit)

Parameter	Min.	Typ.	Max.	Unit
I_d Quiescent drain current		14		mA

VIDEO AMPLIFIER

R_i Input resistance (pin 10)		2.7		$\text{k}\Omega$
C_i Input capacitance (pin 10)		0.8		pF
B Bandwidth (-3 dB)		5		MHz
G_v Voltage gain		9.5		dB
V_i Peak to peak video input voltage (pin 10)	(1)	2		V
V_o Peak to peak video output voltage (pin 12)	(2)	6		V
V Black level at the output (pin 12)	(3)	5		V
I_o Available video peak output current	(4)	20		mA
$\frac{\Delta V_o}{\Delta T_{amb}}$ Video output voltage temperature drift	(5)	1		$\text{mV}/^{\circ}\text{C}$
$\frac{\Delta V}{\Delta T_{amb}}$ Black level temperature drift		0.2		$\text{mV}/^{\circ}\text{C}$
$\frac{\Delta V}{\Delta V_s}$ Black level drift at the output with supply voltage variation		0.5		V/V

VIDEO BLANKING

V_i Peak to peak input voltage (pin 11)		1	5	V
R_i Input resistance (pin 11)		1		$\text{k}\Omega$

AGC CIRCUIT

V Control voltage IF amplifier (pin 4)		0 to 7.5		V
V Control voltage tuner NPN (pin 6)		0 to 6.5		V
PNP (pin 2)		12 to 6		V

TBA 311

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Min.	Typ.	Max.	Unit
$\frac{\Delta V_i}{\Delta V}$	Signal expansion for full control of IF amplifier and tuner		10		%
V	Peak to peak keying input pulse (pin 3) (6)	1		5	V
R_i	Input resistance (pin 3)		2		k Ω

SYNC. CIRCUITS

V_o	Output voltage of horizontal sync. pulse (pin 1)	8.4	10		V
Z_o	Horizontal output impedance (pin 1)		100		Ω
V_o	Output voltage of vertical sync. pulse (pin 15)	8.4	9.5		V
Z_o	Vertical output impedance (pin 15)		2		k Ω

NOTES:

- 1) Negative going video signal (no pre-bias needed for the detector).
- 2) Video signal with negative going sync. pulse.
- 3) Only valid if the video signal is in accordance with the CCIR standard.
- 4) The total load on pin 12 must be such that under nominal conditions $I_o \leq 20$ mA.
- 5) Because the integrated circuit reaches 95% of its final working temperature in 100 seconds, the temperature variations to be considered are those caused by the slower rise in cabinet temperature and by changes in room temperature.
- 6) The TBA 311 may be operated unkeyed but then point 3 must be connected to the positive supply line via a resistor of suitable value (e.g. 10 k Ω). However, the following consequences should be borne in mind:
 - The decoupling capacitors at the IF and tuner control points must be larger to prevent ripple voltages due to the vertical sync pulses. In consequence the AGC will not follow fast signal fluctuations (aircraft flutter).

TBA 331

LINEAR INTEGRATED CIRCUIT

GENERAL PURPOSE

The TBA 331 is an assembly of 5 silicon NPN transistors on a common monolithic substrate in a Jedec TO-116 14-lead dual in-line plastic package. Two transistors are internally connected to form a differential amplifier.

The transistors of the TBA 331 are well suited to low noise general purposes and to a wide variety of applications in low power systems in the DC through VHF range. They may be used as discrete components in conventional circuits, in addition, they provide the very significant inherent integrated circuit advantages of close electrical and thermal matching.

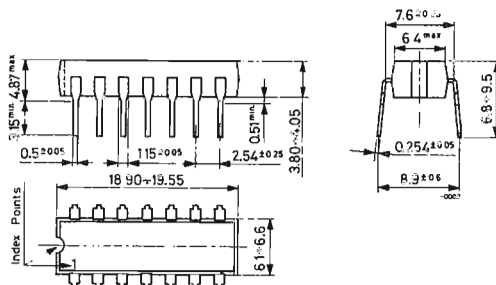
ABSOLUTE MAXIMUM RATINGS

		Each transistor	Total package
V_{CBO}	Collector-base voltage ($I_E = 0$)	20	— V
V_{CEO}	Collector-emitter voltage ($I_B = 0$)	15	— V
V_{CSS}^*	Collector-substrate voltage	20	— V
V_{EBO}	Emitter-base voltage ($I_C = 0$)	5	— V
I_C	Collector current	50	— mA
P_{tot}	Total power dissipation at $T_{amb} \leq 55^\circ\text{C}$ at $T_{amb} > 55^\circ\text{C}$	300	750 mW Derate at 6.67 mW/ $^\circ\text{C}$
T_{stg}	Storage temperature	-25 to 85 $^\circ\text{C}$	
T_{op}	Operating temperature	0 to 85 $^\circ\text{C}$	

* The collector of each transistor of the TBA 331 is isolated from the substrate by an integrated diode. The substrate (pin 13) must be connected to the most negative point in the external circuit to maintain isolation between transistors and to provide for normal transistor action.

MECHANICAL DATA

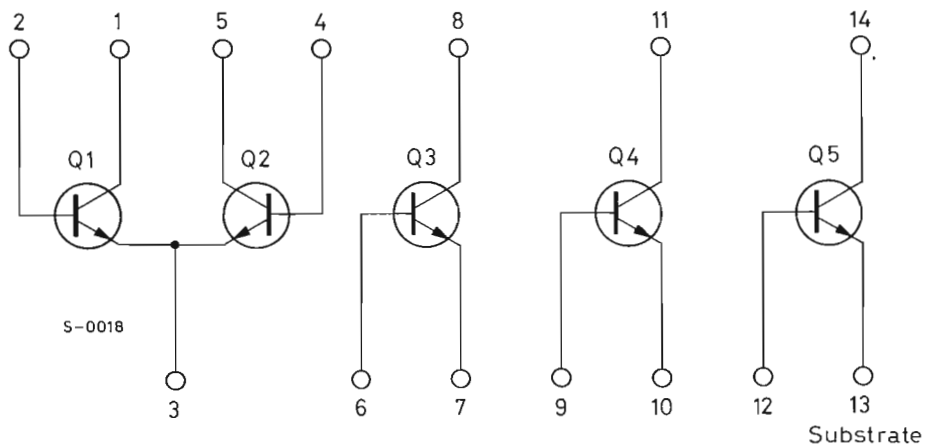
Dimensions in mm



TO-116

TBA 331

SCHEMATIC DIAGRAM



ELECTRICAL CHARACTERISTICS ($T_{amb} = 25^{\circ}\text{C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
I_{CBO} Collector cutoff current ($I_E = 0$)	$V_{CB} = 10\text{ V}$		0.002	40	nA	1
I_{CEO} Collector cutoff current ($I_B = 0$)	$V_{CE} = 10\text{ V}$		see curve	0.5	μA	2
$ I_{B1} - I_{B2} $ Input offset current	$I_C = 1\text{ mA}$ $V_{CE} = 3\text{ V}$		0.3	2	μA	7

TBA 331

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
V_{CBO} Collector-base voltage ($I_E = 0$)	$I_C = 10 \mu A$	20	60		V	—
V_{CEO} Collector-emitter voltage ($I_B = 0$)	$I_C = 1 \text{ mA}$	15	24		V	—
V_{CSS} Collector-substrate voltage ($I_{CSS} = 0$)	$I_C = 10 \mu A$	20	60		V	—
$V_{CE(sat)}$ Collector-emitter saturation voltage	$I_B = 1 \text{ mA}$ $I_C = 10 \text{ mA}$		0.23		V	—
V_{EBO} Emitter-base voltage ($I_C = 0$)	$I_E = 10 \mu A$	5	7		V	—
V_{BE} Base-emitter voltage	$I_E = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $I_E = 10 \text{ mA}$ $V_{CE} = 3 \text{ V}$		0.715		V	4
			0.8		V	4
$ V_{BE1} - V_{BE2} $ Input offset voltage	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$		0.45	5	mV	4-6
$ V_{BE3} - V_{BE4} $ Input offset voltage	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$		0.45	5	mV	4-6
$ V_{BE4} - V_{BE5} $ Input offset voltage	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$		0.45	5	mV	4-6
$ V_{BE5} - V_{BE4} $ Input offset voltage	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$		0.45	5	mV	4-6
$\frac{\Delta V_{BE}}{\Delta T}$ Base-emitter voltage temperature coefficient	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$		-1.9		mV/°C	5
$\frac{ V_{BE1} - V_{BE2} }{\Delta T}$ Input offset voltage temperature coefficient	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$		1.1		$\mu V/^\circ C$	6

TBA 331

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit	Fig.
h_{FE}	DC current gain	$I_C = 10 \text{ mA}$ $V_{CE} = 3 \text{ V}$		100		—	3
		$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$	40	100		—	3
		$I_C = 10 \mu\text{A}$ $V_{CE} = 3 \text{ V}$		54		—	3
f_T	Transition frequency	$I_C = 3 \text{ mA}$ $V_{CE} = 3 \text{ V}$	300	550		MHz	14
NF	Noise figure	$I_C = 100 \mu\text{A}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ kHz}$ $R_g = 1 \text{ k}\Omega$		3.25		dB	8
h_{ie}	Input impedance	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ kHz}$		3.5		$\text{k}\Omega$	9
h_{fe}	Forward current transfer ratio	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ kHz}$		110		—	9
h_{re}	Reverse voltage transfer ratio	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ kHz}$		1.8×10^{-4}		—	9
h_{oe}	Output admittance	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ kHz}$		15.6		μS	9
y_{ie}	Input admittance	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ MHz}$		$0.3 + j0.04$		mS	11
y_{fe}	Forward transadmittance	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ MHz}$		$31 - j1.5$		mS	10
y_{re}	Reverse transadmittance	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ MHz}$		see curve		mS	13

TBA 331

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test Conditions	Min.	Typ.	Max.	Unit	Fig.
y_{oe} Output admittance	$I_C = 1 \text{ mA}$ $V_{CE} = 3 \text{ V}$ $f = 1 \text{ MHz}$		0.001 + j0.03		mS	12
C_{EBO} Emitter-base capacitance	$I_C = 0$ $V_{EB} = 3 \text{ V}$		0.6		pF	—
C_{CBO} Collector-base capacitance	$I_E = 0$ $V_{CB} = 3 \text{ V}$		0.58		pF	—
C_{CSS} Collector-substrate capacitance	$I_C = 0$ $V_{CSS} = 3 \text{ V}$		2.8		pF	—

Fig. 1-Typical collector cutoff current

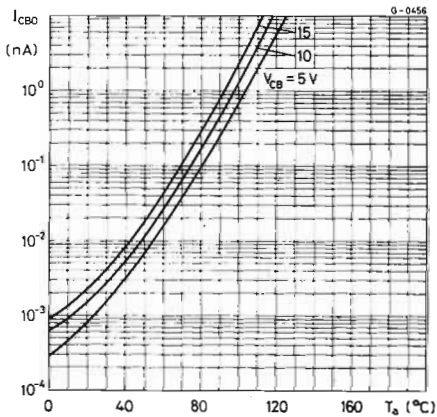
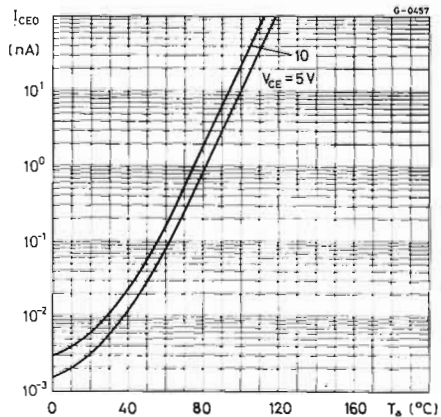


Fig. 2-Typical collector cutoff current



TBA 331

Fig. 3 - Typical DC current gain

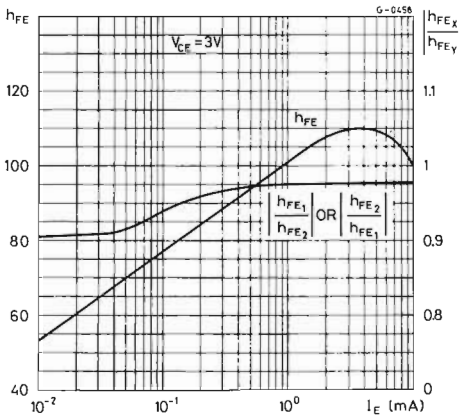


Fig. 4 - Typical input voltage and input voltage offset

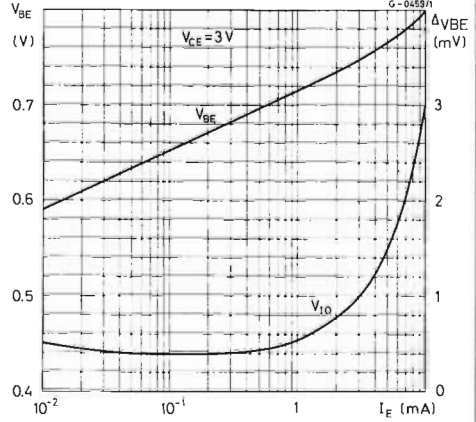


Fig. 5 - Typical input characteristic for each transistor

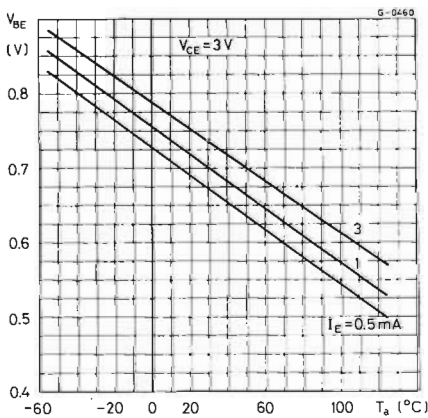
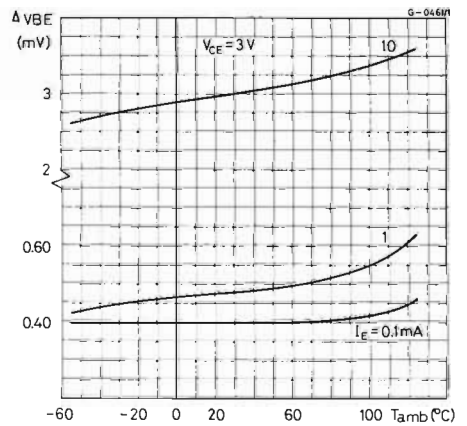


Fig. 6 - Typical input voltage offset



TBA 331

Fig. 7 - Typical input current offset for matched transistor pair

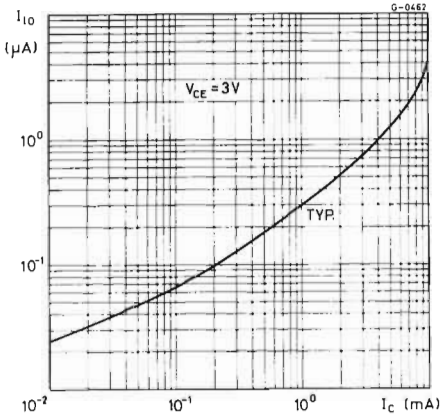


Fig. 8 - Typical noise figure

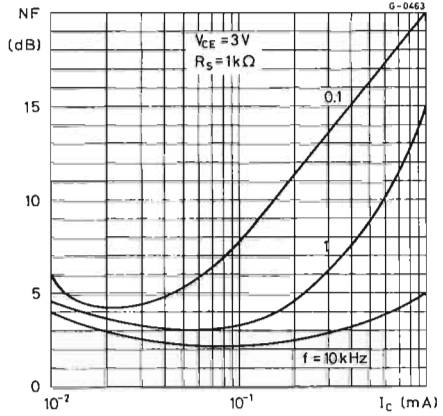


Fig. 9 - Typical normalized h parameters

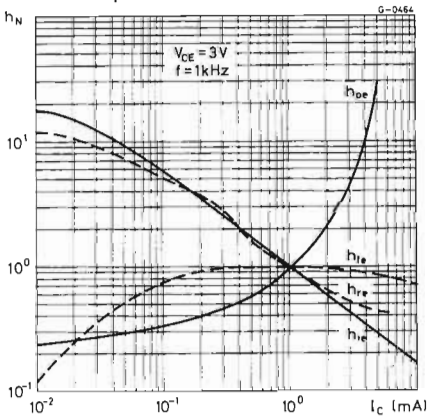
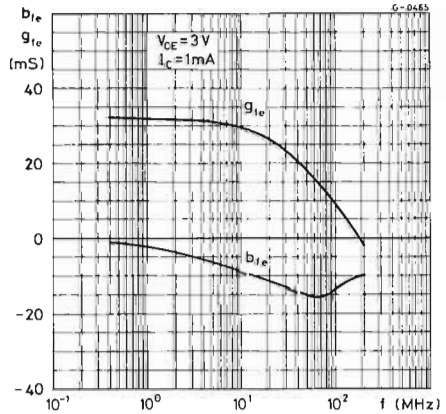


Fig. 10 - Typical forward admittance



TBA 331

Fig. 11 - Typical input admittance

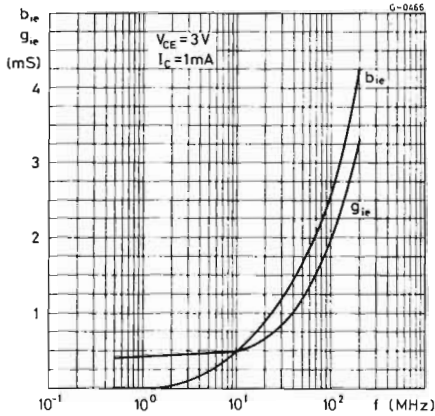


Fig. 12 - Typical output admittance

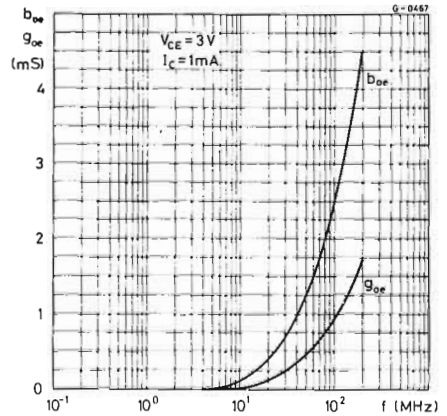


Fig. 13 - Typical reverse admittance

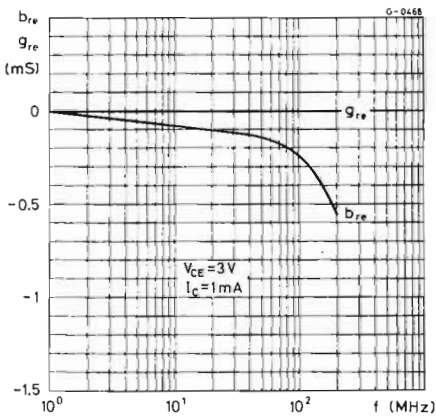
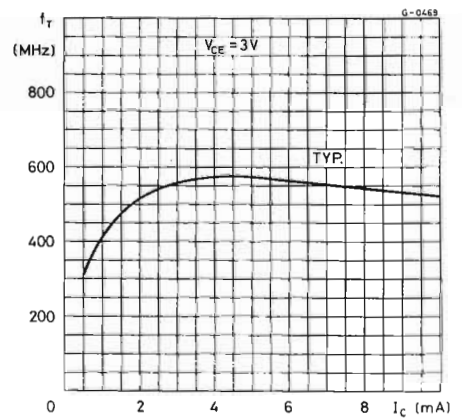


Fig. 14 - Transition frequency



LINEAR INTEGRATED CIRCUIT

TBA 435

VOLTAGE REGULATOR

- OUTPUT CURRENT ≥ 100 mA
- TIGHT TOLERANCE for OUTPUT VOLTAGE
- LOAD REGULATION $\leq 1\%$
- RIPPLE REJECTION 57 dB TYPICAL
- OVERLOAD and SHORT CIRCUIT PROTECTION

The TBA 435 is an integrated monolithic 8.5 V voltage regulator in TO-39 metal case which can supply more than 100 mA. The device features high temperature stability, internal overload and short circuit protection, low output impedance and excellent transient response. The TBA 435 is intended for use as voltage supply for consumer circuits and for any other industrial application.

ABSOLUTE MAXIMUM RATINGS

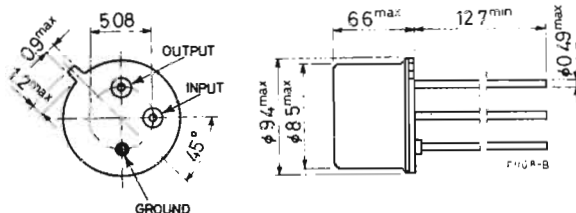
V_i	Input voltage	20	V
P_{tot}	Power dissipation at $T_{amb} = 25^\circ\text{C}$	0.75	W
	at $T_{case} = 25^\circ\text{C}$	4	W
T_{stg}	Storage temperature	-55 to 150	$^\circ\text{C}$
T_j	Junction temperature	175	$^\circ\text{C}$
T_{op}	Operating temperature	0 to 70	$^\circ\text{C}$

ORDERING NUMBER: TBA 435A X5

MECHANICAL DATA

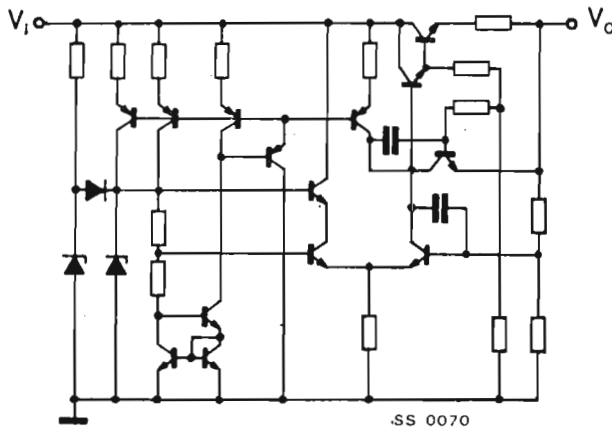
Dimensions in mm

Ground connected to case



TBA 435

SCHEMATIC DIAGRAM



THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	37.5	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	200	°C/W

ELECTRICAL CHARACTERISTICS $(T_j = 25\text{ °C unless otherwise specified})$

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_o Output voltage	$V_i = 11.5\text{ V to }20\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\text{ }\mu\text{F}$	8.1	8.5	8.9	V
$\frac{\Delta V_o}{V_o}$ Load regulation	$V_i = 11.5\text{ V to }20\text{ V}$ $I_o = 5\text{ mA to }100\text{ mA}$ $C_L = 10\text{ }\mu\text{F}$		0.3	1	%
I_o Regulated current	$V_i = 15\text{ V}$ $\frac{\Delta V_o}{V_o} \leq 1\%$	100	140		mA

TBA 435

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
I_o	Max. regulated current	$V_i = 15\text{ V}$	130	150	200	mA
R_o	Output resistance	$V_i = 15\text{ V}$ $I_o = 5\text{ mA to }100\text{ mA}$		0.1		Ω
$\frac{\Delta V_o}{V_o}$	Line regulation	$V_i = 11.5\text{ V to }20\text{ V}$ $I_o = 5\text{ mA}$		0.15	0.6	%
SVR	Supply voltage rejection	$V_i = 13.5\text{ V}$ $\Delta V_i = 4\text{ V}_{pp}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$ $f = 100\text{ Hz}$	46	57		dB
e_N	Output noise voltage	$V_i = 15\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$ $B = 100\text{ Hz to }100\text{ kHz}$		100		μV
I_d	Quiescent drain current	$V_i = 20\text{ V}$ $I_o = 0$	5	9	16	mA
$\frac{\Delta V_o}{\Delta T_{amb}}$	Temperature coefficient	$V_i = 15\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$ $T_{amb} = 0\text{ to }70\text{ }^\circ\text{C}$		0.85		mV/ $^\circ\text{C}$
I_{sc}	Output short circuit current	$V_i = 20\text{ V}$ $V_o = 0$		40	60	mA

TBA 435

Fig. 1 - Typical output voltage vs output current

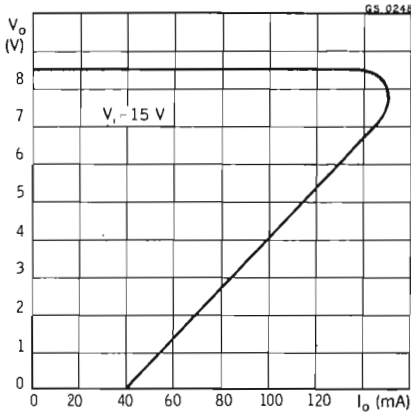


Fig. 2 - Power rating chart

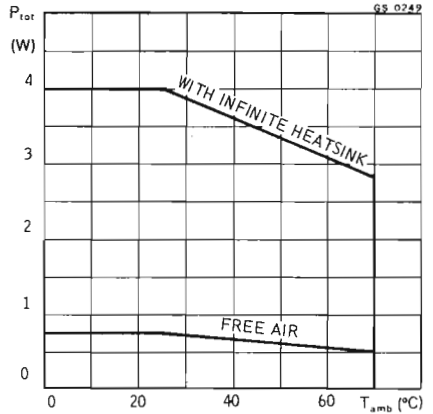


Fig. 3 - Maximum output current vs junction temperature

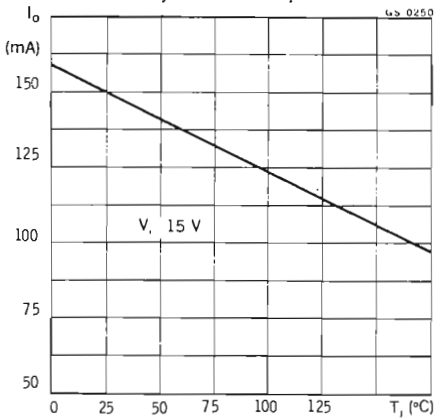
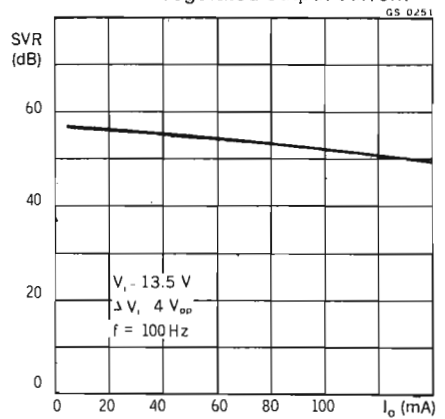


Fig. 4 - Typical ripple rejection vs regulated output current



TBA 435

Fig. 5 - Typical ripple rejection vs frequency

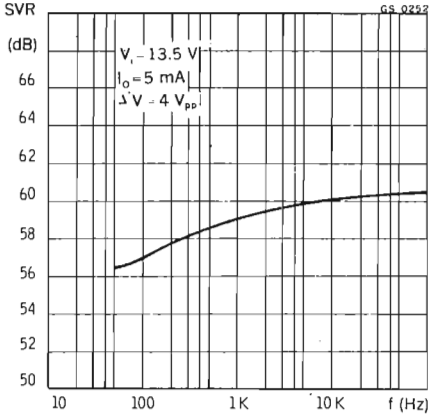


Fig. 6 - Maximum output current vs input voltage

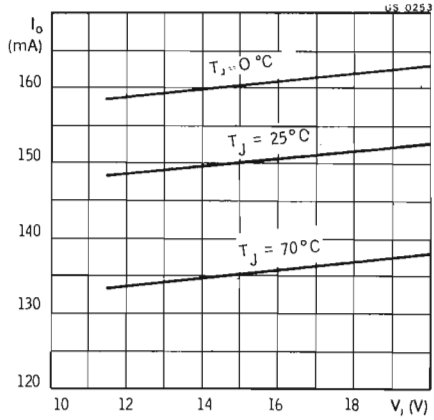


Fig. 7 - Typical short circuit output current vs input voltage

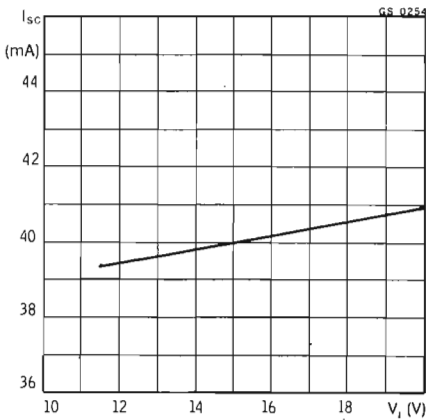
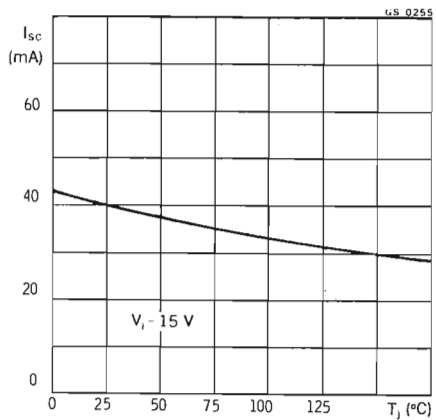


Fig. 8 - Typical short circuit output current vs junction temperature



TBA 435

Fig. 9 - Typical dropout voltage vs output current

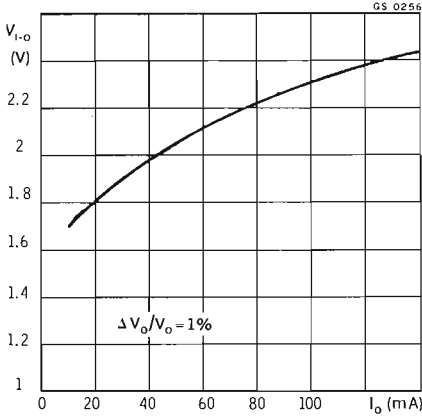


Fig. 10 - Typical quiescent drain current vs junction temperature

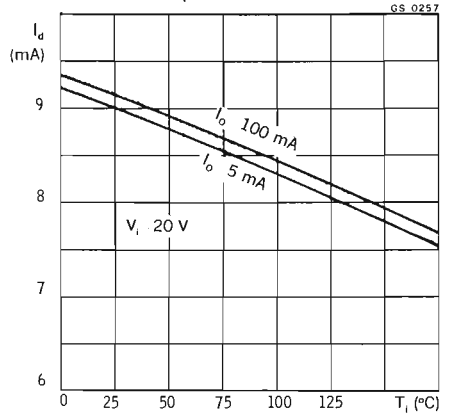


Fig. 11 - Typical quiescent drain current vs input voltage

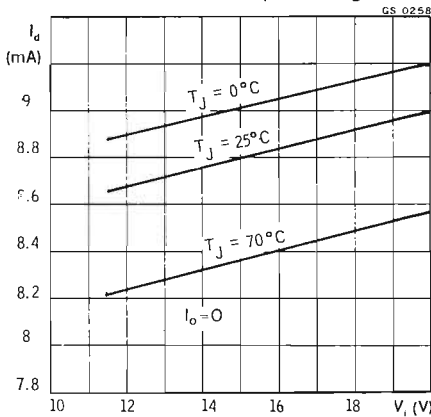
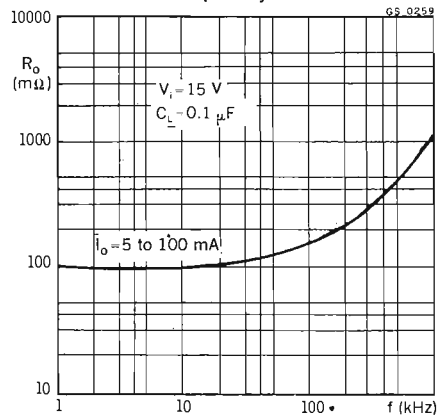
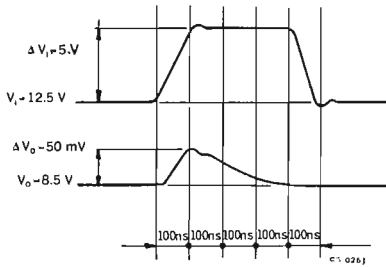


Fig. 12 - Typical output resistance vs frequency

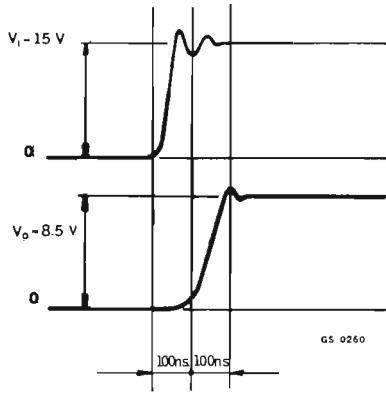


TBA 435

Line transient response
($I_o = 5 \text{ mA}$)



Turn on time
($I_o = 100 \text{ mA}$)



TYPICAL APPLICATIONS

Fig. 13 - Positive output voltage regulator

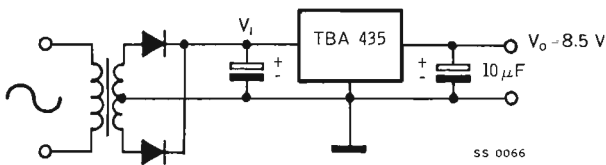
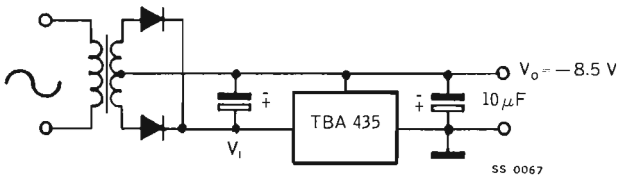
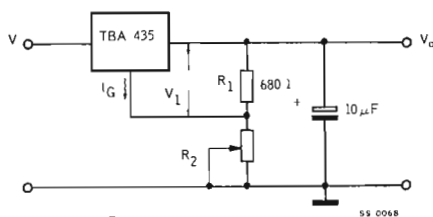


Fig. 14 - Negative output voltage regulator



TBA 435

Fig. 15 - Adjustable output voltage regulator



$$V_o = V_1 \left(1 + \frac{R_2}{R_1}\right) + I_G R_2$$

V_1 18 V

V_o 8.5 to 11 V

$I_G > 80$ mA

$R_D = 100$ mΩ

R_2 potentiometer 0 to 150 Ω

Typical adjustable output voltage vs output current

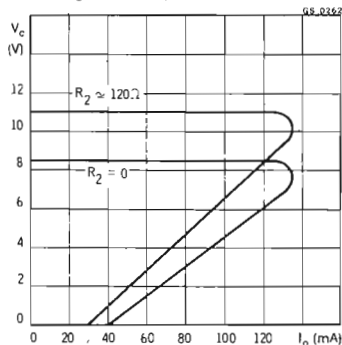
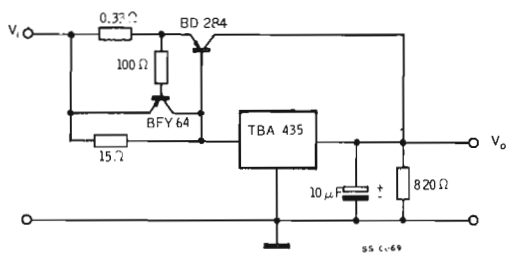


Fig. 16 - PNP current boost circuit



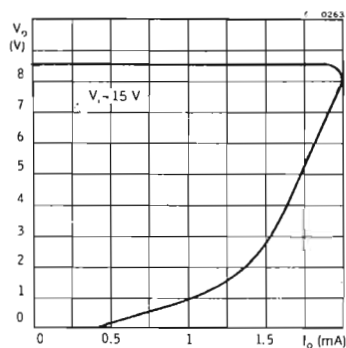
V_1 - 15 V

V_o 8.5 V

$I_o = 2$ A

$R_D = 20$ mΩ

Typical output voltage vs output current



TBA 625A

LINEAR INTEGRATED CIRCUIT

VOLTAGE REGULATOR

- OUTPUT CURRENT ≥ 100 mA
- TIGHT TOLERANCE for OUTPUT VOLTAGE
- LOAD REGULATION $\leq 1\%$
- RIPPLE REJECTION 60 dB TYPICAL
- OVERLOAD and SHORT CIRCUIT PROTECTION

The TBA 625A is an integrated monolithic 5 V voltage regulator in TO-39 metal case which can supply more than 100 mA. The device features high temperature stability, internal overload and short circuit protection, low output impedance and excellent transient response. The TBA 625A is intended for use as voltage supply for digital circuits and for any other industrial application.

ABSOLUTE MAXIMUM RATINGS

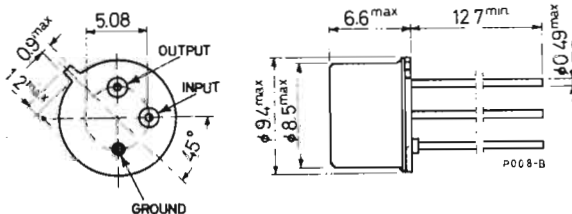
V_i	Input voltage	20	V
P_{tot}	Power dissipation at $T_{amb} = 25^\circ\text{C}$	0.75	W
	at $T_{case} = 25^\circ\text{C}$	4	W
T_{stg}	Storage temperature	-55 to 150	$^\circ\text{C}$
T_j	Junction temperature	175	$^\circ\text{C}$
T_{op}	Operating temperature	0 to 70	$^\circ\text{C}$

ORDERING NUMBER: TBA 625A X5

MECHANICAL DATA

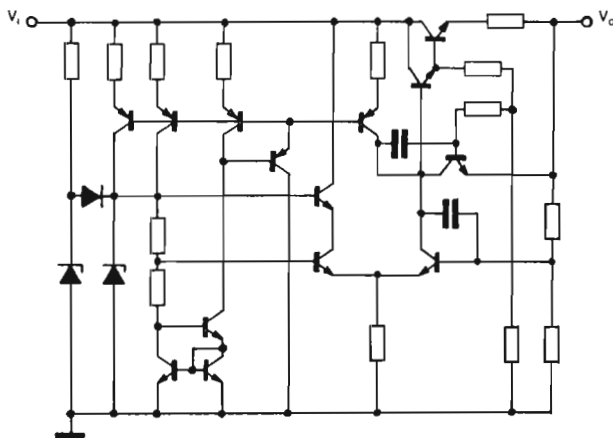
Dimensions in mm

Ground connected to case



TBA 625A

SCHEMATIC DIAGRAM



SS 0071

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	37.5 °C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	200 °C/W

ELECTRICAL CHARACTERISTICS ($T_i = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_o Output voltage	$V_i = 8\text{ V to }20\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$	4.75	5	5.25	V
$\frac{\Delta V_o}{V_o}$ Load regulation	$V_i = 8\text{ V to }20\text{ V}$ $I_o = 5\text{ mA to }100\text{ mA}$ $C_L = 10\ \mu$		0.3	1	%
I_o Regulated current	$V_i = 12\text{ V}$ $\frac{\Delta V_o}{V_o} \leq 1\%$	100	140		mA

TBA 625A

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
I_o	Max. regulated current	$V_i = 12\text{ V}$	130	150	200	mA
R_o	Output resistance	$V_i = 12\text{ V}$ $I_o = 5\text{ mA to }100\text{ mA}$	0.1			Ω
$\frac{\Delta V_o}{V_o}$	Line regulation	$V_i = 8\text{ V to }20\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$	0.2			%
SVR	Supply voltage rejection	$V_i = 10\text{ V}$ $\Delta V_i = 4\text{ V}_{pp}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$ $f = 100\text{ Hz}$	46	60		dB
e_N	Output noise voltage	$V_i = 12\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$ $B = 10\text{ Hz to }100\text{ kHz}$	70			μV
I_d	Quiescent drain current	$V_i = 20\text{ V}$ $I_o = 0$	5	9	16	mA
$\frac{\Delta V_o}{\Delta T_{amb}}$	Temperature coefficient	$V_i = 12\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$ $T_{amb} = 0\text{ to }70\text{ }^\circ\text{C}$	0.5			mV/ $^\circ\text{C}$
I_{sc}	Output short circuit current	$V_i = 20\text{ V}$ $V_o = 0$	45	65		mA

TBA 625A

Fig. 1 - Typical output voltage vs output current

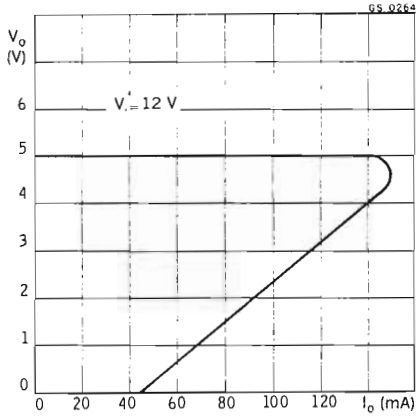


Fig. 2 - Power rating chart

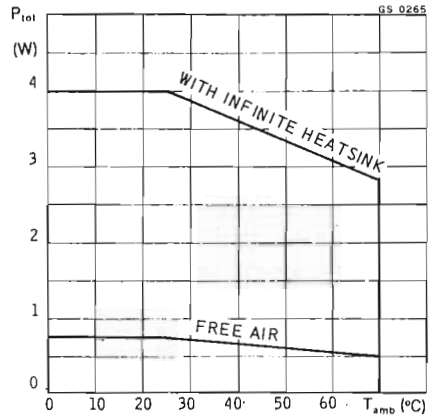


Fig. 3 - Maximum output current vs junction temperature

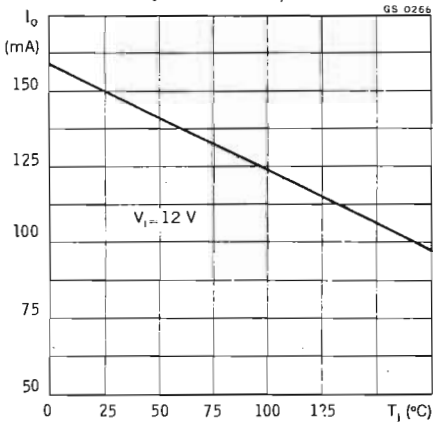
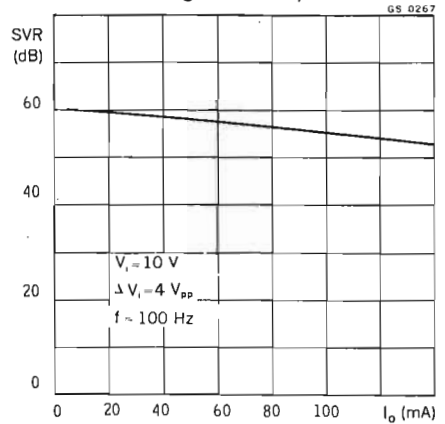


Fig. 4 - Typical ripple rejection vs regulated output current



TBA 625A

Fig. 5 - Typical ripple rejection vs frequency

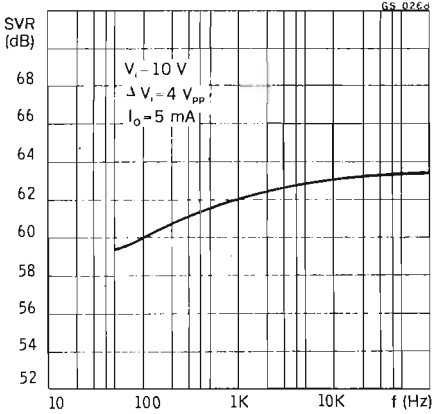


Fig. 6 - Maximum output current vs input voltage

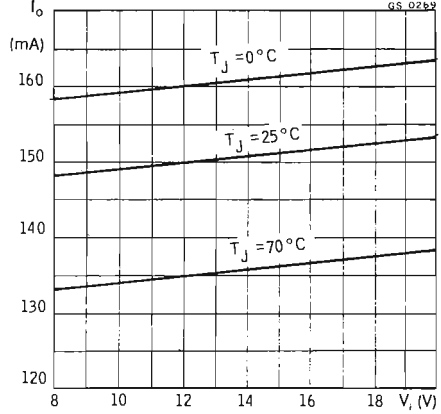


Fig. 7 - Typical short circuit output current vs input voltage

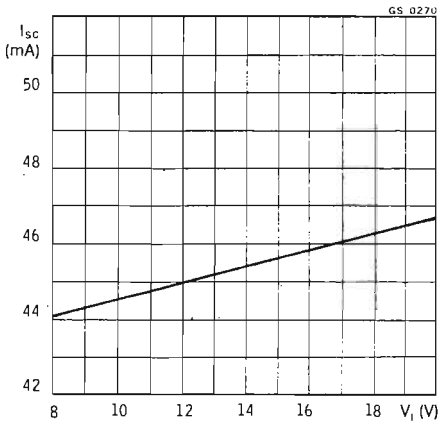
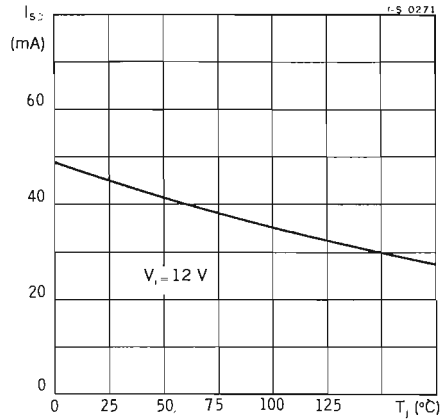


Fig. 8 - Typical short circuit output current vs junction temperature



TBA 625A

Fig. 9 - Typical dropout voltage vs output current

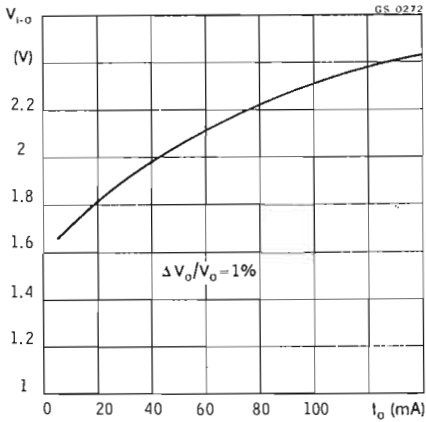


Fig. 10 - Typical quiescent drain current vs junction temperature

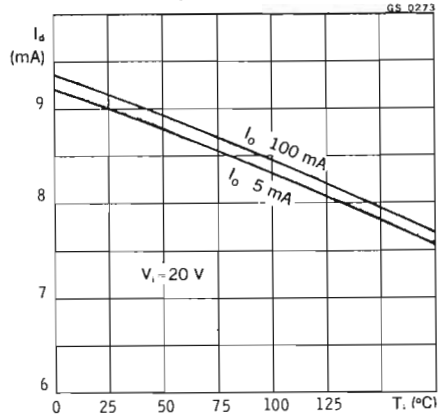


Fig. 11 - Typical quiescent drain current vs input voltage

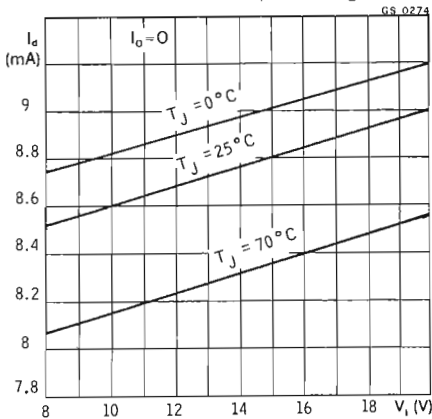
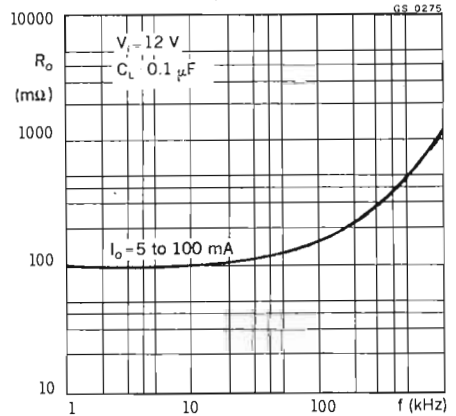
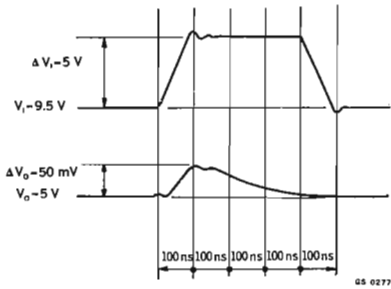


Fig. 12 - Typical output resistance vs frequency

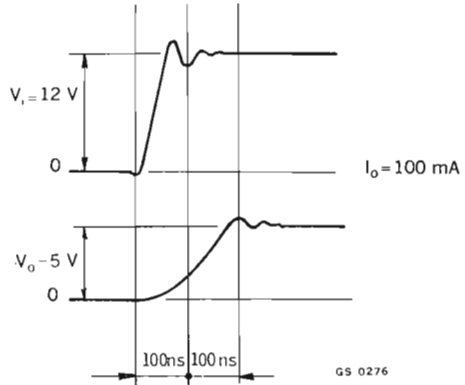


TBA 625A

Line transient response
($I_o = 5 \text{ mA}$)



Turn on time
($I_o = 100 \text{ mA}$)



TYPICAL APPLICATIONS

Fig. 13 - Positive output voltage regulator

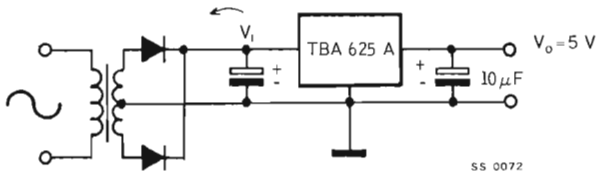
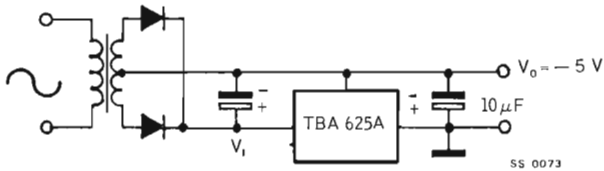
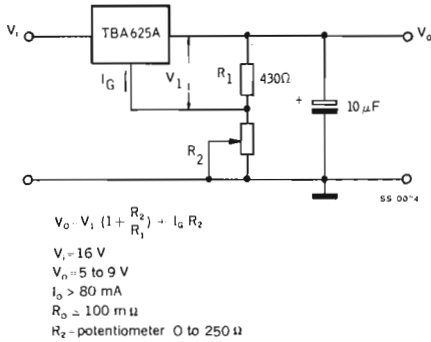


Fig. 14 - Negative output voltage regulator



TBA 625A

Fig. 15 - Adjustable output voltage regulator



Typical adjustable output voltage vs output current

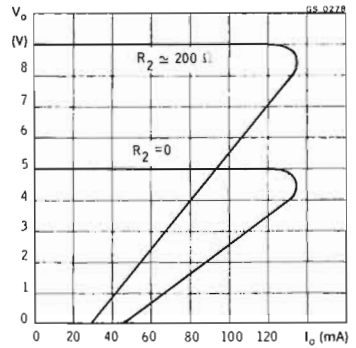
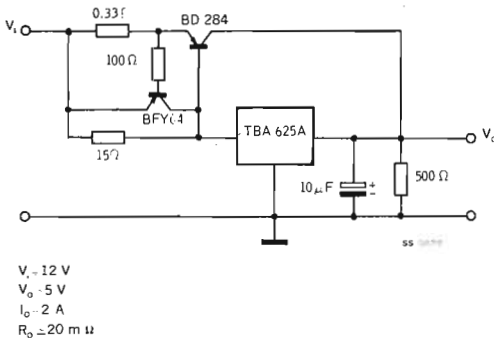
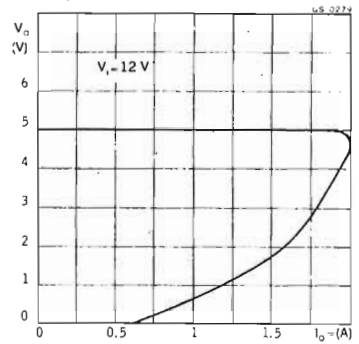


Fig. 16 - PNP current boost circuit



Typical output voltage vs output current



TBA 625B

LINEAR INTEGRATED CIRCUIT

VOLTAGE REGULATOR

- OUTPUT CURRENT ≥ 100 mA
- TIGHT TOLERANCE for OUTPUT VOLTAGE
- LOAD REGULATION $\leq 1\%$
- RIPPLE REJECTION 54 dB TYPICAL
- OVERLOAD and SHORT CIRCUIT PROTECTION

The TBA 625B is an integrated monolithic 12 V voltage regulator in TO-39 metal case which can supply more than 100 mA. The device features high temperature stability, internal overload and short circuit protection, low output impedance and excellent transient response. The TBA 625B is intended for use as voltage supply for digital circuits with high noise immunity, linear integrated circuits and for any other industrial applications.

ABSOLUTE MAXIMUM RATINGS

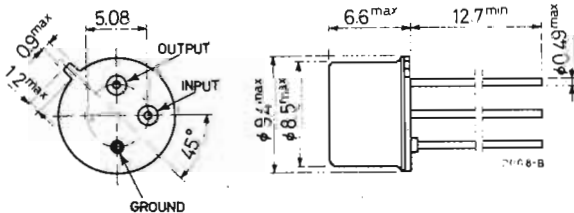
V_i	Input voltage	27 V
P_{tot}	Power dissipation at $T_{amb} = 25^\circ\text{C}$	0.75 W
	at $T_{case} = 25^\circ\text{C}$	4 W
T_{stg}	Storage temperature	-55 to 150 $^\circ\text{C}$
T_j	Junction temperature	175 $^\circ\text{C}$
T_{op}	Operating temperature	0 to 70 $^\circ\text{C}$

ORDERING NUMBER: TBA 625B X5

MECHANICAL DATA

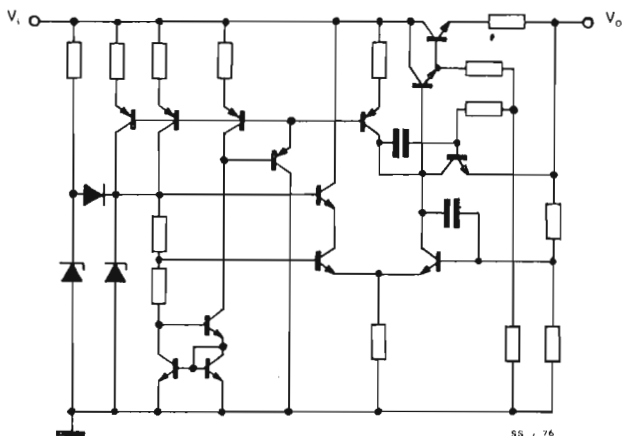
Dimensions in mm

Ground connected to case



TBA 625B

SCHEMATIC DIAGRAM



THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	37.5	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	200	°C/W

ELECTRICAL CHARACTERISTICS $(T_j = 25^\circ\text{C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_o Output voltage	$V_i = 15\text{ V to }27\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\ \mu\text{F}$	11.4	12	12.6	V
$\frac{\Delta V_o}{V_o}$ Load regulation coefficient	$V_i = 15\text{ V to }27\text{ V}$ $I_o = 5\text{ mA to }100\text{ mA}$ $C_L = 10\ \mu\text{F}$		0.3	1	%
I_o Regulated current	$V_i = 12\text{ V}$ $\frac{\Delta V_o}{V_o} \leq 1\%$	100	140		mA

TBA 625B

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
I_o	Max. regulated current	$V_i = 21 \text{ V}$	120	150	200	mA
R_o	Output resistance	$V_i = 21 \text{ V}$ $I_o = 5 \text{ mA to } 100 \text{ mA}$		0.1		Ω
$\frac{\Delta V_o}{V_o}$	Line regulation coefficient	$V_i = 15 \text{ V to } 27 \text{ V}$ $I_o = 5 \text{ mA}$ $C_L = 10 \mu\text{F}$		0.2	0.5	%
SVR	Supply voltage rejection	$V_i = 17 \text{ V}$ $\Delta V_i = 4 \text{ V}_{pp}$ $I_o = 5 \text{ mA}$ $C_L = 10 \mu\text{F}$ $f = 100 \text{ Hz}$		46	54	dB
e_N	Output noise voltage	$V_i = 21 \text{ V}$ $I_o = 5 \text{ mA}$ $C_L = 10 \mu\text{F}$ $B = 10 \text{ Hz to } 100 \text{ kHz}$		150		μV
I_d	Quiescent drain current	$V_i = 27 \text{ V}$ $I_o = 0$	6	10	18	mA
$\frac{\Delta V_o}{\Delta T_{amb}}$	Voltage/temperature coefficient	$V_i = 21 \text{ V}$ $I_o = 5 \text{ mA}$ $C_L = 10 \mu\text{F}$ $T_{amb} = 0 \text{ to } 70 \text{ }^\circ\text{C}$		0.85		mV/ $^\circ\text{C}$
I_{sc}	Output short circuit current	$V_i = 27 \text{ V}$ $V_o = 0$		35	55	mA

TBA 625B

Fig. 1 - Typical output voltage vs output current

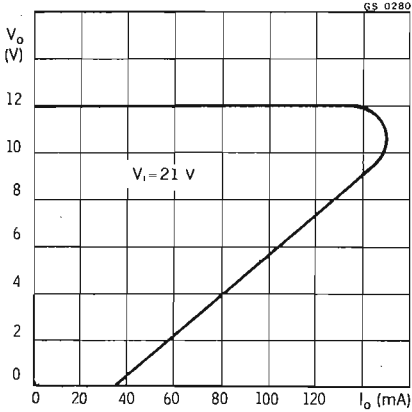


Fig. 2 - Power rating chart

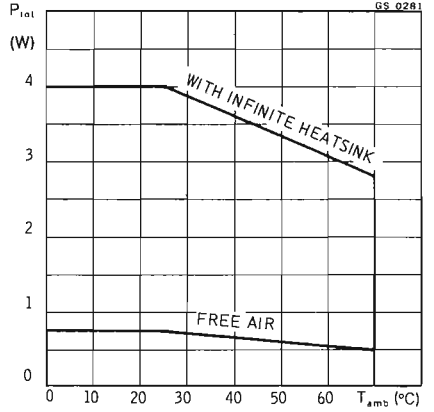


Fig. 3 - Maximum output current vs junction temperature

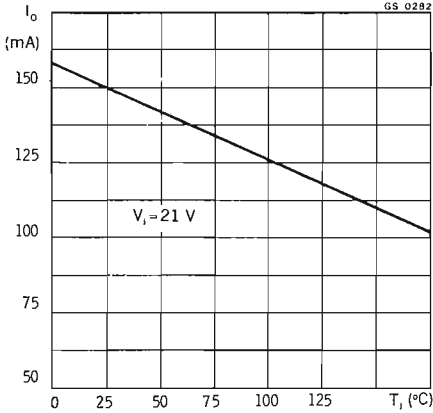
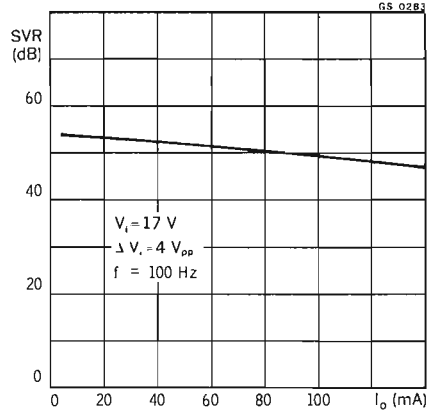


Fig. 4 - Typical ripple rejection vs regulated output current



TBA 625B

Fig. 5 - Typical ripple rejection vs frequency

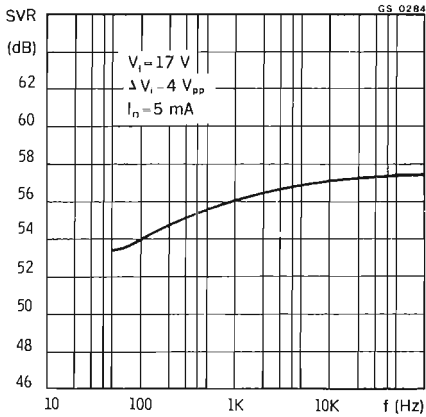


Fig. 6 - Maximum output current vs input voltage

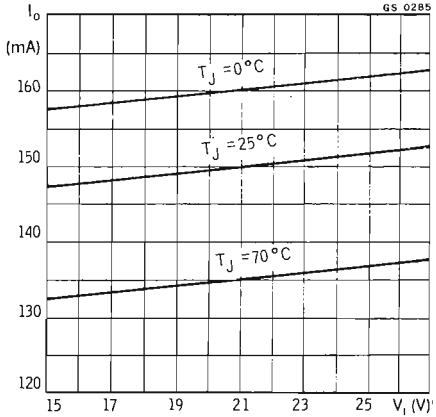


Fig. 7 - Typical short circuit output current vs input voltage

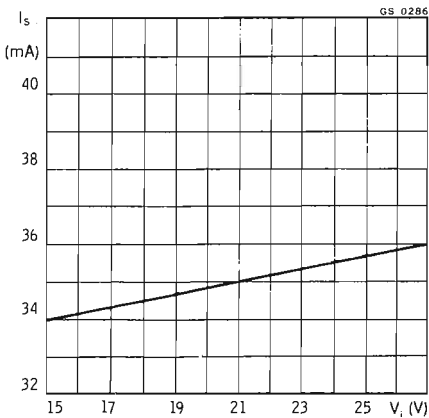
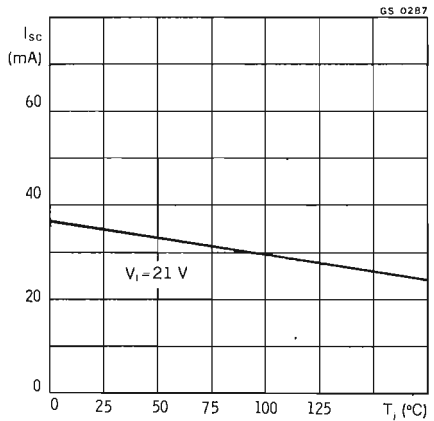


Fig. 8 - Typical short circuit output current vs junction temperature



TBA 625B

Fig. 9 - Typical dropout voltage vs output current

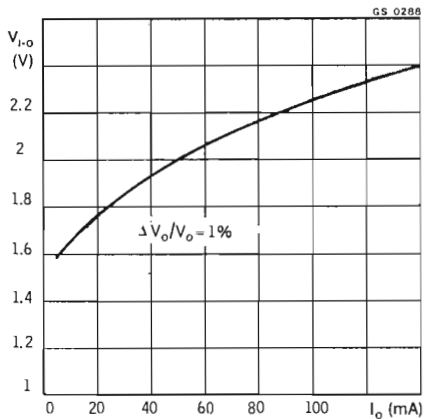


Fig. 10 - Typical quiescent drain current vs junction temperature

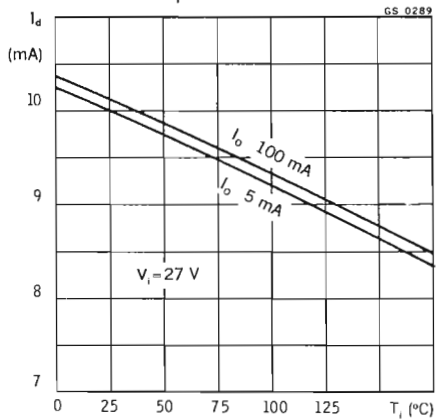


Fig. 11 - Typical quiescent drain current vs input voltage

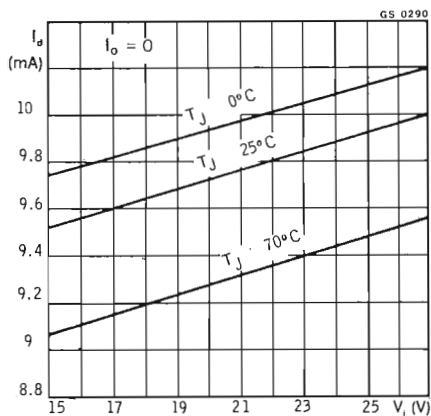
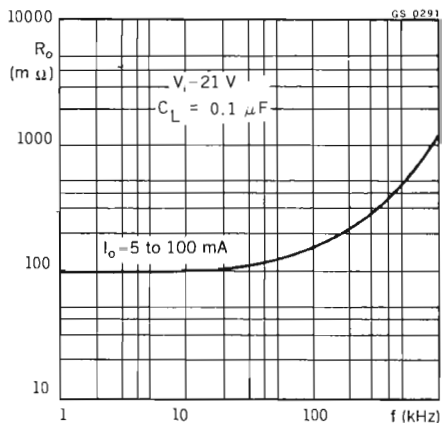
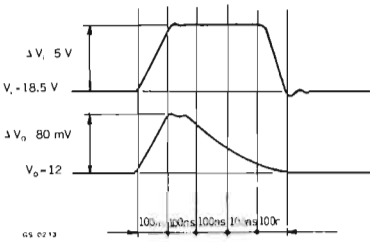


Fig. 12 - Typical output resistance vs frequency

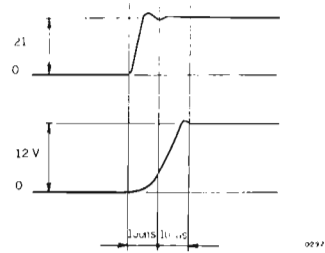


TBA 625B

Line transient response
($I_o = 5 \text{ mA}$)



Turn on time
($I_o = 100 \text{ mA}$)



TYPICAL APPLICATIONS

Fig. 13 - Positive output voltage regulator

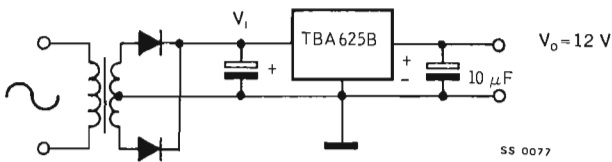
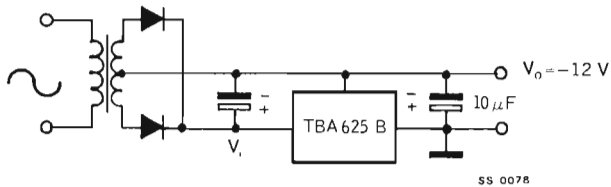
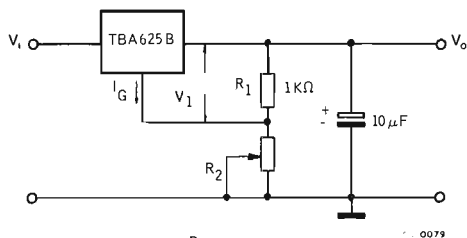


Fig. 14 - Negative output voltage regulator



TBA 625B

Fig. 15 - Adjustable output voltage regulator



$$V_o = V_1 \left(1 + \frac{R_2}{R_1}\right) + I_G R_2$$

$V_1 = 24 \text{ V}$
 $V_o = 12 \text{ to } 15 \text{ V}$
 $I_o > 80 \text{ mA}$
 $R_o \approx 100 \text{ m}\Omega$
 $R_2 = \text{potentiometer } 0 \text{ to } 150 \Omega$

Typical adjustable output voltage vs output current

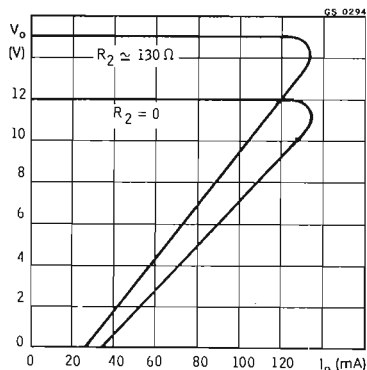
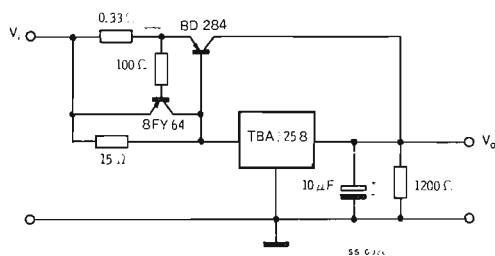
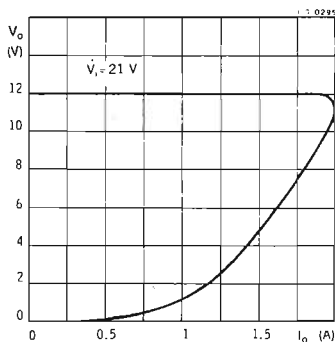


Fig. 16 - PNP current boost circuit



$V_i = 21 \text{ V}$
 $V_o = 12 \text{ V}$
 $I_o = 2 \text{ A}$
 $R_o \approx 20 \text{ m}\Omega$

Typical output voltage vs output current



TBA 625C

LINEAR INTEGRATED CIRCUIT

VOLTAGE REGULATOR

- OUTPUT CURRENT ≥ 100 mA
- TIGHT TOLERANCE for OUTPUT VOLTAGE
- LOAD REGULATION $\leq 1\%$
- RIPPLE REJECTION 51 dB TYPICAL
- OVERLOAD and SHORT CIRCUIT PROTECTION

The TBA 625C is an integrated monolithic 15 V voltage regulator in TO-39 metal case which can supply more than 100 mA. The device features high temperature stability, internal overload and short circuit protection, low output impedance and excellent transient response. The TBA 625C is intended for use as voltage supply for digital circuits with high noise immunity, linear integrated circuits and for any other industrial applications.

ABSOLUTE MAXIMUM RATINGS

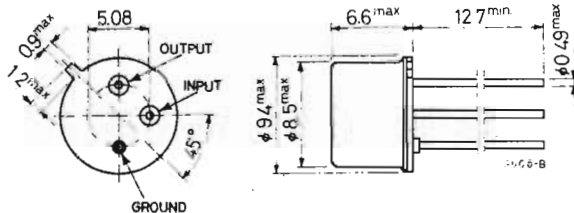
V_i	Input voltage	27 V
P_{tot}	Power dissipation at $T_{amb} = 25^\circ\text{C}$	0.75 W
	at $T_{case} = 25^\circ\text{C}$	4 W
T_{op}	Storage temperature	-55 to 150 $^\circ\text{C}$
T_j	Junction temperature	175 $^\circ\text{C}$
T_{op}	Operating temperature	0 to 70 $^\circ\text{C}$

ORDERING NUMBER: TBA 625C X5

MECHANICAL DATA

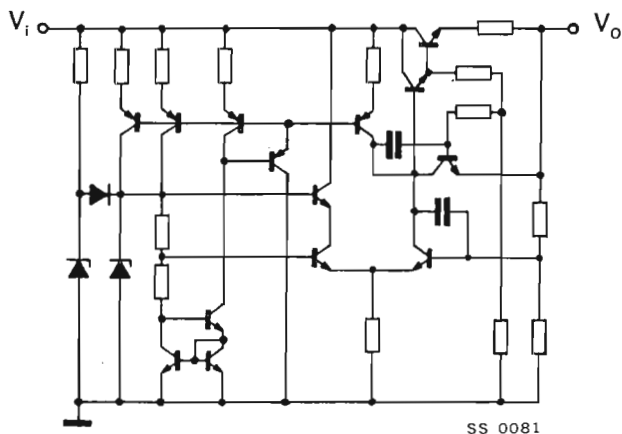
Dimensions in mm

Ground connected to case



TBA 625C

SCHEMATIC DIAGRAM



THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	37.5	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	200	°C/W

ELECTRICAL CHARACTERISTICS $(T_j = 25\text{ °C unless otherwise specified})$

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_o Output voltage	$V_i = 18\text{ V to }27\text{ V}$ $I_o = 5\text{ mA}$ $C_L = 10\text{ }\mu\text{F}$	14.25	15	15.75	V
$\frac{\Delta V_o}{V_o}$ Load regulation	$V_i = 18\text{ V to }27\text{ V}$ $I_o = 5\text{ mA to }100\text{ mA}$ $C_L = 10\text{ }\mu\text{F}$		0.3	1	%
I_o Regulated current	$V_i = 24\text{ V}$ $\frac{\Delta V_o}{V_o} \leq 1\%$	100	140		mA

TBA 625C

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
I_o	Max. regulated current	$V_i = 24 \text{ V}$	120	150	200	mA
R_o	Output resistance	$V_i = 24 \text{ V}$ $I_o = 5 \text{ mA to } 100 \text{ mA}$		0.1		Ω
$\frac{\Delta V_o}{V_o}$	Line regulation	$V_i = 18 \text{ V to } 27 \text{ V}$ $I_o = 5 \text{ mA}$ $C_L = 10 \mu\text{F}$		0.25	0.5	%
SVR	Supply voltage rejection	$V_i = 20 \text{ V}$ $\Delta V_i = 4 \text{ V}_{pp}$ $I_o = 5 \text{ mA}$ $C_L = 10 \mu\text{F}$ $f = 100 \text{ Hz}$	46	51		dB
e_N	Output noise voltage	$V_i = 24 \text{ V}$ $I_o = 5 \text{ mA}$ $C_L = 10 \mu\text{F}$ $B = 10 \text{ Hz to } 100 \text{ kHz}$		200		μV
I_d	Quiescent drain current	$V_i = 27 \text{ V}$ $I_o = 0$	6	10	18	mA
$\frac{\Delta V_o}{\Delta T_{amb}}$	Temperature coefficient	$V_i = 24 \text{ V}$ $I_o = 5 \text{ mA}$ $C_L = 10 \mu\text{F}$ $T_{amb} = 0 \text{ to } 70 \text{ }^\circ\text{C}$		1.5		mV/ $^\circ\text{C}$
I_{sc}	Output short circuit current	$V_i = 27 \text{ V}$ $V_o = 0$		30	50	mA

TBA 625C

Fig. 1 - Typical output voltage vs output current

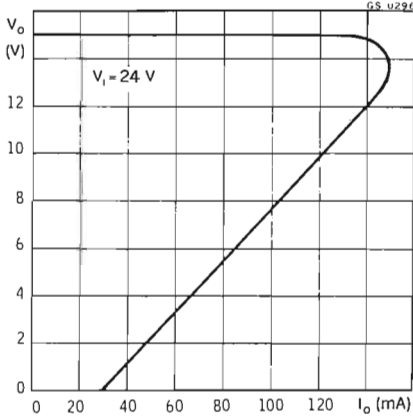


Fig. 2 - Power rating chart

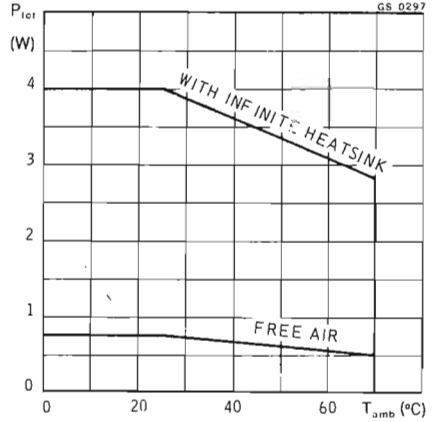


Fig. 3 - Maximum output current vs junction temperature

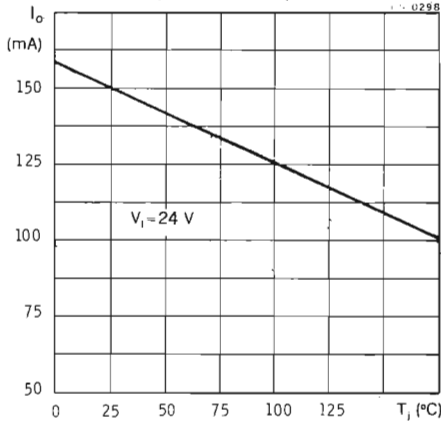
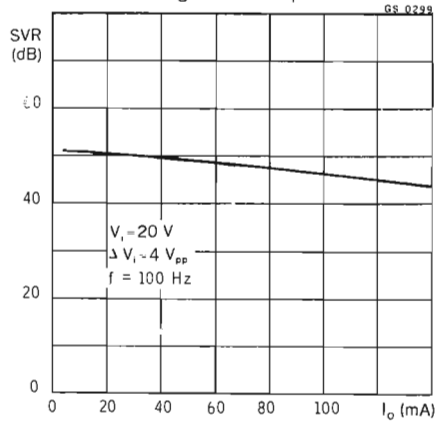


Fig. 4 - Typical ripple rejection vs regulated output current



TBA 625C

Fig. 5 - Typical ripple rejection vs frequency

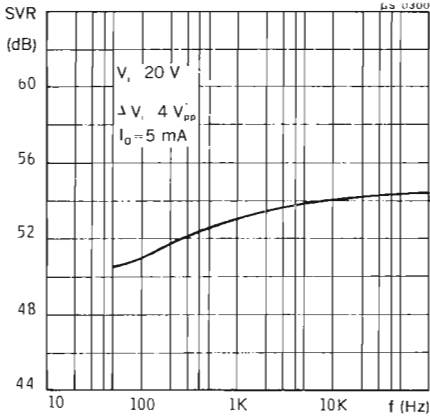


Fig. 6 - Maximum output current vs input voltage

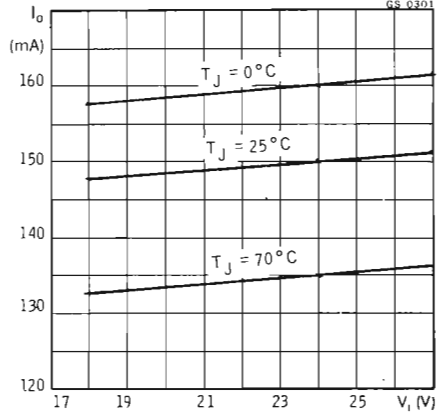


Fig. 7 - Typical short circuit output current vs input voltage

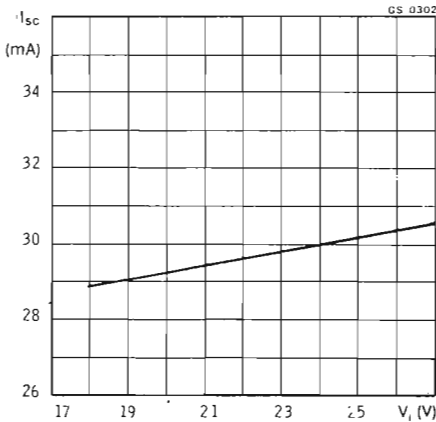
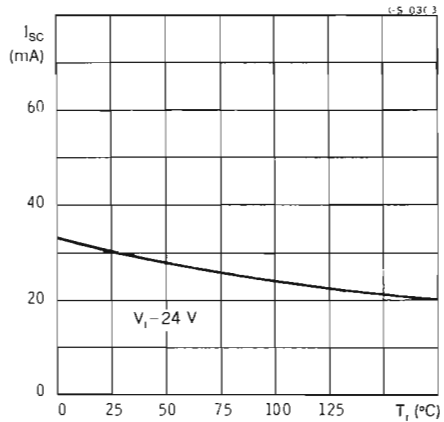


Fig. 8 - Typical short circuit output current vs junction temperature



TBA 625C

Fig. 9 - Typical dropout voltage vs output current

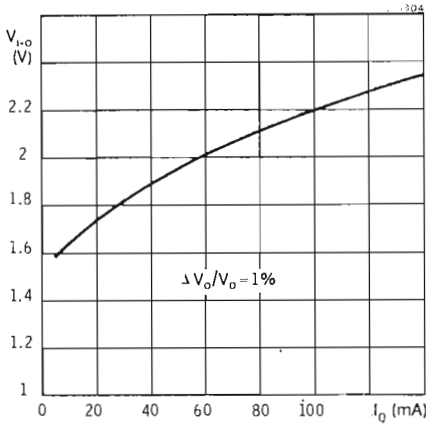


Fig. 10 - Typical quiescent drain current vs junction temperature

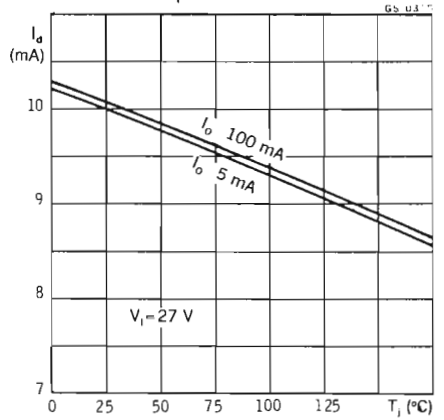


Fig. 11 - Typical quiescent drain current vs input voltage

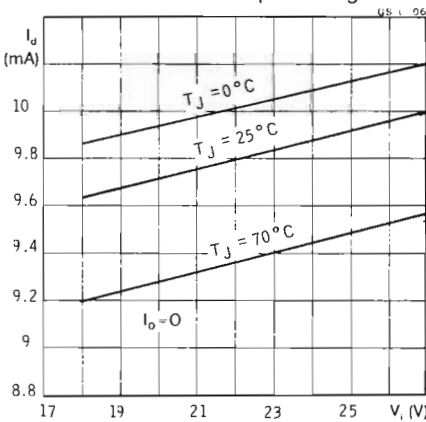
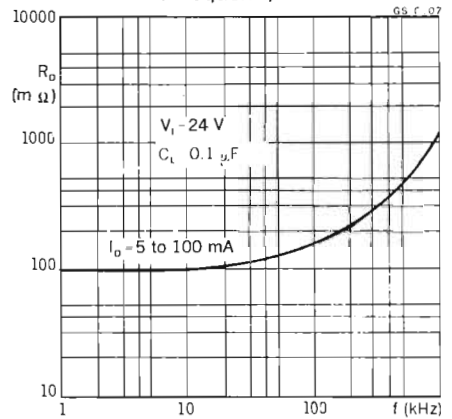
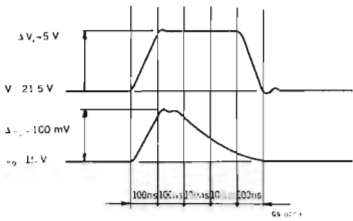


Fig. 12 - Typical output resistance vs frequency

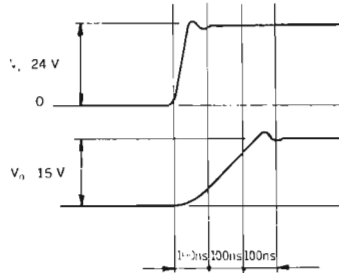


TBA 625C

Line transient response
($I_o = 5 \text{ mA}$)



Turn on time
($I_o = 100 \text{ mA}$)



TYPICAL APPLICATIONS

Fig. 13 - Positive output voltage regulator

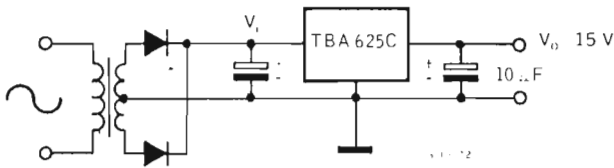
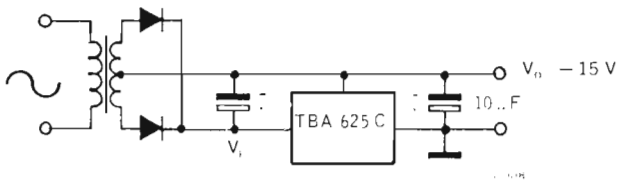
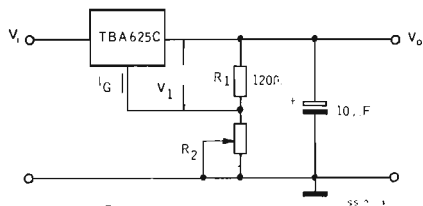


Fig. 14 - Negative output voltage regulator



TBA 625C

Fig. 15 - Adjustable output voltage regulator



$$V_0 = V_1 \left(1 + \frac{R_2}{R_1}\right) + I_G R_2$$

$V_1 = 26 \text{ V}$

$V_0 = 15 \text{ to } 17 \text{ V}$

$I_G > 80 \text{ mA}$

$R_0 \approx 100 \text{ m}\Omega$

$R_2 = \text{potentiometer } 0 \text{ to } 150 \Omega$

Typical adjustable output voltage vs output current

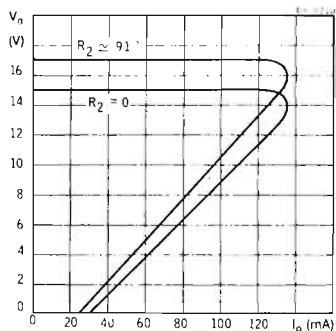
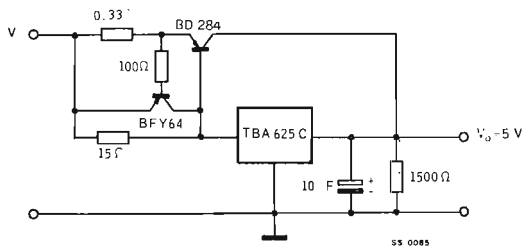


Fig. 16 - PNP current boost circuit



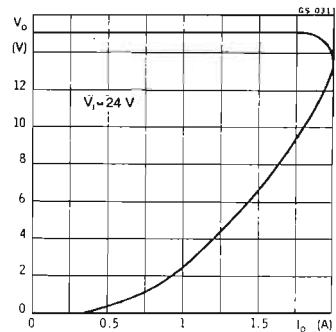
$V_1 = 24 \text{ V}$

$V_0 = 15 \text{ V}$

$I_0 = 2 \text{ A}$

$R_0 = 20 \text{ m}\Omega$

Typical output voltage vs output current



TBA 631

LINEAR INTEGRATED CIRCUIT

TV SOUND SECTION

- OUTPUT POWER 3 W (24 V - 16 Ω)
- LOW THRESHOLD LIMITING VOLTAGE
- LOW DISTORTION
- HIGH AM REJECTION
- SUPPLY VOLTAGE RANGING from 6 V to 18 V for IF STAGE and from 12 V to 27 V for POWER AMPLIFIER STAGE.

The TBA 631 is an integrated monolithic circuit in a 16-lead quad in-line plastic package with external heat-sink. It is especially designed as the whole sound section for TV receivers, from video preamplifier to load-speaker.

It combines the following functions: limiter amplifier, detector and audio power amplifier.

ABSOLUTE MAXIMUM RATINGS

V_{s1}	Supply voltage (IF stage)	18 V
V_{s2}	Supply voltage (Power stage)	27 V
I_o	Output peak current	1 A
P_{tot}	Power dissipation at $T_{amb} \leq 25^\circ\text{C}$	2 W
	at $T_{case} \leq 70^\circ\text{C}$	4.5 W
T_{stg}, T_j	Storage and junction temperature	-55 to 150 $^\circ\text{C}$

ORDERING NUMBERS:

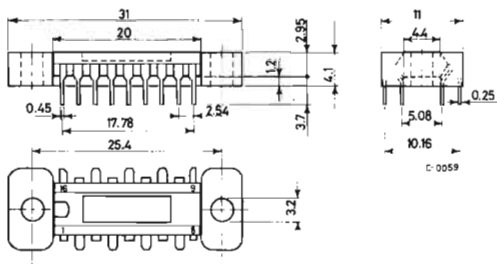
TBA 631 A72 (for quad in-line plastic package with spacer)

TBA 631 A51 (for quad in-line plastic package with external bar)

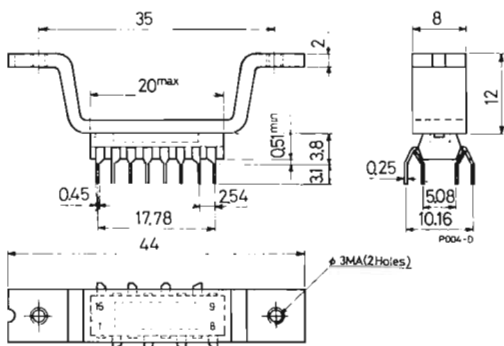
TBA 631 A61 (for quad in-line plastic package with inverted external bar)

TBA 631

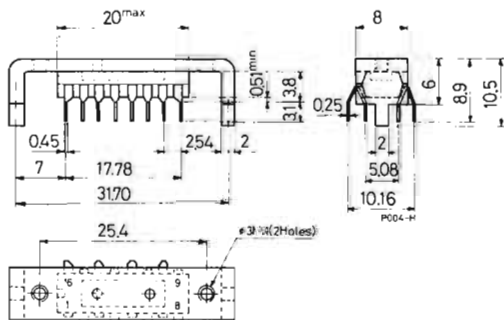
Quad in-line plastic package with spacer for TBA 631 A72 (see also "MOUNTING INSTRUCTIONS")



Quad in-line plastic package with external bar for TBA 631 A51



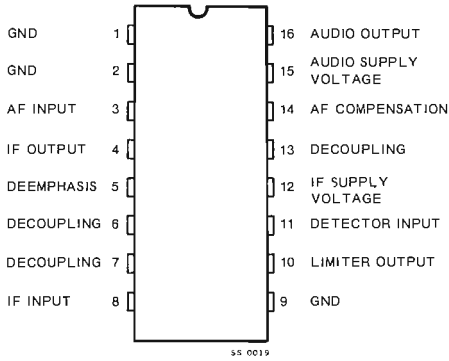
Quad in-line plastic package with inverted external bar for TBA 631 A61



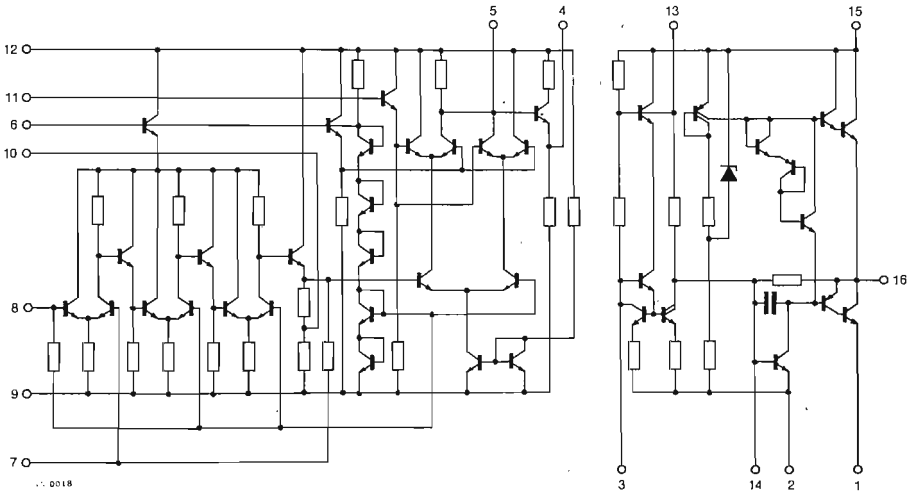
TBA 631

CONNECTION DIAGRAM

(top view)

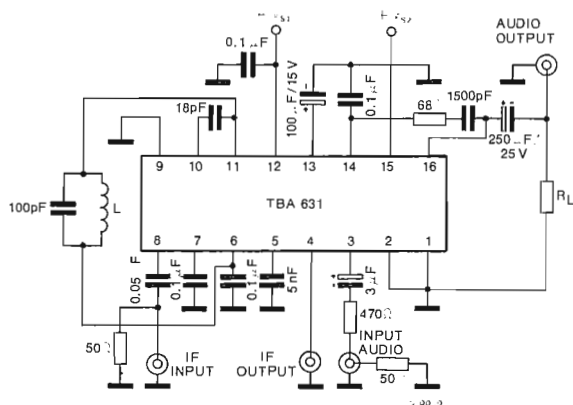


SCHEMATIC DIAGRAM



TBA 631

TEST CIRCUIT



THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	max	17	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	63	°C/W

ELECTRICAL CHARACTERISTICS

($T_{amb} = 25\text{ °C}$, $V_{s1} = 12\text{ V}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
-----------	-----------------	------	------	------	------

IF STAGE

I_d	Quiescent drain current		18		mA
$V_{i(threshold)}$	Input limiting voltage	$f = 5.5\text{ MHz}$ $f = 10.7\text{ MHz}$	100 230		μV μV
V_o	Recovered output voltage	$V_i = 10\text{ mV}$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 25\text{ kHz}$	1		V
d	Distortion	$V_i = 10\text{ mV}$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 25\text{ kHz}$	1.8		%

TBA 631

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
AMR Amplitude modulation rejection	$V_i = 10 \text{ mV}$ $f = 5.5 \text{ MHz}$ $f_m = 1 \text{ kHz}$ $\Delta f = \pm 25 \text{ kHz}$ $m = 0.3$		49		dB
Z_i Input impedance at pin 8	$f = 5.5 \text{ MHz}$ $f = 10.7 \text{ MHz}$		2.5 2		$k\Omega$ $k\Omega$
Z_o Output impedance at pin 4	$f = 1 \text{ kHz}$		100		Ω

AUDIO POWER STAGE ($R_L = 16 \Omega$ unless otherwise specified)

I_d Quiescent drain current	$V_{s2} = 18 \text{ V}$ $V_{s2} = 24 \text{ V}$		9 12		mA mA
P_o Output power	$d = 10\%$ $f = 1 \text{ kHz}$ $V_{s2} = 18 \text{ V}$ $V_{s2} = 24 \text{ V}$ $d = 1\%$ $f = 1 \text{ kHz}$ $G_v = 30 \text{ dB}$ $V_{s2} = 18 \text{ V}$ $V_{s2} = 24 \text{ V}$		1.8 3 1.4 2.25		W W W W
d Distortion	$P_o = 50 \text{ mW}$ $f = 1 \text{ kHz}$ $G_v = 30 \text{ dB}$ $V_{s2} = 18 \text{ V}$ $V_{s2} = 24 \text{ V}$		0.3 0.2		% %
I_d Drain current	$P_o = 3 \text{ W}$ $V_{s2} = 24 \text{ V}$ $P_o = 1.8 \text{ W}$ $V_{s2} = 18 \text{ V}$		200 165		mA mA

TBA 631

Fig. 1 - Typical relative audio output voltage (pin 4) vs RF input voltage

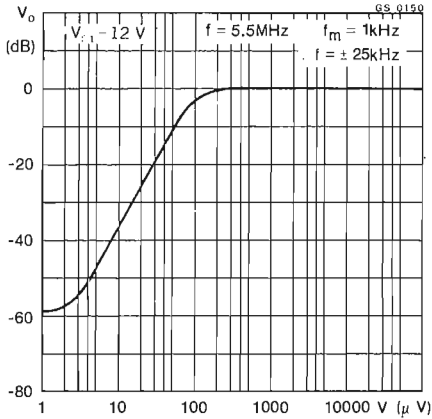


Fig. 2 - Typical AM rejection vs RF input voltage

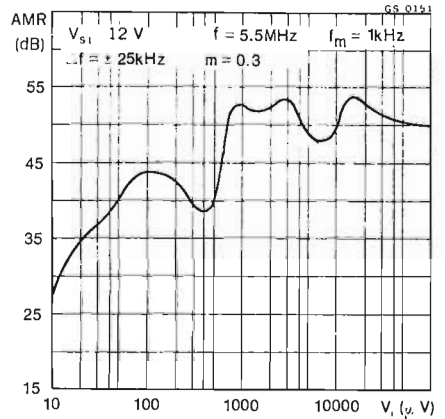


Fig. 3 - Typical output power of the AF amplifier vs supply voltage

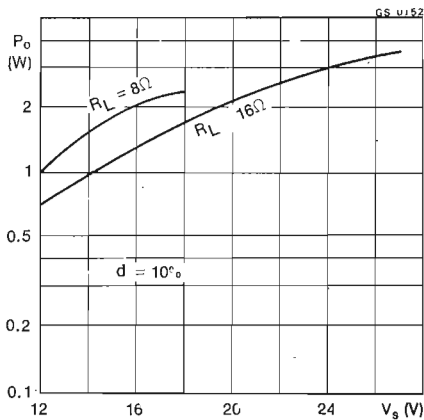


Fig. 4 - Typical distortion vs output power

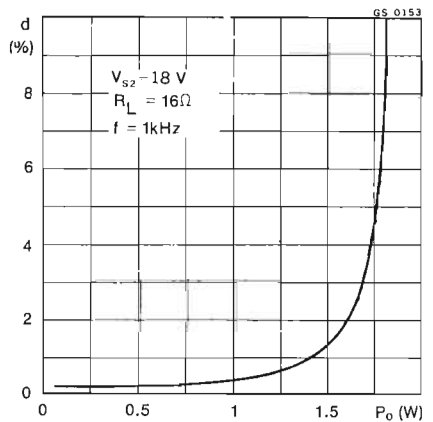


Fig. 5 - Typical distortion vs output power

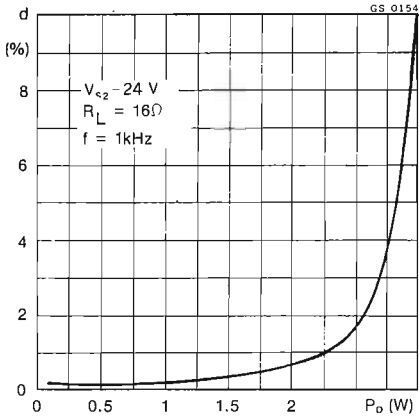


Fig. 6 - Typical relative voltage gain of the AF amplifiers vs frequency (see test circuit)

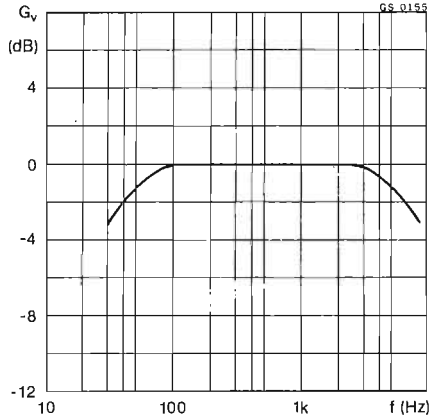


Fig. 7 - Typical output power of the AF amplifier vs input voltage (pin 3)

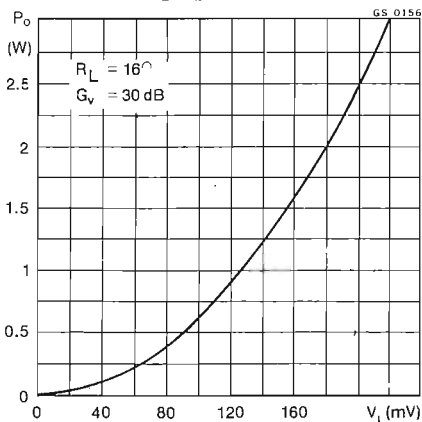
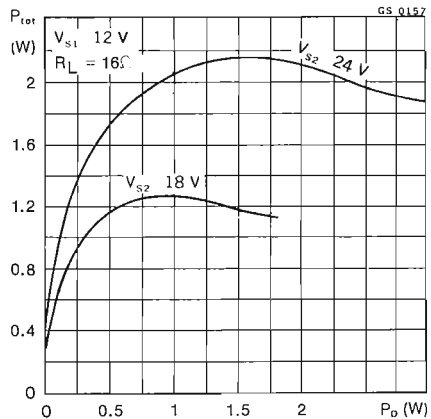


Fig. 8 - Typical power dissipation of TBA 631 vs output power



TBA 631

Fig. 9 - Typical efficiency vs output power

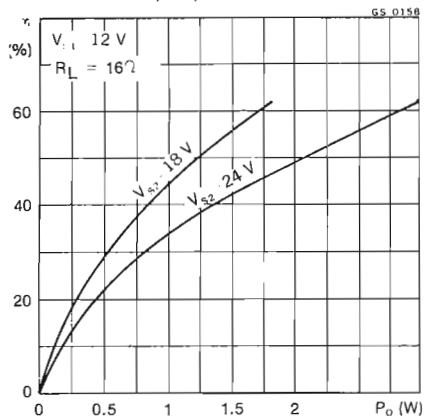


Fig. 10 - Typical drain current of the AF amplifier vs output power

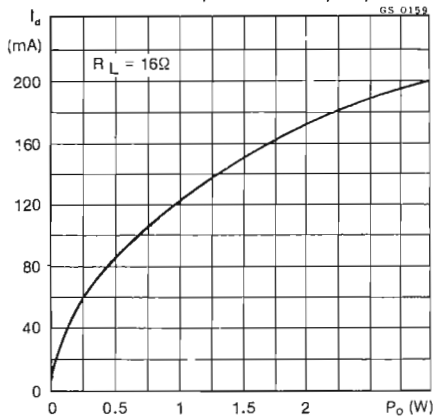


Fig. 11 - Maximum power dissipation vs AF amplifier supply voltage

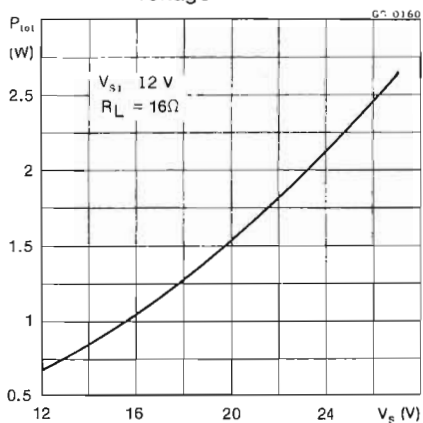
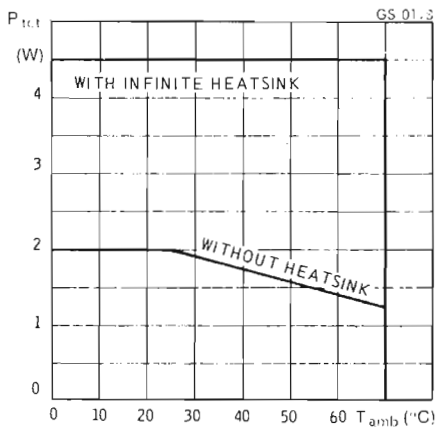


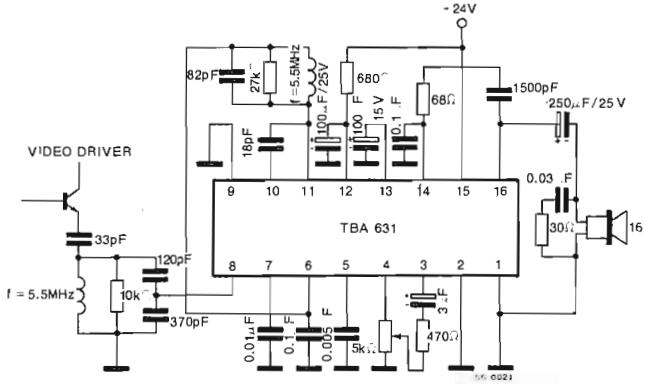
Fig. 12 - Power rating chart



TBA 631

TYPICAL APPLICATION

Sound section of a TV receiver.

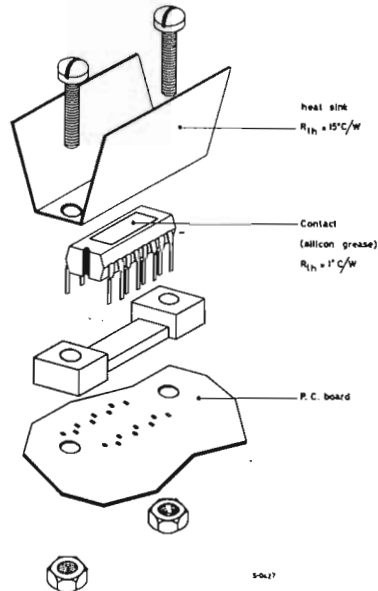


MOUNTING INSTRUCTIONS

Heat-sinking with spacer.

Fig. 13 - Shows a method of mounting the TBA 631 with the spacer, satisfactory both mechanically and from the point of view of heat dissipation. Better thermal contact between package and heat-sink can be obtained by using a small quantity of silicon grease. For heat dissipation the desired thermal resistance is obtained by fixing the elements shown to a heat-sink of suitable dimensions.

Fig. 13



TBA 631

MOUNTING INSTRUCTIONS (continued)

Heat-sinking with external bar

Power dissipation can be achieved by means of an additional external heat-sink fixed with two screws (both packages) or by soldering the pins of the external bar to suitable copper areas on the p.c. board (TBA 631 A61).

- A. In the former case, the thermal resistance case-ambient of the added heat-sink can be calculated as follows:

$$R_{th} = \frac{(T_{jmax} - T_{amb}) - P_{tot} \cdot R_{th j-case}}{P_{tot}}$$

where:

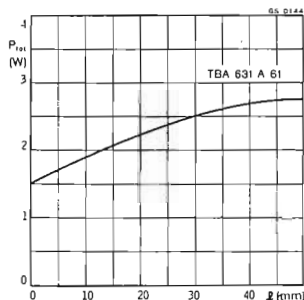
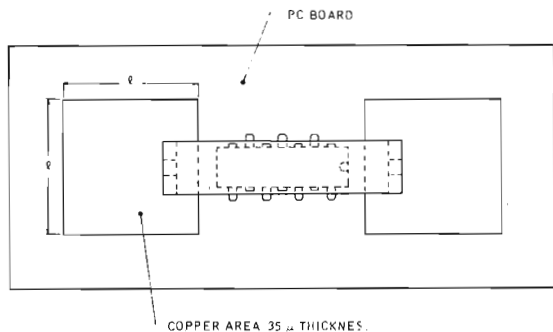
T_{jmax} = Max junction temperature

T_{amb} = Ambient temperature

P_{tot} = Power dissipation

$R_{th j-case}$ = Thermal resistance junction-case

- B. If copper areas on the p.c. board are used (TBA 631 A61) the diagrams enclosed give the maximum power dissipation as a function of copper area, with copper thickness 35μ and ambient temperature 55°C .



TBA 641A

LINEAR INTEGRATED CIRCUIT

AUDIO AMPLIFIER

- OUTPUT POWER 2.2 W (9 V - 4 Ω)
- LOW DISTORTION
- LOW QUIESCENT CURRENT
- SELF CENTERING BIAS
- HIGH INPUT IMPEDANCE

The TBA 641 A is a monolithic integrated circuit in a 14-lead quad in-line plastic package. It is particularly designed for use as audio power amplifier in portable radio receivers, tape recorders, record players and in industrial applications which require high output power, low distortion and high reliability performance.

Special features of the circuit include a low quiescent current, self centering bias operation at supply voltage ranging from 6 V to 12 V, direct coupling of the input. The circuit requires a minimum of external components.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	12	V
V_i	Input voltage	-0.5 to $+V_s$	V
I_o	Output peak current	2	A
P_{tot}^*	Power dissipation at $T_{amb} \approx 25^\circ\text{C}$	1.5	W
T_{stg}	Storage temperature	-40 to 150	$^\circ\text{C}$
T_j	Junction temperature	150	$^\circ\text{C}$

* P_{tot} value refers to TBA 641 A12

ORDERING NUMBERS:

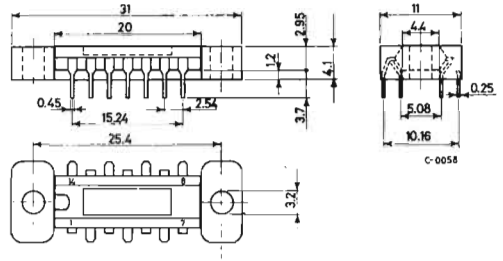
TBA 641 A72 for quad in-line plastic package with spacer

TBA 641 A12 for quad in-line plastic package

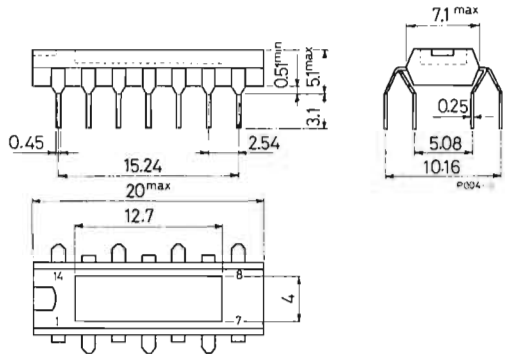
TBA 641A

MECHANICAL DATA (Dimensions in mm)

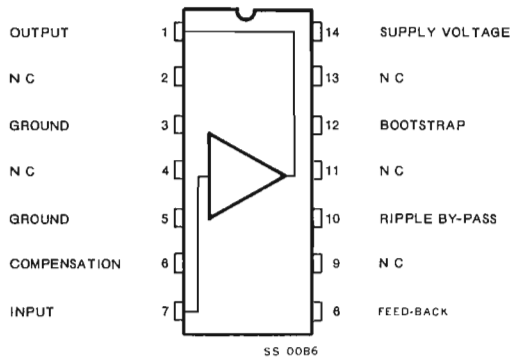
Quad in-line plastic package with spacer for TBA 641 A72 (see also "MOUNTING INSTRUCTIONS")



Quad in-line plastic package for TBA 641 A12

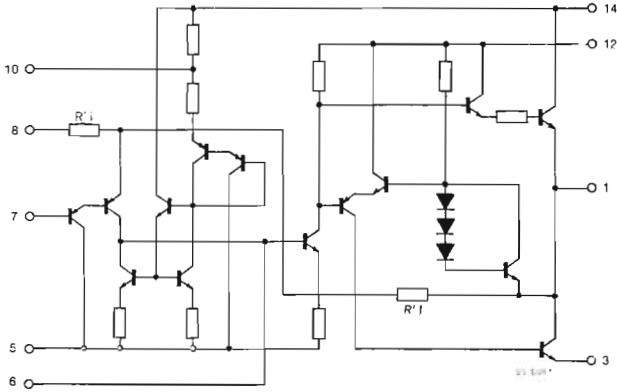


CONNECTION DIAGRAM (top view)

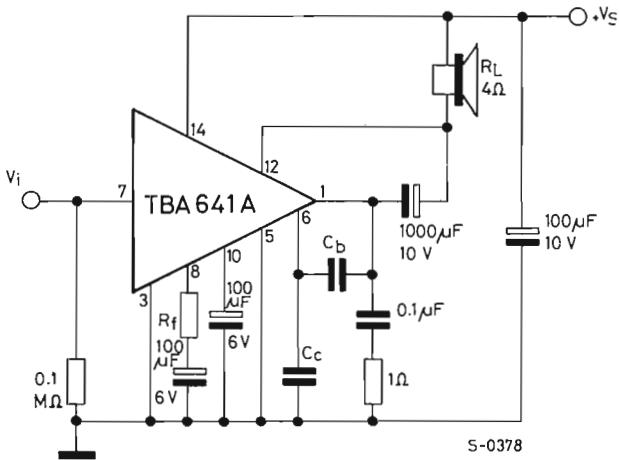


TBA 641A

SCHEMATIC DIAGRAM



TEST AND APPLICATION CIRCUIT



TBA 641A

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	typ	13	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	typ	83	°C/W

ELECTRICAL CHARACTERISTICS

(See test circuit; $T_{amb} = 25\text{ °C}$, $V_s = 9\text{ V}$ and $R_L = 4\ \Omega$ unless otherwise specified)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
V_o	Quiescent output voltage (pin 1)		4	4.5	5	V
I_d	Total quiescent drain current	$P_o = 0$		8	18	mA
I_d	Quiescent drain current of output transistors	$P_o = 0$		6		mA
I_d	Drain current	$P_o = 2.2\text{ W}$		340		mA
I_b	Bias current (pin 7)			100		nA
P_o	Output power	$d = 10\%$ $f = 1\text{ kHz}$ $G_v = 46\text{ dB}$	1.8	2.2		W
R'_f	Internal feedback resistance	See schematic diagram		7		k Ω
R'_i	Internal feedback resistance	See schematic diagram		35		Ω
Z_i	Input impedance (pin 7)	$f = 1\text{ kHz}$ $G_v = 46\text{ dB}$		3		M Ω
d	Distortion	$f = 1\text{ kHz}$ $G_v = 46\text{ dB}$ $P_o = 50\text{ mW}$ $P_o = 1\text{ W}$		0.6	0.6	% %
G_v	Voltage gain	$R_f = 0$		46		dB
e_N	Input noise voltage	$R_s = 22\text{ k}\Omega$ $B = 10\text{ kHz}$		2.5		μV

TBA 641A

Fig. 1 - Typical output power vs supply voltage

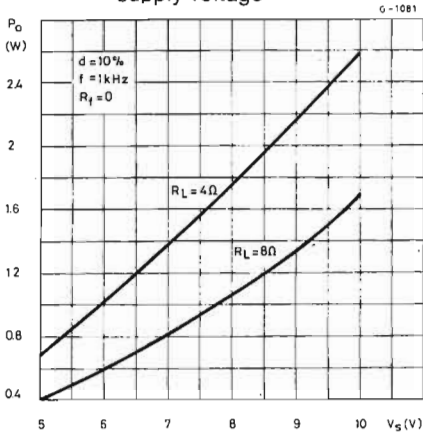


Fig. 2 - Typical distortion vs output power

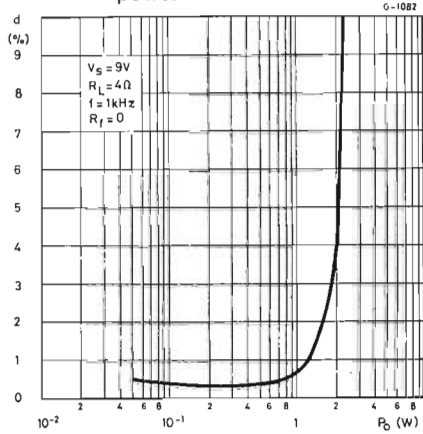


Fig. 3 - Typical voltage gain vs feedback resistance (R_f)

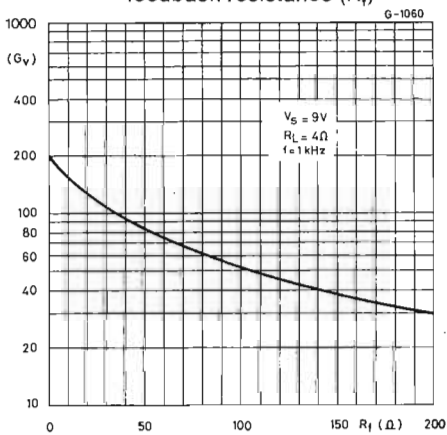
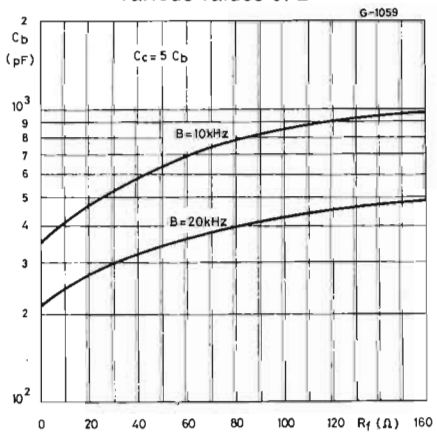


Fig. 4 - Typical value of C_b vs R_f for various values of B



TBA 641A

Fig. 5 - Typical output power vs input voltage

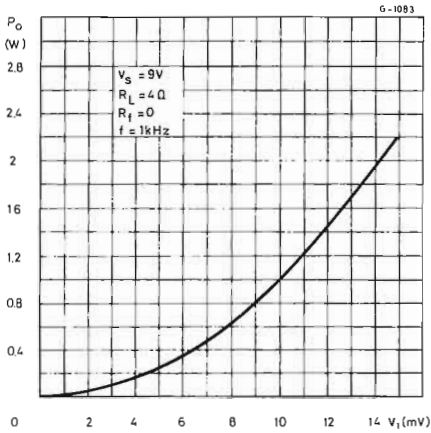


Fig. 6 - Typical power dissipation and efficiency vs output power

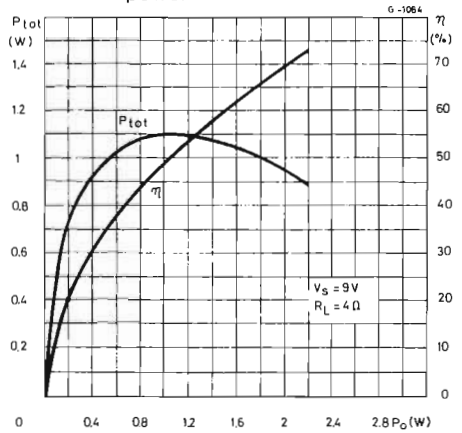


Fig. 7 - Typical drain current vs output power

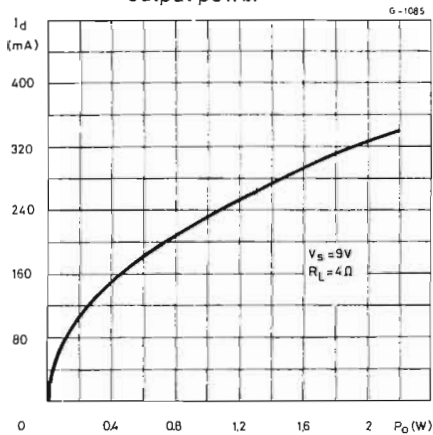
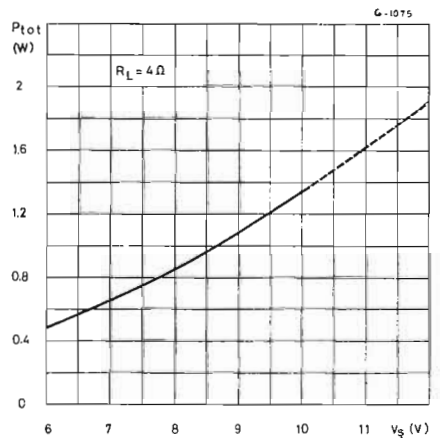


Fig. 8 - Maximum power dissipation *



* The dotted line refers to TBA 641 A72 with additional heat-sink.

TBA 641A

Fig. 9 - Power rating chart

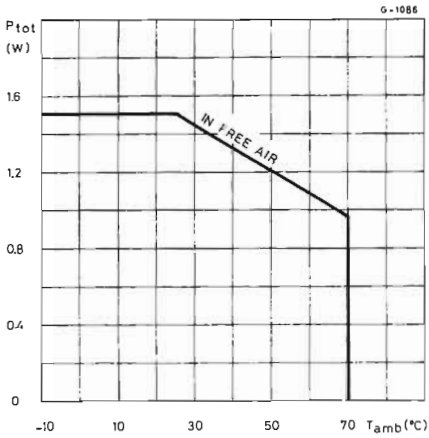


Fig. 10 - Typical quiescent drain current vs supply voltage

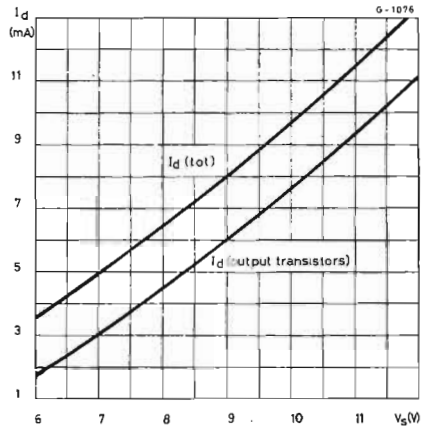


Fig. 11 - Typical quiescent drain current vs ambient temperature

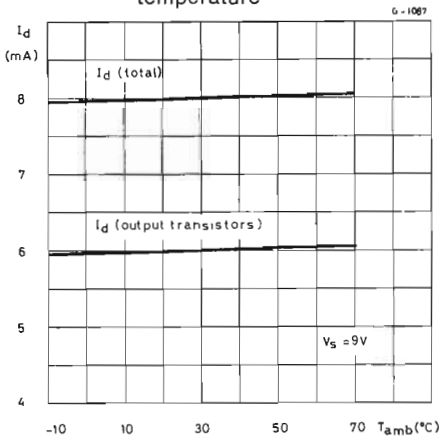
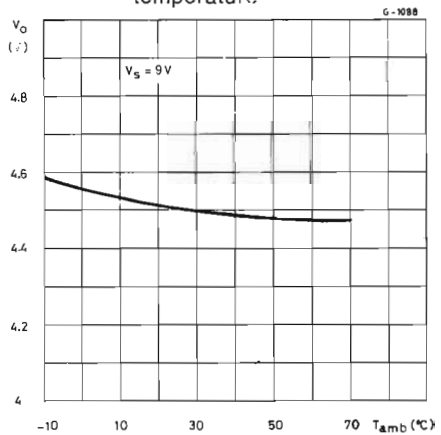
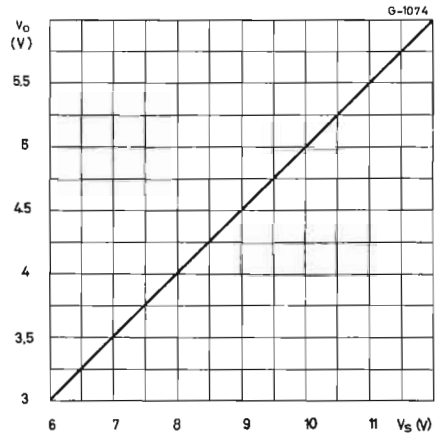


Fig. 12 - Typical quiescent output voltage vs ambient temperature



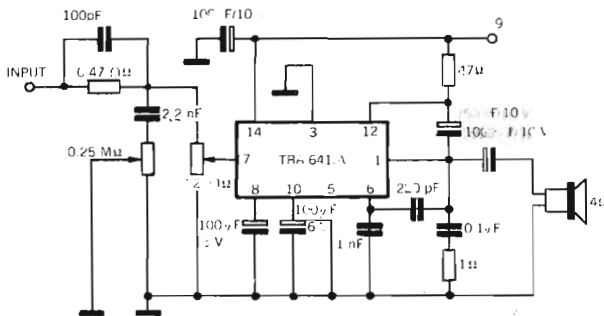
TBA 641A

Fig. 13 - Typical quiescent output voltage vs supply voltage



TYPICAL APPLICATION

Fig. 14 - Portable record-player amplifier

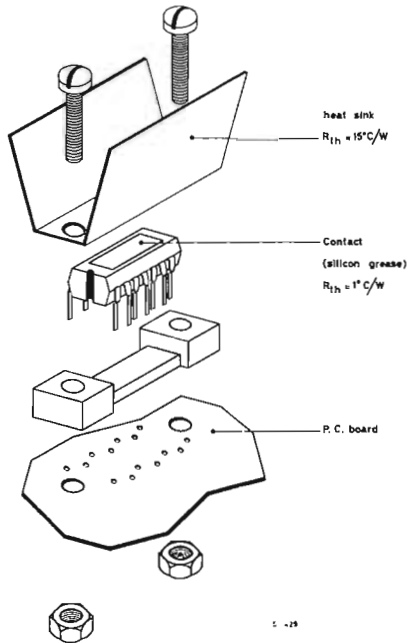


TBA 641A

MOUNTING INSTRUCTIONS

Fig. 15 shows a method of mounting the TBA 641 A with the spacer, satisfactory both mechanically and from the point of view of heat dissipation. Better thermal contact between package and heat-sink can be obtained by using a small quantity of silicon grease. For heat dissipation the desired thermal resistance is obtained by fixing the elements shown to a heat-sink of suitable dimensions.

Fig. 15



1 - 25

TBA 641B

LINEAR INTEGRATED CIRCUIT

AUDIO AMPLIFIER

- OUTPUT POWER 4.5 W (14 V - 4 Ω)
- LOW DISTORTION
- LOW QUIESCENT CURRENT
- HIGH INPUT IMPEDANCE

The TBA 641 B is a monolithic integrated circuit in a 14-lead quad in-line plastic package. It is particularly designed for use as audio power amplifier in radio and television receivers, and in industrial applications which require high output power, low distortion and high reliability performance. Special features of the circuit include a low quiescent current, self centering bias for operation at supply voltage ranging from 6 V to 16 V, direct coupling of the input. The circuit requires a minimum of external components.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage (no signal)	18	V
V_s	Operating supply voltage	16	V
V_i	Input voltage	-0.5 to + V_s	V
I_o	Peak output current	2.5	A
P_{tot}^*	Power dissipation at $T_{amb} = 25^\circ\text{C}$	2.3	W
	$T_{amb} = 70^\circ\text{C}$	1.45	W
	$T_{case} = 70^\circ\text{C}$	6	W
T_{stg}, T_j	Storage and junction temperature	-40 to 150	$^\circ\text{C}$

* P_{tot} values refer to TBA 641 BX1 and TBA 641 B11.

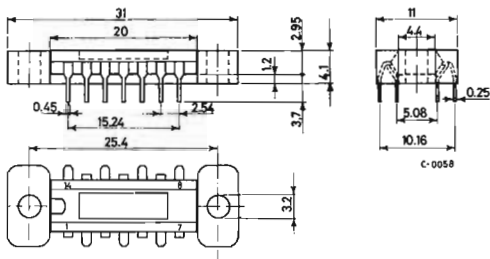
ORDERING NUMBERS:

TBA 641 B72 for quad in-line plastic package with spacer
TBA 641 BX1 for quad in-line plastic package with external bar
TBA 641 B11 for quad in-line plastic package with inverted external bar

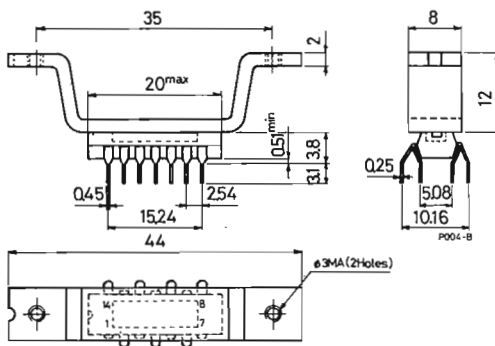
TBA 641B

MECHANICAL DATA (Dimensions in mm)

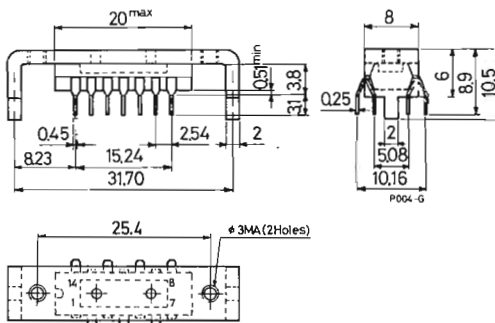
Quad in-line plastic package with spacer for TBA 641 B72 (see also "MOUNTING INSTRUCTIONS")



Quad in-line plastic package with external bar for TBA 641 BX1

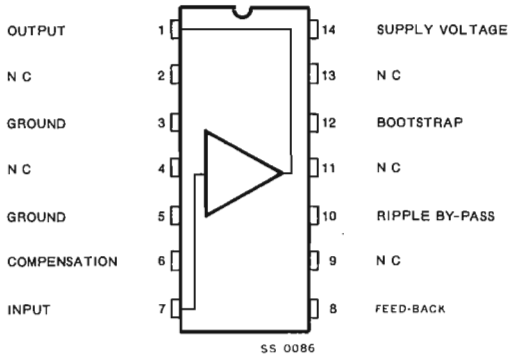


Quad in-line plastic package with inverted external bar for TBA 641 B11

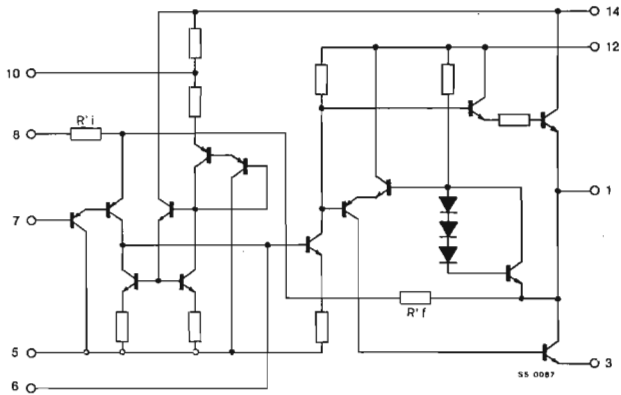


TBA 641B

CONNECTION DIAGRAM

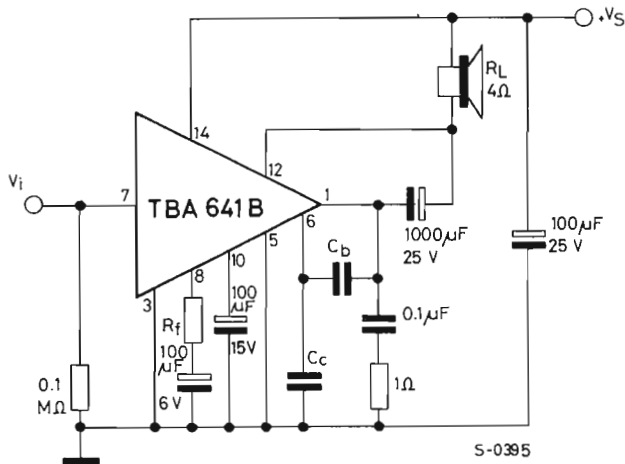


SCHEMATIC DIAGRAM



TBA 641B

TEST AND APPLICATION CIRCUIT



THERMAL DATA

$R_{th \ j-case}$	Thermal resistance junction-case	typ	13	$^{\circ}C/W$
$R_{th \ j-amb}$	Thermal resistance junction-ambient	typ	55	$^{\circ}C/W$

ELECTRICAL CHARACTERISTICS

(See test circuit; $T_{amb} = 25^{\circ}C$, $V_s = 14V$ and $R_L = 4\Omega$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_o	Quiescent output voltage (pin 1)	6.5	7	8	V
I_d	Total quiescent drain current	$P_o = 0$	16	32	mA
I_d	Quiescent drain current of output transistors	$P_o = 0$	13		mA
I_d	Drain current	$P_o = 4.5W$	485		mA
I_b	Bias current (pin 7)		250		nA

TBA 641B

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
P_o Output power	$d = 10\%$ $f = 1 \text{ kHz}$ $G_v = 46 \text{ dB}$	4	4.5		W
R'_f Internal feedback resistance	See schematic diagram		7		$k\Omega$
R'_i Internal feedback resistance	See schematic diagram		35		Ω
Z_i Input impedance (pin 7)	$f = 1 \text{ kHz}$ $G_v = 46 \text{ dB}$		3		$M\Omega$
d Distortion	$f = 1 \text{ kHz}$ $G_v = 46 \text{ dB}$ $P_o = 50 \text{ mW}$ $P_o = 2 \text{ W}$		0.3	0.8	% %
G_v Voltage gain	$R_f = 0$		46		dB
e_N Input noise voltage	$R_s = 22 \text{ k}\Omega$ $B = 10 \text{ kHz}$		3.4		μV

Fig. 1 - Typical output power vs supply voltage

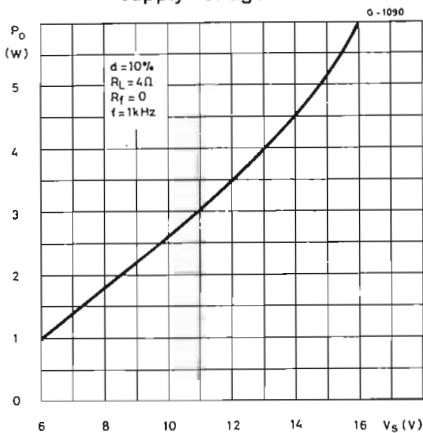
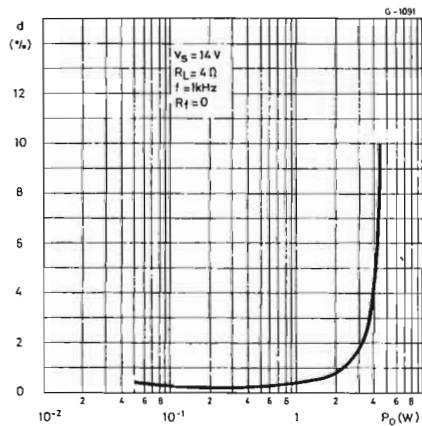


Fig. 2 - Typical distortion vs output power



TBA 641B

Fig. 3 - Typical voltage gain vs feedback resistance (R_f)

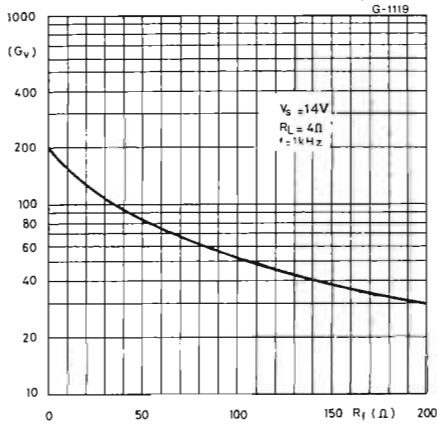


Fig. 4 - Typical value of C_b vs R_f for various values of B

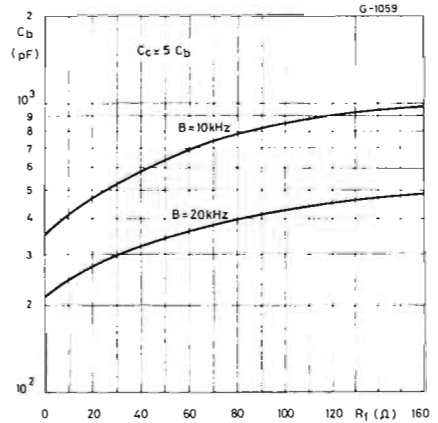


Fig. 5 - Typical output power vs input voltage

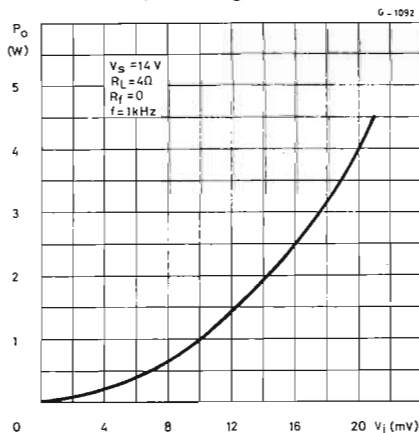
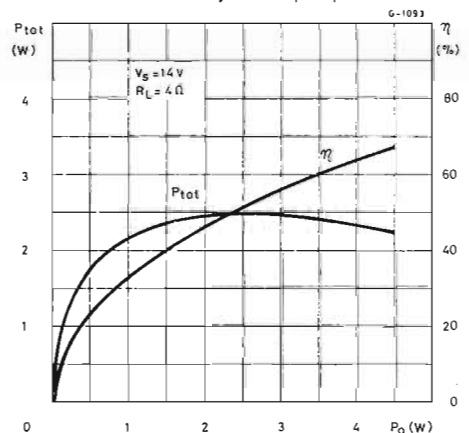


Fig. 6 - Typical power dissipation and efficiency vs output power



TBA 641B

Fig. 7 - Typical drain current vs output power

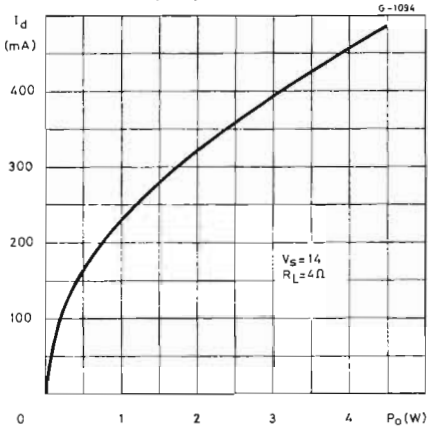


Fig. 8 - Maximum power dissipation vs supply voltage

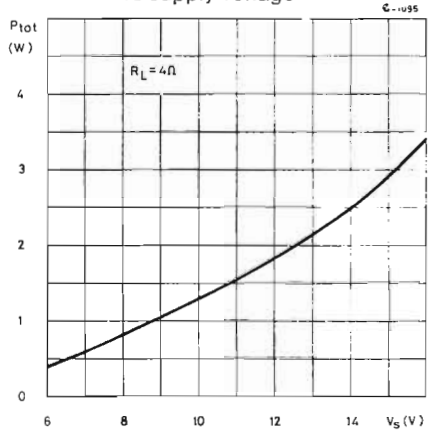


Fig. 9 - Power rating chart

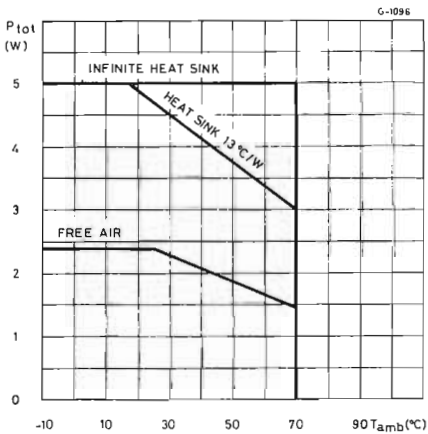
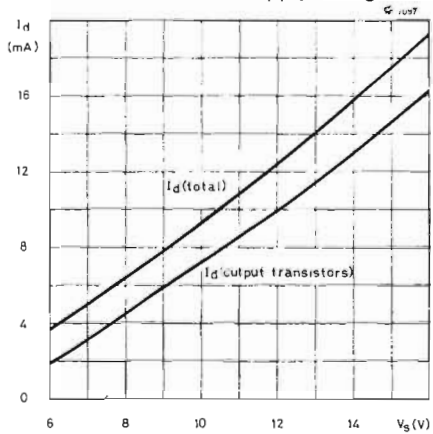


Fig. 10 - Typical quiescent drain current vs supply voltage



TBA 641B

Fig. 11 - Typical quiescent drain current vs ambient temperature

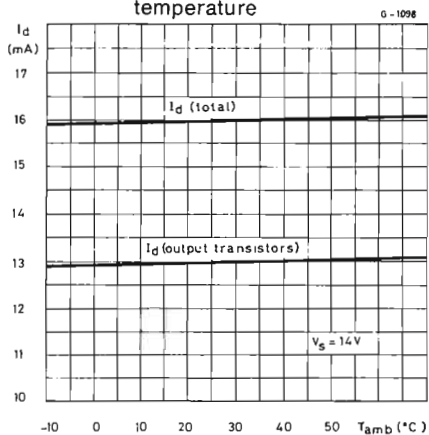


Fig. 12 - Typical quiescent output voltage vs ambient temperature

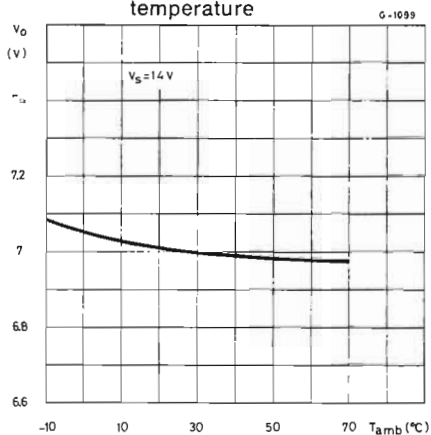
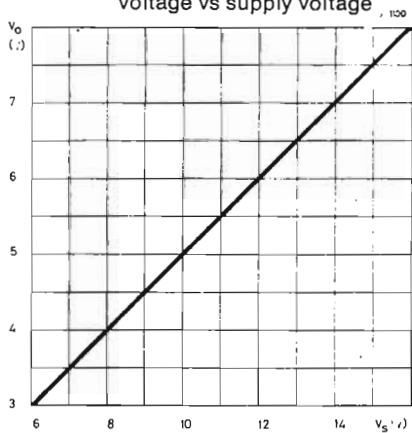
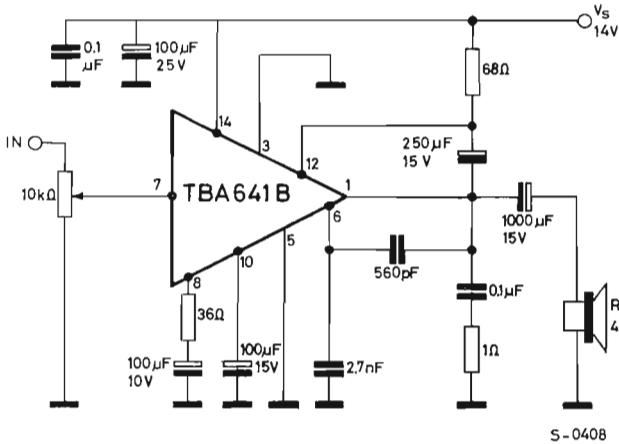


Fig. 13 - Typical quiescent output voltage vs supply voltage



TBA 641B

TYPICAL APPLICATION



MOUNTING INSTRUCTIONS

Fig. 14

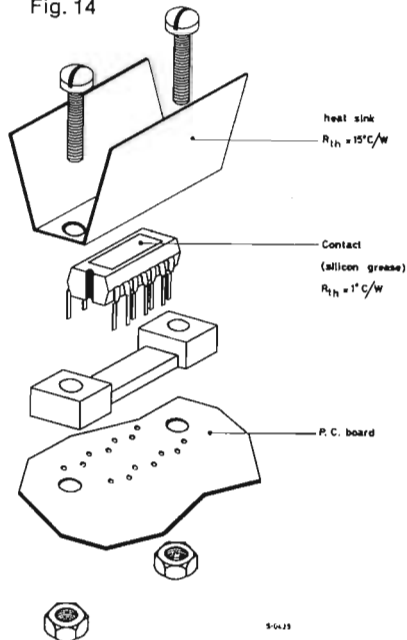


Fig. 14 shows a method of mounting the TBA 641 B with the spacer, satisfactory both mechanically and from the point of view of heat dissipation. Better thermal contact between package and heat-sink can be obtained by using a small quantity of silicon grease. For heat dissipation the desired thermal resistance is obtained by fixing the elements shown to a heat-sink of suitable dimensions.

TBA 641B

MOUNTING INSTRUCTIONS (continued)

Power dissipation can be achieved by means of an additional external heat-sink fixed with two screws (both packages) or by soldering the pins of the external bar to suitable copper areas on the p.c. board (TBA 641 B11)

- A. In the former case, the thermal resistance case-ambient of the added heat-sink can be calculated as follows:

$$R_{th} = \frac{T_{jmax} - T_{amb} - P_{tot} \cdot R_{th j-case}}{P_{tot}}$$

where:

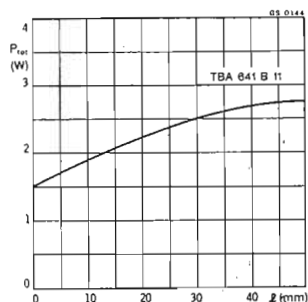
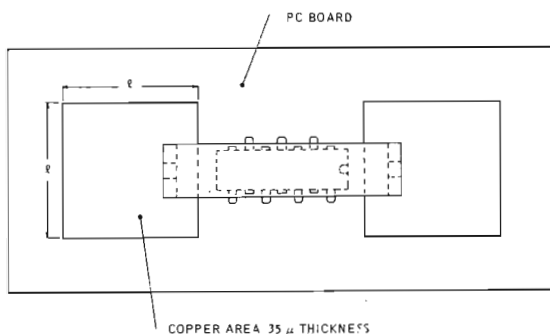
T_{jmax} = Max junction temperature

T_{amb} = Ambient temperature

P_{tot} = Power dissipation

$R_{th j-case}$ = Thermal resistance junction-case

- B. If copper areas on the p.c. board are used (TBA 641 B11) the diagrams enclosed give the maximum power dissipation as a function of copper area, with copper thickness 35μ and ambient temperature 55°C .



TBA 651

LINEAR INTEGRATED CIRCUIT

TUNER AND IF AMPLIFIER FOR AM RADIO

- AUDIO OUTPUT VOLTAGE 0.6 V
- LOW NOISE and HIGH GAIN
- WIDE VOLTAGE SUPPLY RANGE 4.5 V to 18 V
- HIGH SIGNAL HANDLING CAPABILITY 1 V

The TBA 651 is a monolithic integrated circuit in a 16-lead dual in-line plastic package which processes the high frequency signal from antenna to detector in AM receivers. It is particularly intended for car radios and high quality radio receivers.

The TBA 651 consists of five stages: RF amplifier, mixer, oscillator, IF amplifier and AGC control. It features wide voltage supply range (4.5 to 18 V), high gain, low noise and high sensitivity.

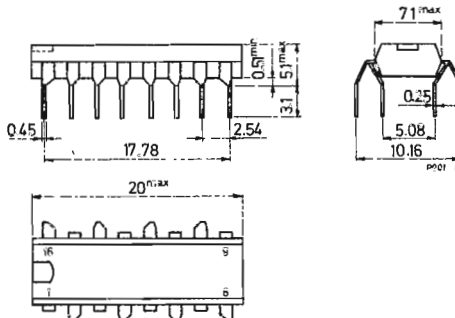
ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	18 V
P_{tot}	Power dissipation at $T_{amb} \leq 80^\circ\text{C}$	250 mW
T_{stg}	Storage temperature	-55 to 150 °C
T_{op}	Operating temperature	-20 to 80 °C

ORDERING NUMBER: TBA 651

MECHANICAL DATA

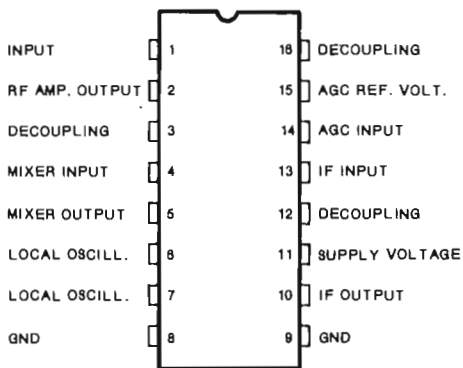
Dimensions in mm



TBA 651

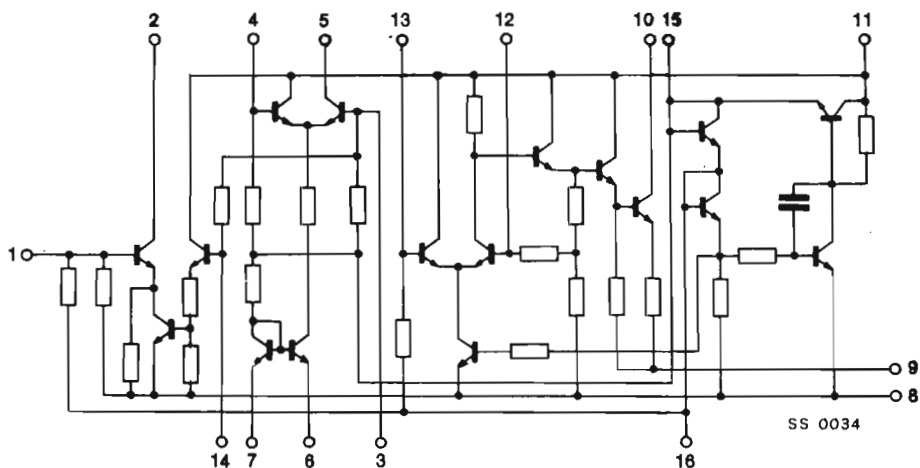
CONNECTION DIAGRAM

(top view)



SS 0033

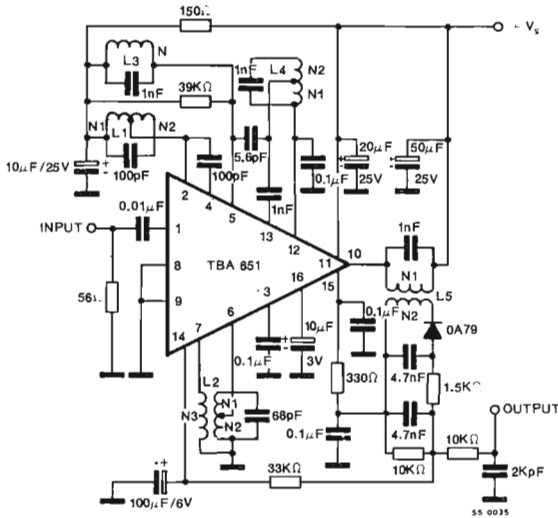
SCHEMATIC DIAGRAM



SS 0034

TBA 651

TEST CIRCUIT (f = 1.6 MHz)



ELECTRICAL CHARACTERISTICS

($T_{amb} = 25^\circ\text{C}$, $V_s = 12\text{V}$ unless otherwise specified)

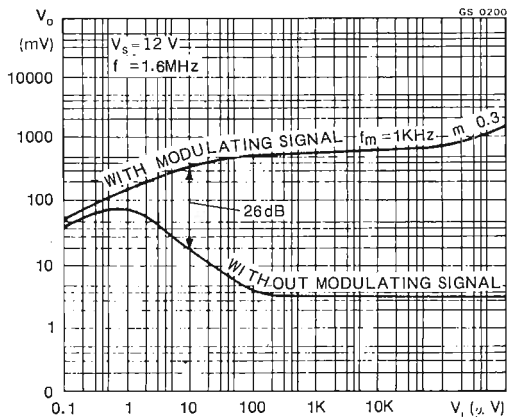
Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_d	Quiescent drain current		11.5		mA
V_i	Input voltage at pin 1 signal to noise ratio = 26 dB		10		μV
		$d = 5\%$ $f = 1.6\text{ MHz}$ $f_m = 1\text{ kHz}$ $m = 0.8$		100	
V_o	Recovered audio output voltage	$f = 1.6\text{ MHz}$ $f_m = 1\text{ kHz}$ $m = 0.3$ $V_i = 100\ \mu\text{V}$ $V_i = 1.5\ \mu\text{V}$	0.5	180	V mV

TBA 651

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_i	Signal handling capability at pin 1		1		V
	AGC range	for 10 dB expansion of output voltage	80		dB
R_i	rf amplifier input resistance at pin 1	$f = 1.6 \text{ MHz}$	1.4		$k\Omega$
R_i	Mixer input resistance at pin 4	$f = 1.6 \text{ MHz}$	2.5		$k\Omega$
R_i	IF amplifier input resistance at pin 13	$f = 455 \text{ kHz}$	4		$k\Omega$

Fig. 1 - Typical output voltage vs input voltage



TBA 780

LINEAR INTEGRATED CIRCUIT

WIDE-BAND AMPLIFIER, FM DETECTOR, AUDIO PREAMPLIFIER/DRIVER

The TBA 780 provides, in a single monolithic silicon chip, a major subsystem for the sound section of TV receivers in a 14-lead quad in-line or dual in-line plastic package. As shown in the schematic diagram the TBA 780 contains a multistage wide-band IF amplifier/limiter section, active filter, an FM-detector stage, electronic attenuator, a Zener diode regulated power supply section and AF amplifier section specifically designed to directly drive an NPN power transistor or high-transconductance tube.

In the TBA 780, the demodulation can be effected by a single tuned discriminator coil (differential peak detector).

Because of the circuit being so inclusive, a minimum number of external components is required. A particular feature of the TBA 780 is the electronic attenuator, which performs the conventional volume control function.

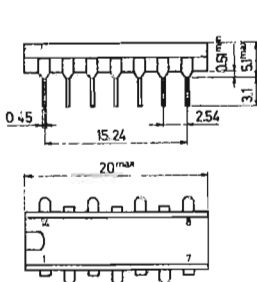
ABSOLUTE MAXIMUM RATINGS

→ I_s	Supply current (pin 5)	50 mA
→ I_o	Output current (pin 12)	6 mA
V_i	Input-signal voltage (between pins 1 and 2)	± 3 V
P_{tot}	Total power dissipation: at $T_{amb} \leq 25^\circ\text{C}$	850 mW
→ T_{stg}	Storage temperature	-25 to 150 °C
→ T_{op}	Operating temperature	0 to 85 °C

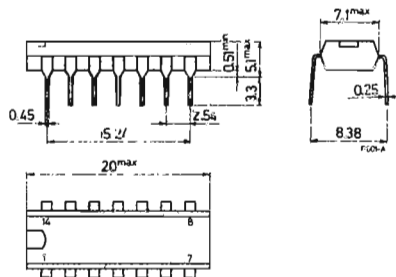
ORDERING NUMBERS: TBA 780 X2 for quad in-line plastic package
TBA 780 X7 for dual in-line plastic package

MECHANICAL DATA

Dimensions in mm



TBA 780 X2



TBA 780 X7

TBA 780

THERMAL DATA

→ $R_{th\ j-amb}$	Thermal resistance junction-ambient	typ.	150 °C/W
-------------------	-------------------------------------	------	----------

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25\text{ °C}$, DC volume control $P_2 = 0$ and $V_S = +30\text{ V}$ applied to terminal 5 through a $620\ \Omega$ resistor, unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
I_S Supply current	$V_S = 9\text{ V}$ (applied direct. to pin 5)	10	16	24	mA	—
$V_{i(threshold)}$ Input limiting voltage (pin 2)	$f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 50\text{ kHz}$		200	400	μV	—
V_o Recovered audio voltage (pin 8)	$V_i = 100\text{ mV}$ $f = 5.5\text{ MHz}$ $f_m = 1\text{ kHz}$ $\Delta f = \pm 50\text{ kHz}$	0.5	0.75		V_{rms}	3
d Distortion (pin 8)			0.9	2	%	
V_o Audio output voltage (pin 12)	d = 5% f = 1 kHz	2	2.5		V_{rms}	4
→ V_o DC output voltage (pin 12)		8.5		11.75	V	—
DC volume control range	$P_2 = \infty$	60	80		dB	3
Max. play-through voltage			0.075	1	mV	
R_i Input resistance (pin 2)	f = 5.5 MHz		17		k Ω	—
R_o Output resistance (pin 9)			3.25		k Ω	

TBA 780

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
R_o Output resistance (pin 12)	$f = 1 \text{ kHz}$	270			Ω	—
R_o Output resistance (pin 7)		7.5			$k\Omega$	
R_o Output resistance (pin 8)		300			Ω	
C_i Input capacitance (pin 2)	$f = 5.5 \text{ MHz}$	4			pF	—
C_o Output capacitance (pin 9)		7.5			pF	
G_v Audio voltage gain	$f = 1 \text{ kHz}$ $V_i = 0.1 \text{ V}$	17.5	20		dB	4
P_{tot} Total power dissipation		343	370	400	mW	—
AMR Amplitude modulation rejection	$f = 5.5 \text{ MHz}$	40	50		dB	3

TBA 780

SCHEMATIC DIAGRAM

Fig. 1

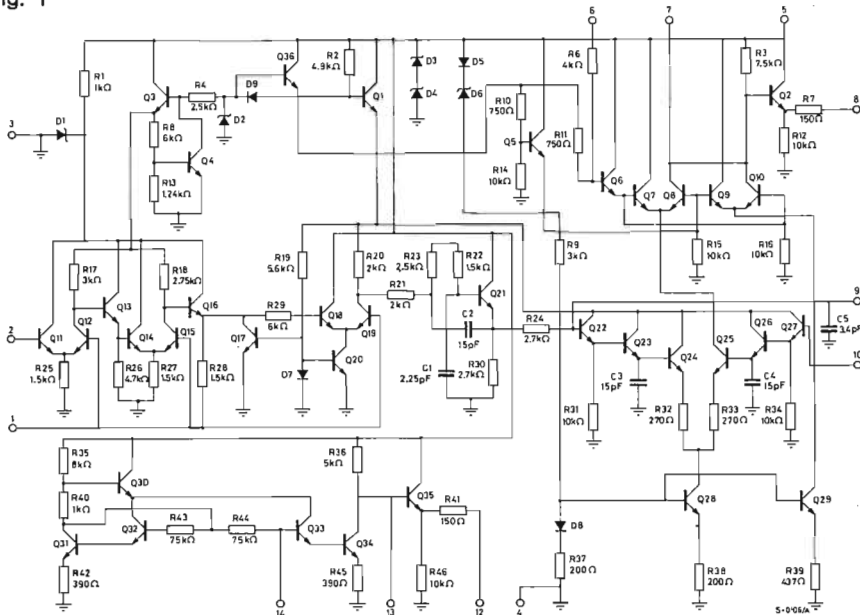
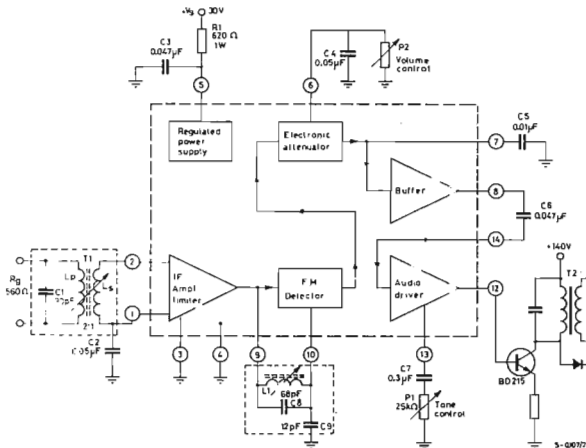


Fig. 2 - Typical circuit utilizing the TBA 780 and BD 215 transistor



1) $T_1 = 5.5$ MHz transformer

$L_p = 5.5 \mu\text{H}$; $Q_0=80$; 19 turns
 $\varnothing 0.15$ mm silk-covered
 copper wire with powdered-iron core

$L_s = 9$ turns $\varnothing 0.15$ mm

2) $T_2 =$ Audio output transformer:

The dimensions of the transformer and of the circuit parameters are to be evaluated on the basis of the output power desired and of the load to be used

3) $L_1 =$ Single tuned discriminator coil: $12 \mu\text{H}$; $Q_0 = 50$ (58 turns $\varnothing 0.08$ mm with powdered-iron core)

TBA 780

Fig. 3 - Input limiting voltage, AM rejection, recovered audio, total harmonic distortion, maximum attenuation, maximum "play-through" test circuit

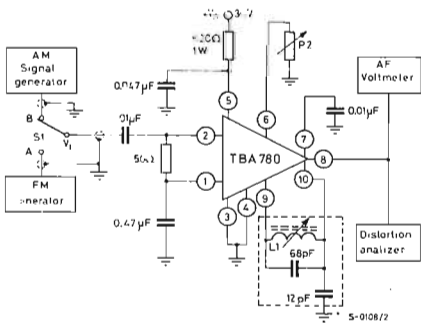


Fig. 4 - Audio voltage gain (undistorted output) test circuit

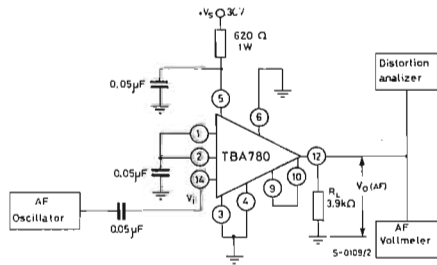
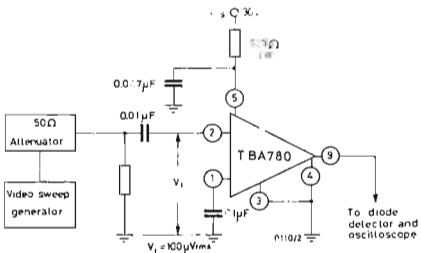


Fig. 5 - IF amplifier voltage gain test circuit



TBA 780

Fig. 6 - Typical IF amplifier voltage gain

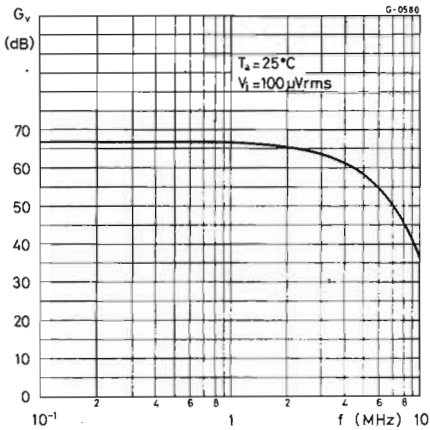


Fig. 7 - Typical AF amplifier voltage gain

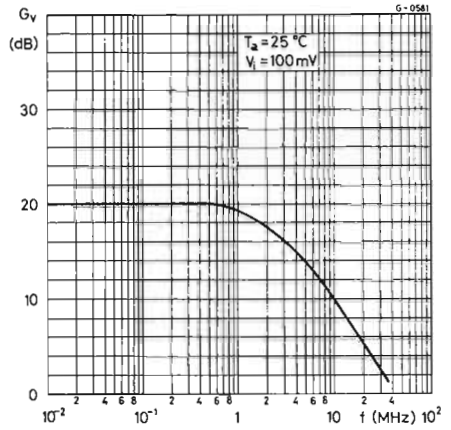


Fig. 8 - Typical FM detector output voltage versus input voltage

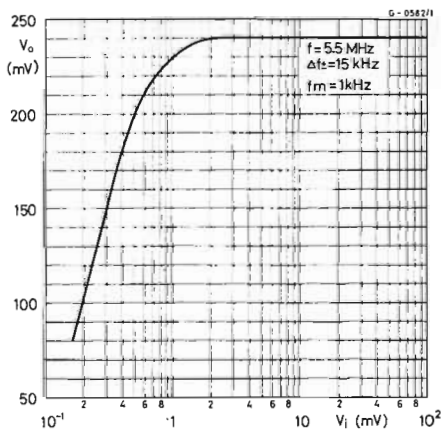
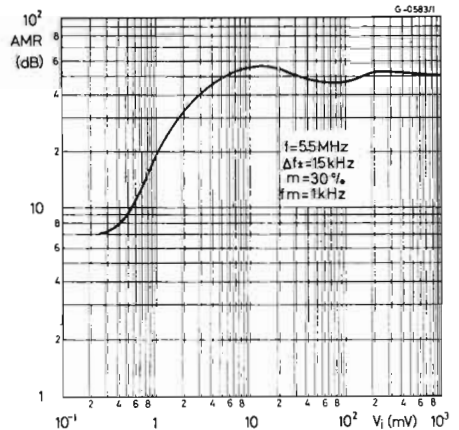


Fig. 9 - Typical amplitude-modulation rejection versus input voltage



TBA 780

Fig. 10 - Typical gain reduction versus resistance (P2) (terminal 6 to gnd)

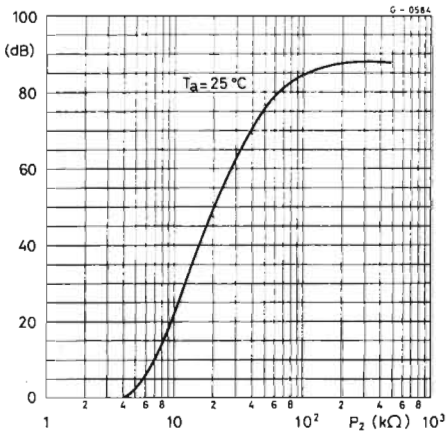
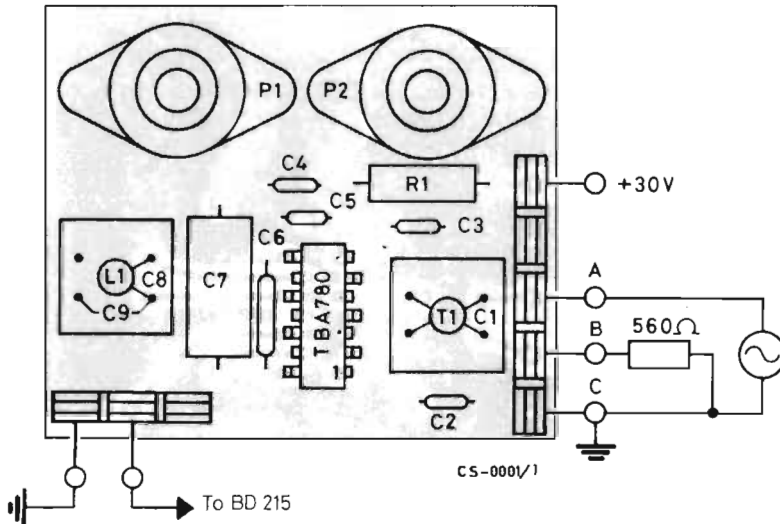


Fig. 11 - P.C. board layout, 1:1 scale (fig. 2 circuit)



TBA 800

LINEAR INTEGRATED CIRCUIT

PRELIMINARY DATA

AUDIO POWER AMPLIFIER

The TBA 800 is an integrated monolithic power amplifier in a 12-lead quad in-line plastic package. The external cooling tabs enable 2.5 W output power to be achieved without external heat-sink and 5 W output power using a small area of the P.C. board Copper as a heat sink.

It is intended for use as a low frequency Class B amplifier.

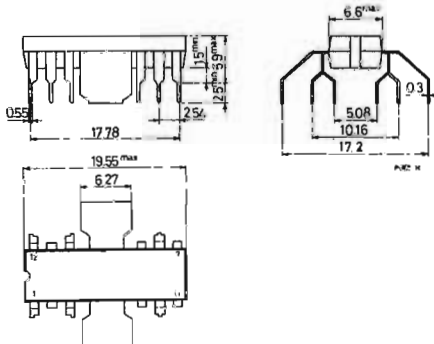
The TBA 800 provides 5 W output power at 24 V/16 Ω and works with a wide range of supply voltages (5-30 V); it gives high output current (up to 1.5 A), high efficiency (70% at 4 W output), very low harmonic distortion and no cross-over distortion.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	30	V
I_o	Output peak current (non repetitive)	2	A
$\rightarrow I_o$	Output current (repetitive)	1.5	A
$\rightarrow P_{tot}$	Power dissipation at $T_{amb} = 80^\circ\text{C}$	1	W
	at $T_{tab} = 90^\circ\text{C}$	5	W
$\rightarrow T_{stg}; T_j$	Storage and junction temperature	-40 to 150	$^\circ\text{C}$

MECHANICAL DATA

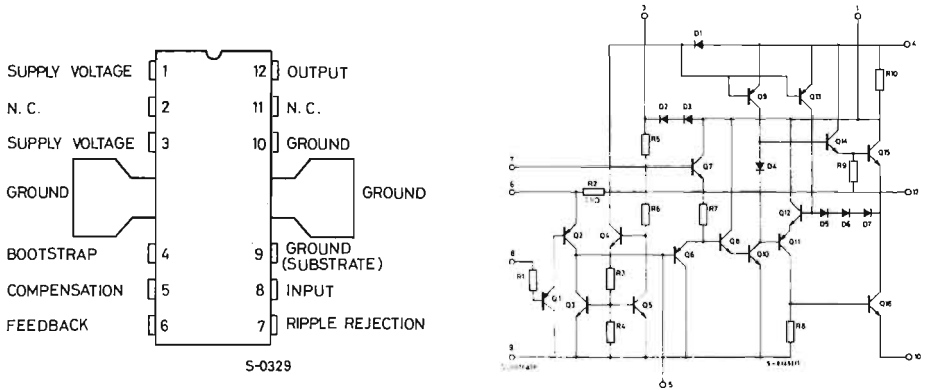
Dimensions in mm



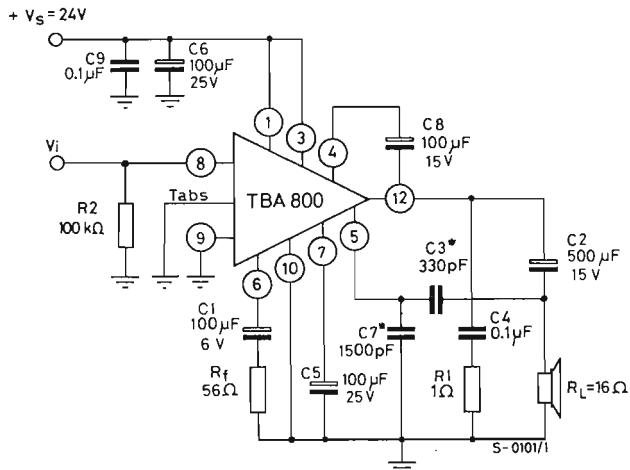
TBA 800

CONNECTION AND SCHEMATIC DIAGRAM

(top view)



TEST AND APPLICATION CIRCUIT



* C3, C7 see fig. 5

TBA 800

THERMAL DATA

→ $R_{th\ j-tab}$	Thermal resistance junction-tab	max	12	°C/W
→ $R_{th\ j-amb}$	Thermal resistance junction-ambient	max	70*	°C/W

* Obtained with tabs soldered to printed circuit with minimized copper area.

ELECTRICAL CHARACTERISTICS (Refer to the TEST CIRCUIT, $T_{amb} = 25\text{°C}$)

	Parameter	Test conditions	Min.	Typ.	Max.	Unit
	V_s Supply voltage		5		30	V
	V_o Quiescent output voltage (pin 12)	$V_s = 24\text{ V}$	11	12	13	V
→	I_d Quiescent drain current (pin 12)	$V_s = 24\text{ V}$		9	20	mA
→	I_b Bias current (pin 8)	$V_s = 24\text{ V}$		1	5	μA
→	P_o Output power	$d = 10\%$ $V_s = 24\text{ V}$ $R_L = 16\ \Omega$ $f = 1\text{ kHz}$	4.4	5		W
→	$V_{i(rms)}$ Input voltage				220	mV
→	$V_{i(rms)}$ Input sensitivity	$P_o = 5\text{ W}$ $V_s = 24\text{ V}$ $R_L = 16\ \Omega$ $f = 1\text{ kHz}$		80		mV
→	R_i Input resistance (pin 8)			5		MΩ
→	B Frequency response (-3 dB)	$V_s = 24\text{ V}$ $R_L = 16\ \Omega$ $C3 = 330\text{ pF}$		40 to 20000		Hz
	d Distortion	$P_o = 50\text{ mW to } 2.5\text{ W}$ $V_s = 24\text{ V}$ $R_L = 16\ \Omega$ $f = 1\text{ kHz}$		0.5		%
→	G_v Voltage gain (open loop)	$V_s = 24\text{ V}$ $R_L = 16\ \Omega$ $f = 1\text{ kHz}$		80		dB
→	G_v Voltage gain (closed loop)	$V_s = 24\text{ V}$ $R_L = 16\ \Omega$ $f = 1\text{ kHz}$	39	42	45	dB

TBA 800

ELECTRICAL CHARACTERISTICS (continued)

	Parameter	Test conditions	Min. Typ. Max.	Unit
→	e_N Input noise voltage	$V_s = 24\text{ V}$ $R_g = 0$ $B(-3\text{ dB}) = 40\text{ to }20,000\text{ Hz}$	5	μV
→	i_N Input noise current	$V_s = 24\text{ V}$ $B(-3\text{ dB}) = 40\text{ to }20,000\text{ Hz}$	0.2	nA
	η Efficiency	$P_o = 4\text{ W}$ $V_s = 24\text{ V}$ $R_L = 16\ \Omega$ $f = 1\text{ kHz}$	70	%

Fig. 1 - Typical output power versus supply voltage

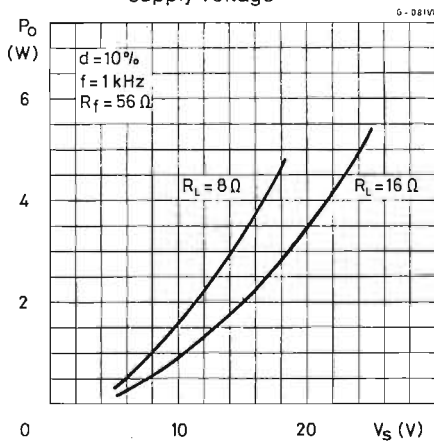
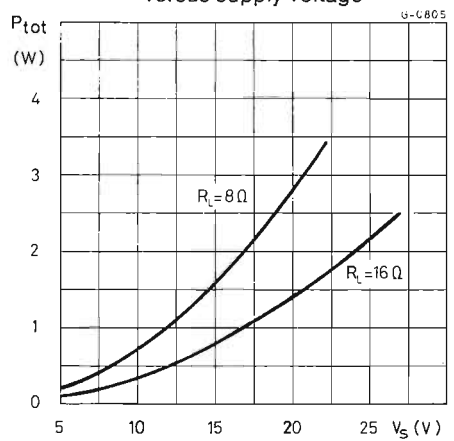


Fig. 2 - Maximum power dissipation versus supply voltage



TBA 800

Fig. 3 - Typical distortion versus output power

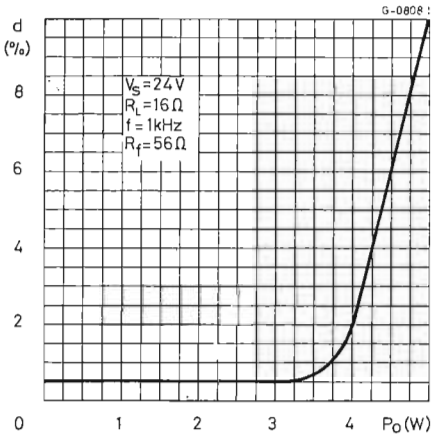


Fig. 4 - Typical distortion versus frequency

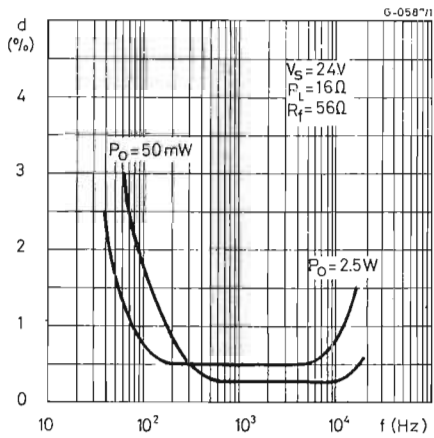


Fig. 5 - Value of C3 versus R_f for various values of B

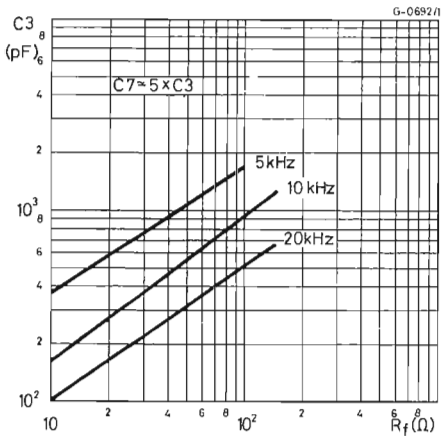
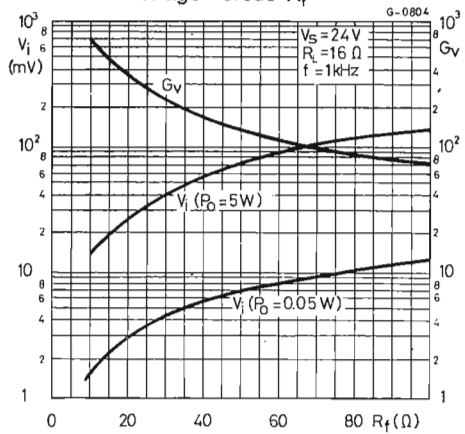


Fig. 6 - Typical voltage gain (closed loop) and typical input voltage versus R_f



TBA 800

Fig. 7 - Typical power dissipation and efficiency versus output power

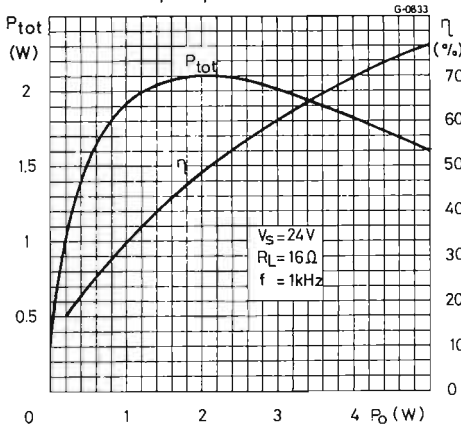


Fig. 8 - Typical quiescent output voltage (pin 12) versus supply voltage

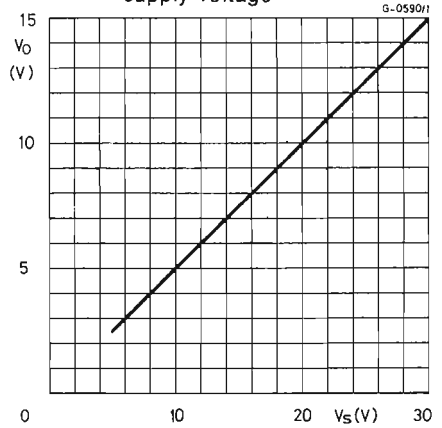
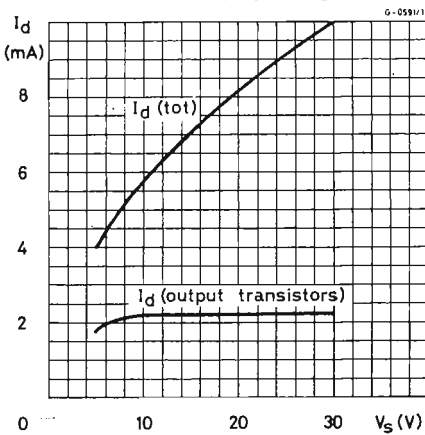
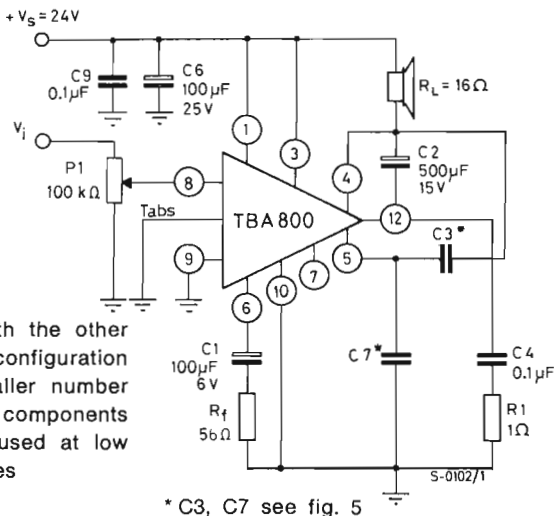


Fig. 9 - Typical quiescent current versus supply voltage



APPLICATION INFORMATION

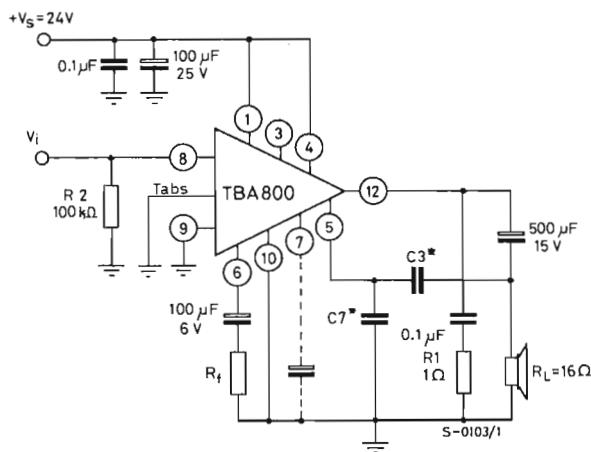
Fig. 10 - Circuit with the load connected to the supply voltage



Compared with the other circuits, this configuration entails a smaller number of external components and can be used at low supply voltages

* C3, C7 see fig. 5

Fig. 11 - Circuit with load connected to ground without bootstrap



* C3, C7 see fig. 5

There is no bootstrap connection and hence there is a greater loss of potential output swing.

This circuit is only for use at high voltages.

In the absence of "bootstrap", the reduction in the upper part of the wave is greater than that in the lower part: if pin 3 is left open circuit, this automatically inserts diodes D2 - D3 (see schematic diagram) and this enables a symmetrical wave to be obtained at the output.

TBA 800

Fig. 12 - Typical distortion versus output power (fig. 11 circuit)

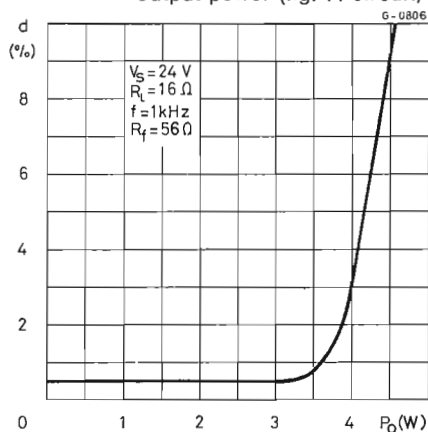


Fig. 13 - Typical output power versus supply voltage (fig. 11 circuit)

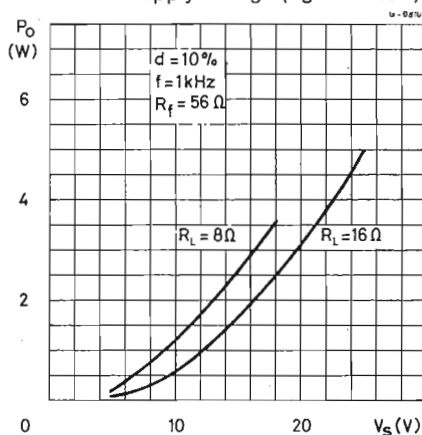
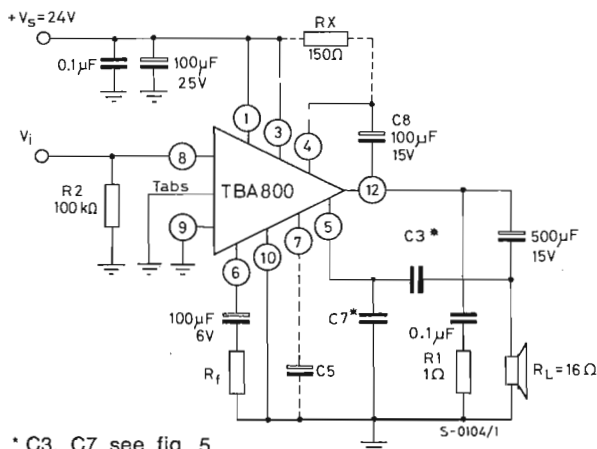


Fig. 14 - Circuit with load connected to ground with bootstrap



* C3, C7 see fig. 5

The bootstrap capacitor C8 enables the same electrical characteristics as circuit of to test circuit to be achieved. For low supply voltage operation (e.g. 9-14 V), 150 Ω is connected between pin 1 and pin 4

N.B. - For the circuits of figures 11 and 14 an excellent supply voltage ripple rejection is obtained by connecting the capacitor C5 (10 to 100 μF - 25 V) between pin 7 and ground.

TBA 800

MOUNTING INSTRUCTIONS

The tabs on the TBA 800 can be used to conduct away the heat generated in the integrated circuit so that the junction temperature does not exceed the permissible maximum (150 °C).

This may be done by connecting tabs to an external heat sink, or by soldering it to a suitable Copper area of the printed circuit board (fig. 15 a).

Fig. 15 b shows a simple type of heat sink. Assuming an area of copper on the printed circuit board of only 2 cm², the total R_{th} between junction to ambient is approximately 30 °C/W.

External heat sink or printed circuit copper area must be connected to electrical ground.

In the latter case, fig. 16 shows the maximum dissipated power (for $T_{amb} = 55$ °C) as a function of the side of two equal square Copper areas having a thickness of 35 μ (1.4 mils).

Fig. 15 a - Example of an area of P.C. board copper soldered to the tabs of the TBA 800, which is used as a heat dissipator.

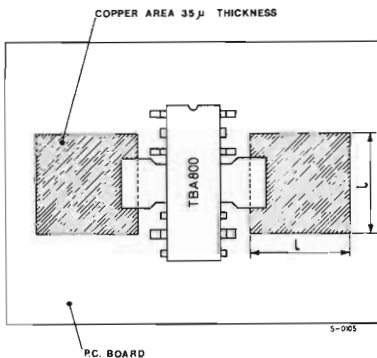
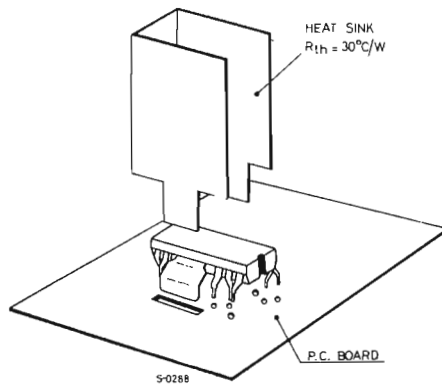


Fig. 15 b - Example of TBA 800 with external heat-sink.



TBA 800

Fig. 16 - Power that can be dissipated versus "l"

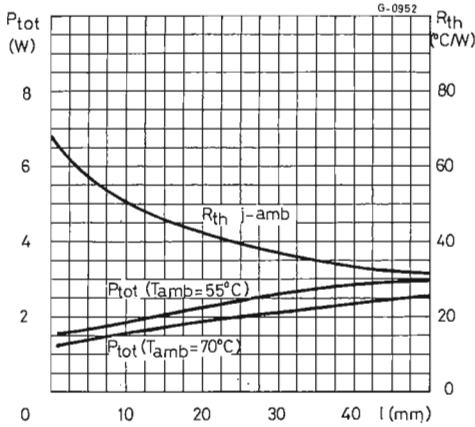


Fig. 17 - Power rating characteristics

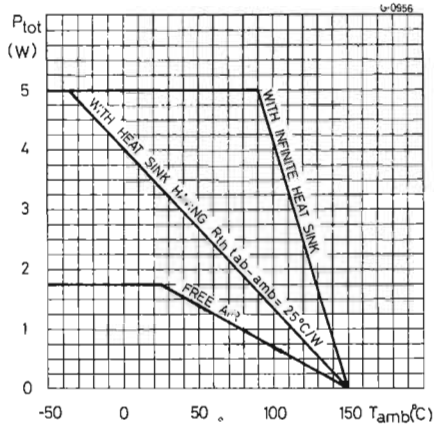
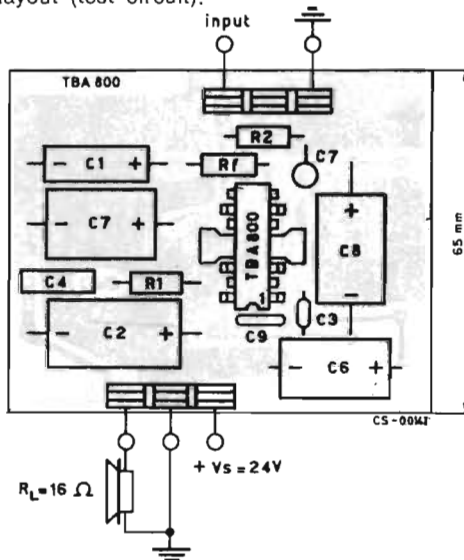
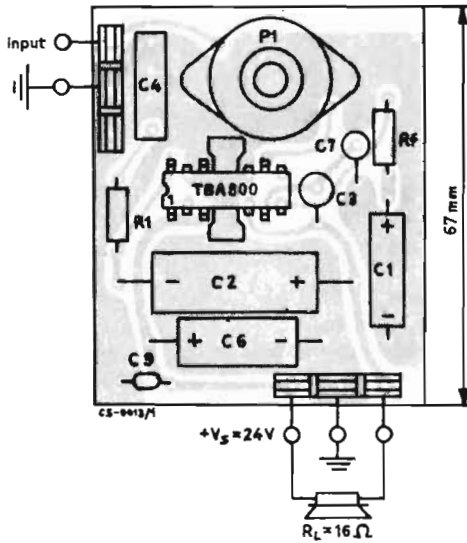


Fig. 18 - P.C. board layout (test circuit).



TBA 800

Fig. 19 - P.C. board layout (fig. 10 circuit)



PROCEDURE TO CALCULATE AREA OF COPPER NEEDED

1) Calculate maximum power dissipation

$$P_{\text{tot}} = 0.4 \cdot \frac{V_{S \text{ max}}^2}{8 R_L} + V_{S \text{ max}} I_d$$

where

- $V_{S \text{ max}}$ = maximum value of supply voltage (increase 10% if not stabilized)
- R_L = load resistance
- I_d = quiescent drain current for typical value see fig. 10; maximum value at $V_S = 24 \text{ V}$ is 20 mA (for worst case design)
- $T_{\text{amb max}}$ = 70 °C

TBA 800

PROCEDURE TO CALCULATE AREA OF COPPER NEEDED (continued)

2) Fig. 16 gives ℓ

Examples:

a) V_S (not stabilized) = 24 V; $R_L = 16 \Omega$

$$P_{\text{tot}} = 0.4 \cdot \frac{(24 + 2.4)^2}{8 \cdot 16} + (24 + 2.4) \cdot 20 \cdot 10^{-3} = 2.6 \text{ W}$$

From fig. 16 $\ell \cong 25 \text{ mm}$

For geometries different from the one of fig. 15 note that copper areas near the tabs have better efficiency as regards power dissipation. Therefore additional safety factors must be added for worst case designs

b) V_S (stabilized) = 12 V; $R_L = 8 \Omega$

$$P_{\text{tot}} = 0.4 \cdot \frac{12^2}{8 \cdot 8} + 0.02 \cdot 12 = 1 \text{ W}$$

The fig. 16 shows that no heat sink is required if $T_{\text{amb}} \leq 55^\circ\text{C}$

TBA 810S TBA 810AS

PRELIMINARY DATA

7 W AUDIO POWER AMPLIFIER WITH THERMAL SHUT-DOWN

The TBA 810 S is a monolithic integrated circuit in a 12-lead quad in-line plastic package, intended for use as a low frequency class B amplifier.

The TBA 810 S provides 7 W power output at 16 V/4 Ω , 6 W at 14.4 V/4 Ω , 2.5 W at 9 V/4 Ω , 1 W at 6 V/4 Ω and works with a wide range of supply voltages (4 to 20 V); it gives high output current (up to 2.5 A), high efficiency (75% at 6 W output), very low harmonic and cross-over distortion. In addition, the circuit is provided with a thermal limiting circuit which fundamentally changes the criteria normally used in determining the size of the heat sink.

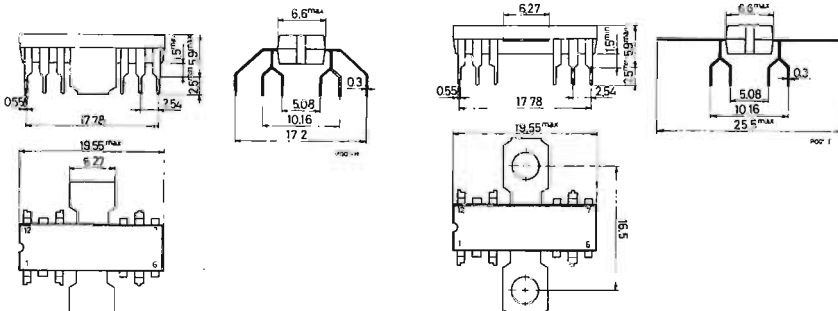
The TBA 810 AS has the same electrical characteristics as the TBA 810 S, but its cooling tabs are flat and pierced so that an external heat sink can easily be attached.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	20	V
I_o	Output peak current (non-repetitive)	3.5	A
I_o	Output current (repetitive)	2.5	A
P_{tot}	Power dissipation: at $T_{amb} = 70^\circ\text{C}$	1	W
	at $T_{tab} = 100^\circ\text{C}$	5	W
T_{stg}, T_j	Storage and junction temperature	-40 to 150	$^\circ\text{C}$

MECHANICAL DATA

Dimensions in mm

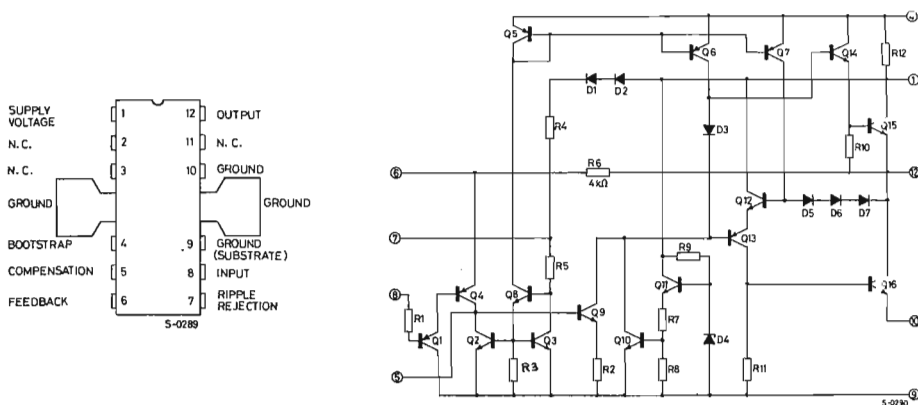


TBA 810 S

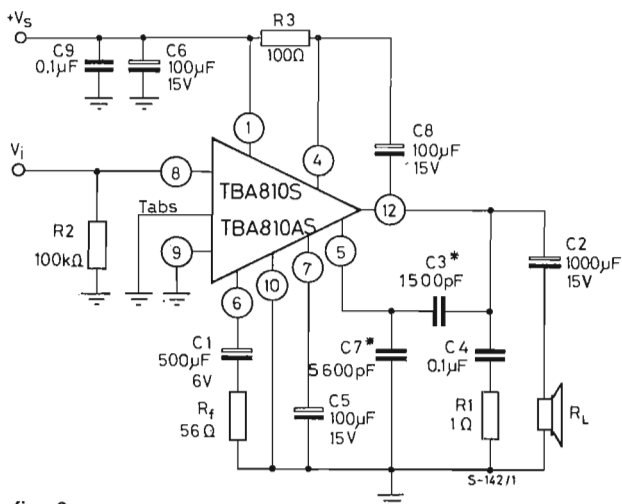
TBA 810 AS

TBA 810S TBA 810AS

CONNECTION AND SCHEMATIC DIAGRAM (top view)



TEST AND APPLICATION CIRCUIT



* C_3 , C_7 see fig. 6.

TBA 810S TBA 810AS

THERMAL DATA			TBA 810S	TBA 810AS
$R_{th\ j-tab}$	Thermal resistance junction-tab	max	12 °C/W	10 °C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	70* °C/W	80 °C/W

* Obtained with tabs soldered to printed circuit with minimized copper area.

ELECTRICAL CHARACTERISTICS (Refer to the test circuit; $T_{amb} = 25\text{ °C}$)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	
V_s	Supply voltage (pin 1)	4		20	V	
V_o	Quiescent output voltage (pin 12)	6.4	7.2	8	V	
I_d	Quiescent drain current	$V_s = 14.4\text{ V}$		12	20	mA
I_b	Bias current (pin 8)			0.4		μA
P_o	Power output	$d = 10\%$ $R_L = 4\ \Omega$ $f = 1\text{ kHz}$ $V_s = 16\text{ V}$ $V_s = 14.4\text{ V}$ $V_s = 9\text{ V}$ $V_s = 6\text{ V}$		7 6 2.5 1	W W W W	
$V_{i(rms)}$	Input voltage			220	mV	
V_i	Input sensitivity	$P_o = 6\text{ W}$ $V_s = 14.4\text{ V}$ $R_L = 4\ \Omega$ $f = 1\text{ kHz}$ $R_f = 56\ \Omega$ $R_f = 22\ \Omega$		80 35	mV mV	
R_i	Input resistance (pin 8)			5	M Ω	
B	Frequency response (-3 dB)	$V_s = 14.4\text{ V}$ $R_L = 4\ \Omega$ $C3 = 820\text{ pF}$ $C3 = 1500\text{ pF}$		40 to 20,000 40 to 10,000	Hz Hz	

TBA 810S

TBA 810AS

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
d	Distortion	$P_o = 50 \text{ mW to } 3 \text{ W}$ $V_s = 14.4 \text{ V}$ $R_L = 4 \Omega$ $f = 1 \text{ kHz}$		0.3		%
G_v	Voltage gain (open loop)	$V_s = 14.4 \text{ V}$ $R_L = 4 \Omega$ $f = 1 \text{ kHz}$		80		dB
G_v	Voltage gain (closed loop)	$V_s = 14.4 \text{ V}$ $R_L = 4 \Omega$ $f = 1 \text{ kHz}$	34	37	40	dB
e_N	Input noise voltage	$V_s = 14.4 \text{ V}$ $R_g = 0$ $B(-3 \text{ dB}) = 20 \text{ Hz to } 20,000 \text{ Hz}$		2		μV
i_N	Input noise current	$V_s = 14.4 \text{ V}$ $B(-3 \text{ dB}) = 20 \text{ Hz to } 20,000 \text{ Hz}$		0.1		nA
η	Efficiency	$P_o = 5 \text{ W}$ $V_s = 14.4 \text{ V}$ $R_L = 4 \Omega$ $f = 1 \text{ kHz}$		70		%
SVR	Supply voltage rejection	$V_s = 14.4 \text{ V}$ $R_L = 4 \Omega$ $f_{\text{ripple}} = 100 \text{ Hz}$		38		dB

TBA 810S TBA 810AS

Fig. 1 - Typical power output versus supply voltage

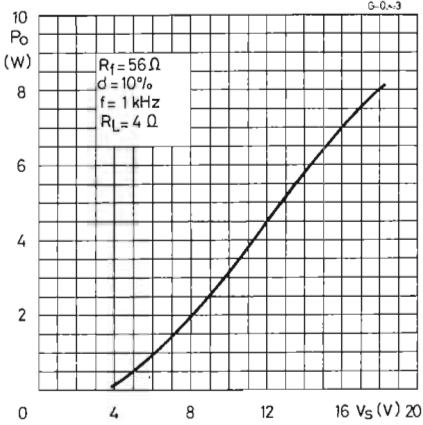


Fig. 2 - Maximum power dissipation versus supply voltage (sine wave operation)

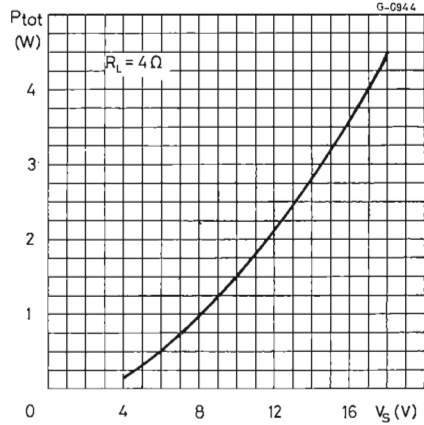


Fig. 3 - Typical distortion versus output power

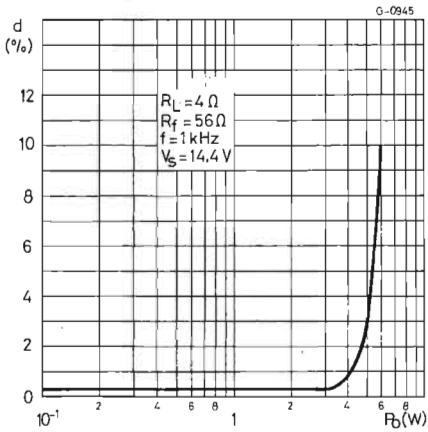
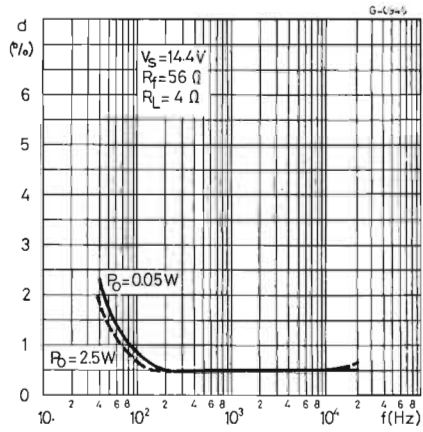


Fig. 4 - Typical distortion versus frequency



TBA 810S TBA 810AS

Fig. 5 - Typical relative voltage gain (closed loop) and typical input voltage versus feedback resistance (R_f)

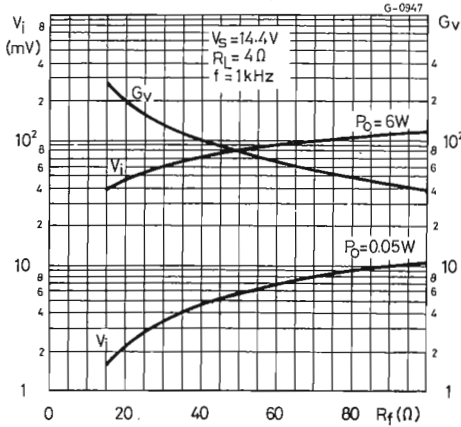


Fig. 6 - Typical value of C_3 versus R_f for various values of B

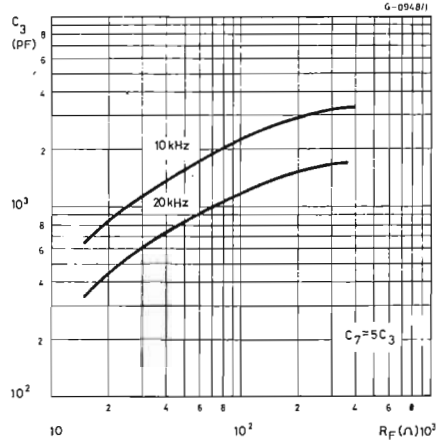


Fig. 7 - Typical power dissipation and efficiency versus output power

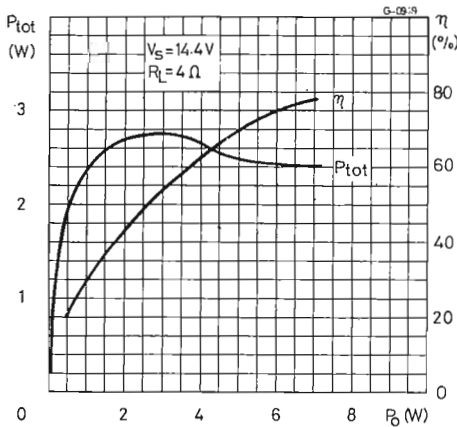
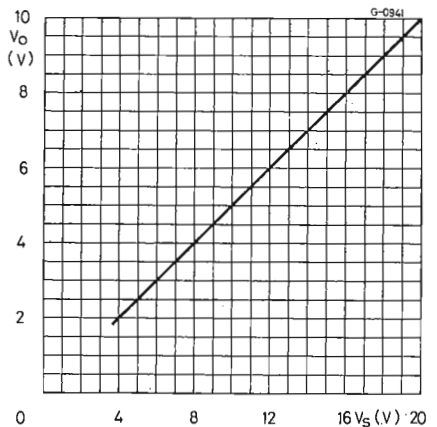


Fig. 8 - Typical quiescent output voltage (pin 12) versus supply voltage



TBA 810S TBA 810AS

Fig. 9 - Typical quiescent current versus supply voltage

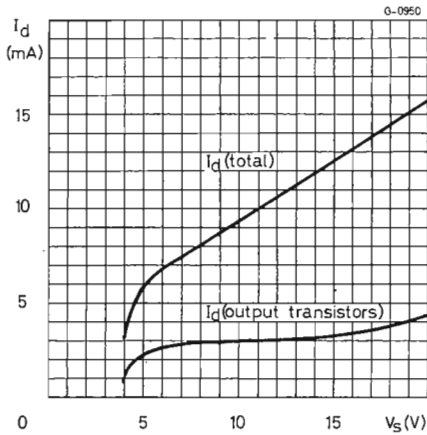
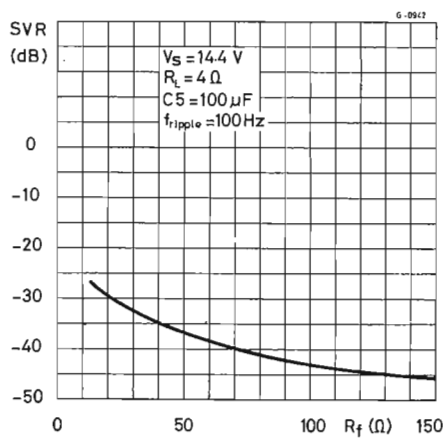
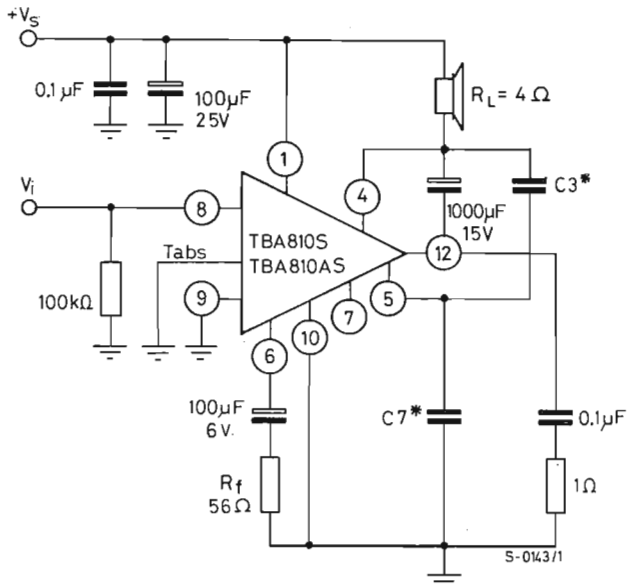


Fig. 10 - Typical supply voltage rejection



For portable equipment the circuit in Fig. 11 has the advantages of fewer external components and a better behaviour at low supply voltages (down to 4 V).

Fig. 11 - Typical circuit with load connected to the supply voltage



* C_3 , C_7 see fig. 6.

TBA 810S TBA 810AS

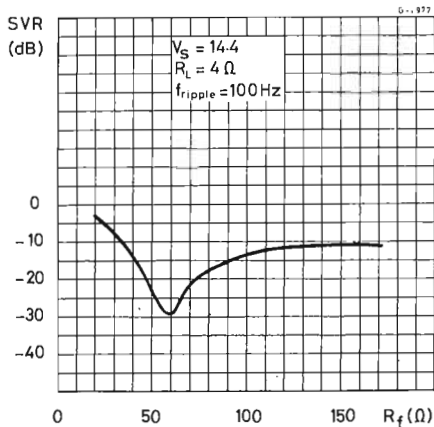


Fig. 12 - Typical supply voltage rejection versus R_f (fig. 11 circuit)

MOUNTING INSTRUCTIONS

The thermal power dissipated in the circuit may be removed by connecting the tabs to an external heat sink (TBA 810 AS - fig. 13) or by soldering them to an area of copper on the printed circuit board (TBA 810 S - fig. 14).

During soldering the tabs temperature must not exceed 260 °C and the soldering time must not be longer than 12 seconds.

Fig. 15a and 15b show two ways that can be used for mounting the device.

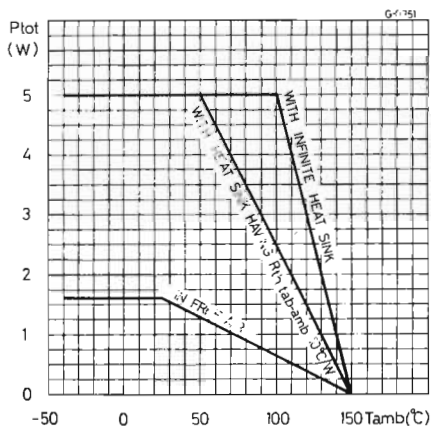
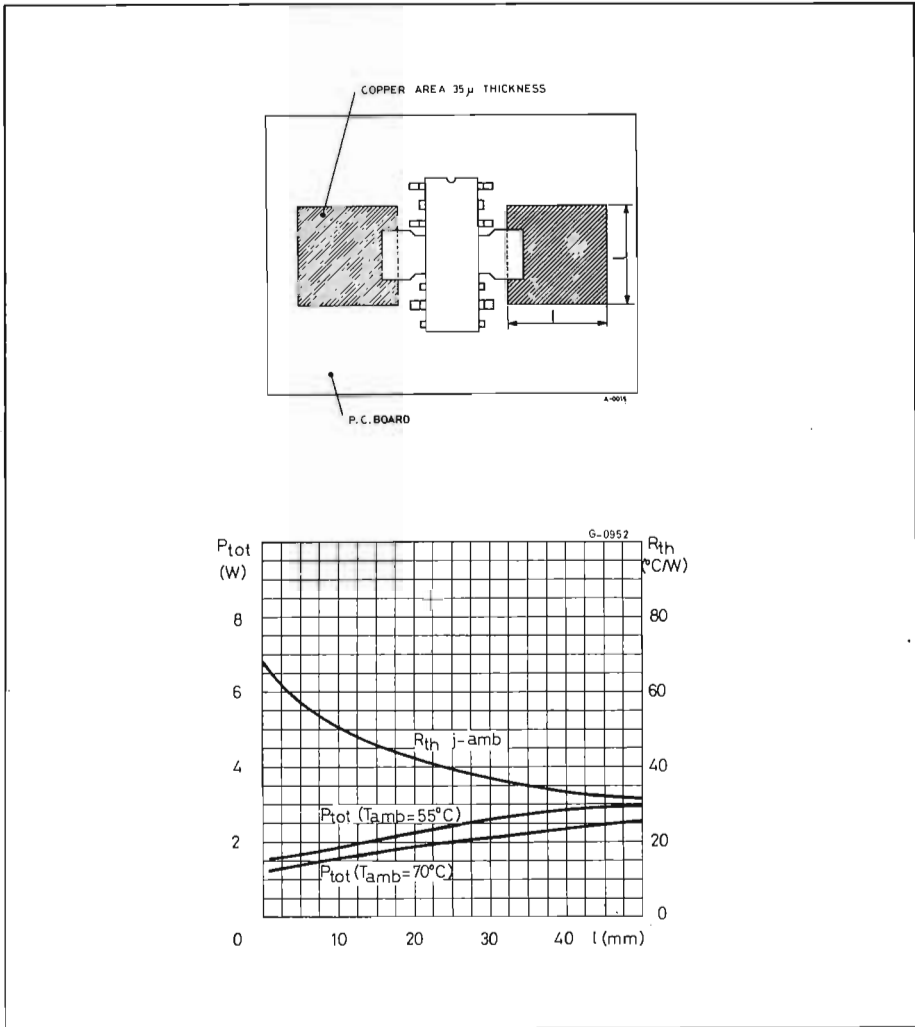


Fig. 13 - Maximum power dissipation versus ambient temperature (for TBA 810 AS only)

TBA 810S TBA 810AS

Fig. 14 - Maximum power dissipation versus copper area of the P.C. board
(for TBA 810 S only)



TBA 810S TBA 810AS

Fig. 15a shows a method, of mounting the TBA 810 S, that is satisfactory both from the point of view of heat dissipation and from mechanical considerations. For TBA 810 AS the desired thermal resistance is obtained by fixing the elements shown in fig. 15b, to a suitably dimensioned plate. This plate can also act as a support for the whole printed circuit board; the mechanical stresses do not damage the integrated circuit. This is firmly fixed to the element, in fig. 15b.

Fig. 15a

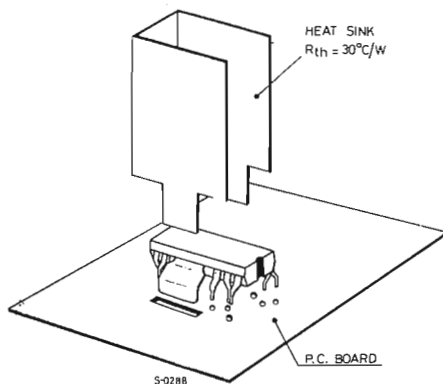
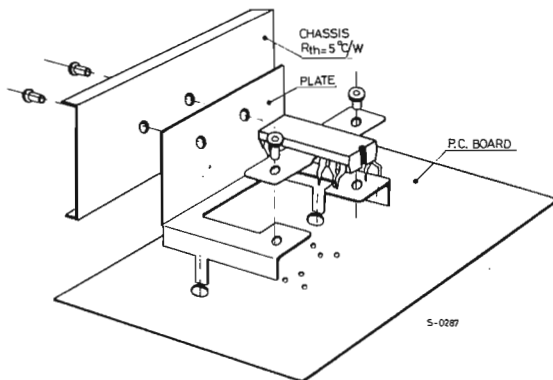


Fig. 15b

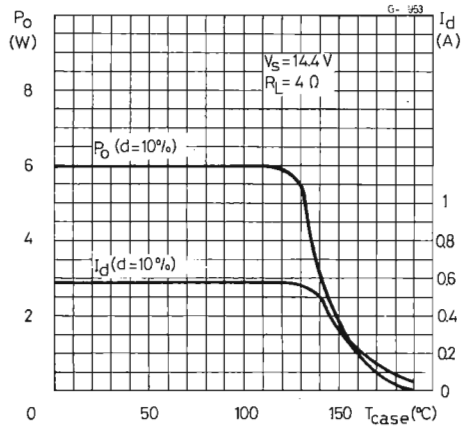


THERMAL SHUT-DOWN

The presence of a thermal limiting circuit offers the following advantages:

- 1) An overload on the output (even if it is permanent), or an above-limit ambient temperature can be easily supported.
- 2) The heat sink can have a smaller factor of safety compared with that of a conventional circuit. There is no device damage in the case of too high a junction temperature: all that happens is that P_o (and therefore P_{tot}) and I_d are reduced (fig. 16).

Fig. 16 - Output power and drain current versus package temperature



Manufactured under SGS-ATES patents held in several countries, and corresponding to one or more of the following Italian patents: 155139 and others pending.

LINEAR INTEGRATED CIRCUIT

TBA 820

PRELIMINARY DATA

AUDIO AMPLIFIER

The TBA 820 is an integrated monolithic audio amplifier in a 14-lead quad in-line plastic package.

It is intended for use as low frequency class B amplifier with wide range of supply voltage: 3 to 16 V.

Main features are: minimum working voltage of 3 V, low quiescent current, low number of external components, good ripple rejection, no cross-over distortion, mounting compatibility with TAA 611 (see note on last page).

Output power:

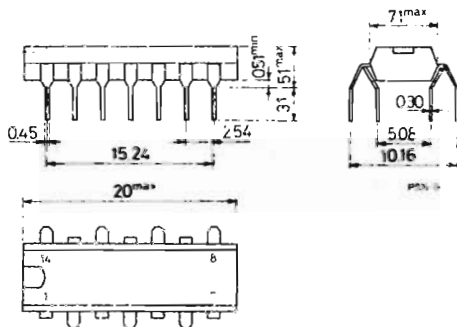
$$P_o = 2 \text{ W at } 12 \text{ V} - 8\Omega \bullet P_o = 1.6 \text{ W at } 9 \text{ V} - 4\Omega \bullet P_o = 1.2 \text{ W at } 9 \text{ V} - 8\Omega$$

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	16	V
I_o	Output peak current	1.5	A
P_{tot}	Power dissipation at $T_{amb} = 50^\circ\text{C}$	1.25	W
$T_{stg}; T_j$	Storage and junction temperature	-40 to 150	$^\circ\text{C}$

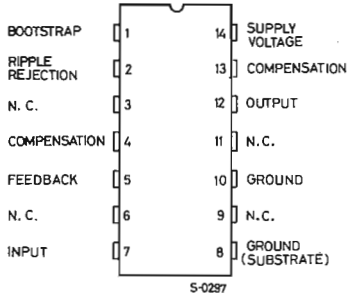
MECHANICAL DATA

Dimensions in mm

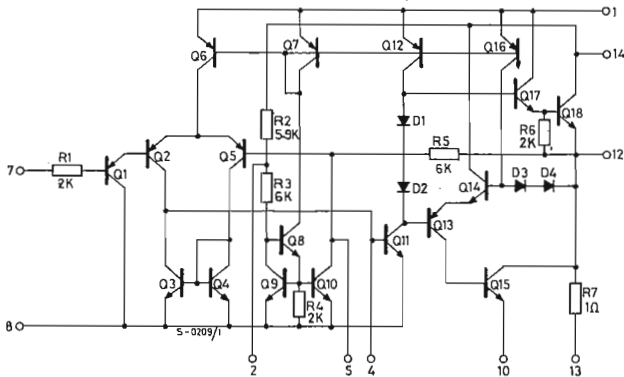


TBA 820

CONNECTION DIAGRAM



SCHEMATIC DIAGRAM



TBA 820

TEST AND APPLICATION CIRCUITS

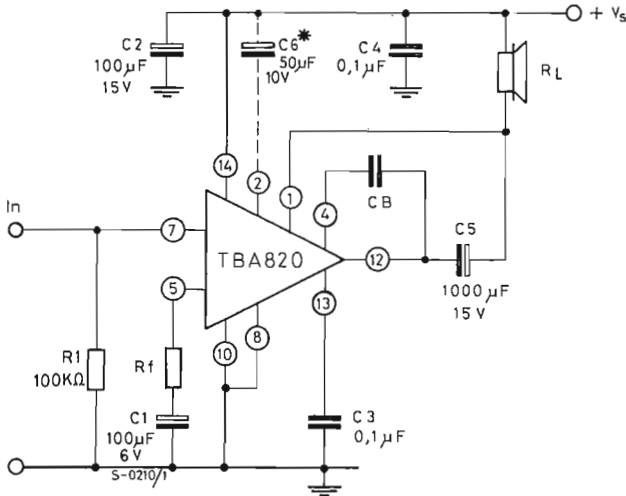


Fig. 1
Circuit diagram with
load connected to the
supply voltage

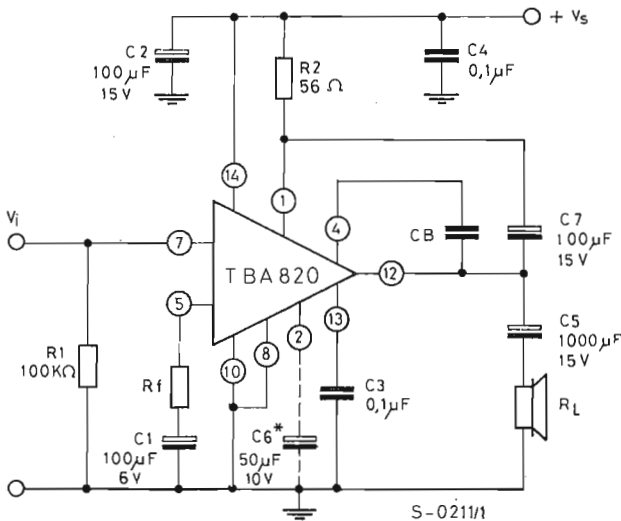


Fig. 2
Circuit diagram
with load connected
to ground

* Capacitor C_6 must be used when high ripple rejection is requested.

TBA 820

THERMAL DATA

$R_{th\ j-amb}$	Thermal resistance junction-ambient (copper frame)	80 °C/W
-----------------	--	---------

ELECTRICAL CHARACTERISTICS

(Output powers measured at pin 12, $T_{amb} = 25\text{ °C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.	
V_s	Supply voltage	3		16	V	—	
V_o	Quiescent output voltage (pin 12)	$V_s = 9\text{ V}$	4	4.5	5	V	—
I_d	Quiescent drain current	$V_s = 9\text{ V}$		4	mA	—	
I_b	Bias current (pin 7)	$V_s = 9\text{ V}$		0.1	μA	—	
P_o	Output power	$d = 10\%$ $f = 1\text{ kHz}$ $R_f = 120\ \Omega$ $V_s = 12\text{ V}$ $R_L = 8\ \Omega$ $V_s = 9\text{ V}$ $R_L = 4\ \Omega$ $V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $V_s = 6\text{ V}$ $R_L = 4\ \Omega$ $V_s = 3.5\text{ V}$ $R_L = 4\ \Omega$			2 1.6 1.2 0.75 0.22	W W W W W	1
$V_{i(rms)}$	Input sensitivity	$P_o = 1.2\text{ W}$ $V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $f = 1\text{ kHz}$ $R_f = 33\ \Omega$ $R_f = 120\ \Omega$			16 60	mV mV	1
$V_{i(rms)}$	Input sensitivity	$P_o = 50\text{ mW}$ $V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $f = 1\text{ kHz}$ $R_f = 33\ \Omega$ $R_f = 120\ \Omega$			3.5 12	mV mV	1
R_i	Input resistance			5	$\text{M}\Omega$	—	
B	Frequency response (-3 dB)	$V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $R_f = 120\ \Omega$ $C_B = 680\text{ pF}$ $C_B = 220\text{ pF}$			25 to 7000 25 to 20000	Hz Hz	1
d	Distortion	$P_o = 500\text{ mW}$ $V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $f = 1\text{ kHz}$ $R_f = 33\ \Omega$ $R_f = 120\ \Omega$			0.8 0.4	% %	1

TBA 820

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
G_v Voltage gain (open loop)	$V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $f = 1\text{ kHz}$		75		dB	—
G_v Voltage gain (closed loop)	$V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $f = 1\text{ kHz}$ $R_f = 33\ \Omega$ $R_f = 120\ \Omega$		45 34		dB dB	—
e_N Input noise voltage	$V_s = 9\text{ V}$ $B (-3\text{ dB}) =$ $= 25\text{ to }20000\text{ Hz}$		3		μV	—
i_N Input noise current	$V_s = 9\text{ V}$ $B (-3\text{ dB}) =$ $= 25\text{ to }20000\text{ Hz}$		0.4		nA	—
$\frac{S+N}{N}$ Signal and noise to noise ratio	$V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $R_f = 120\ \Omega$ $B (-3\text{ dB}) =$ $= 25\text{ to }20000\text{ Hz}$ $R_1 = 100\text{ k}\Omega$ $P_o = 1.2\text{ W}$		70		dB	—
SVR Supply voltage rejection	$V_s = 9\text{ V}$ $R_L = 8\ \Omega$ $f(\text{ripple}) = 100\text{ Hz}$ $C_6 = 50\ \mu\text{F}$ $R_f = 120\ \Omega$		42		dB	2

Fig. 3 - Typical power output

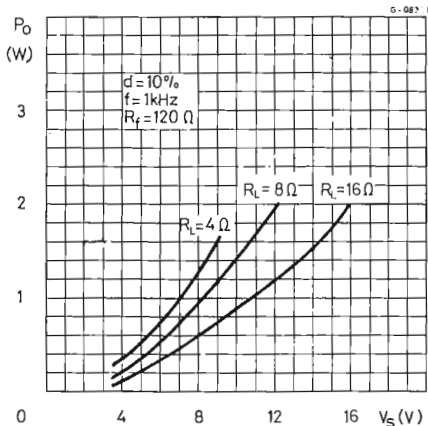
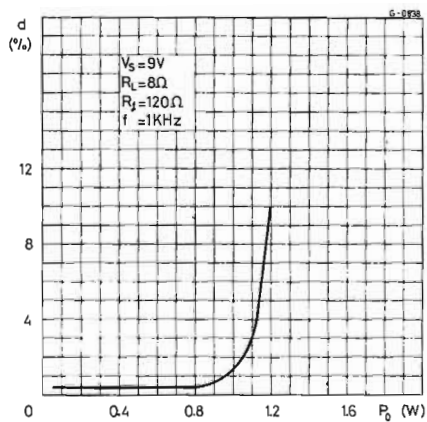


Fig. 4 - Typical distortion



TBA 820

Fig. 5 - Typical power dissipation and efficiency

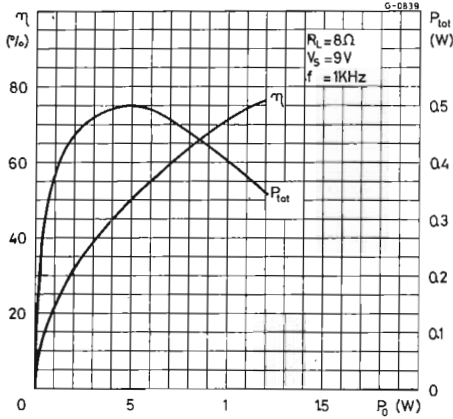


Fig. 6 - Maximum power dissipation (sine wave operation)

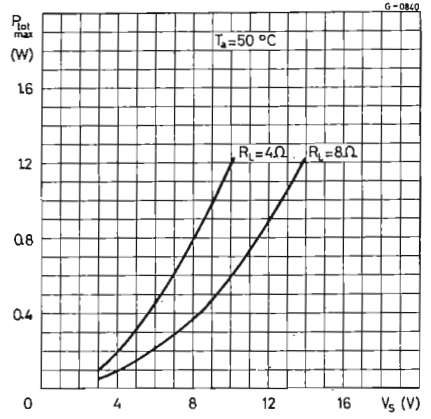


Fig. 7 - Typical value of C_B versus R_f

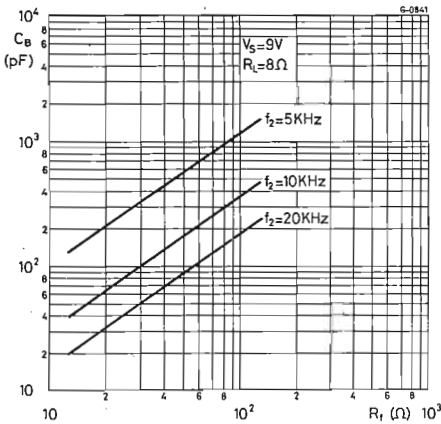


Fig. 8 - Typical relative frequency response

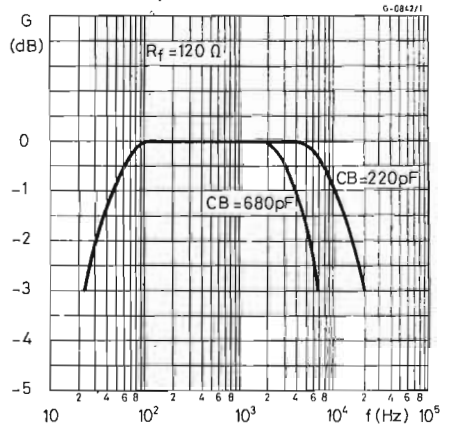


Fig. 9 - Typical input sensitivity

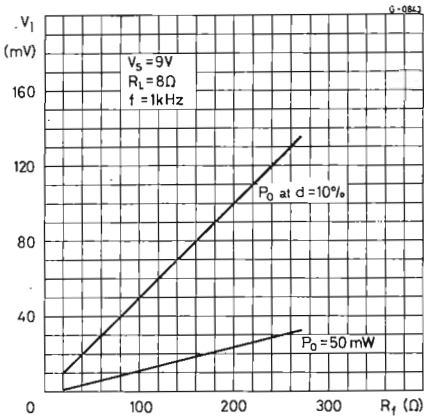


Fig. 10 - Typical voltage gain (closed loop)

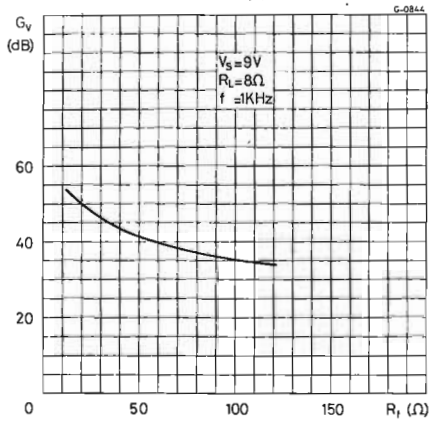


Fig. 11 - Typical distortion

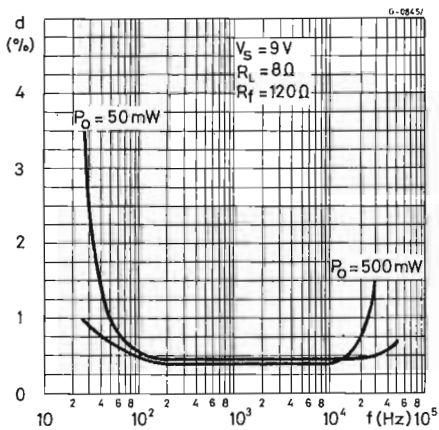
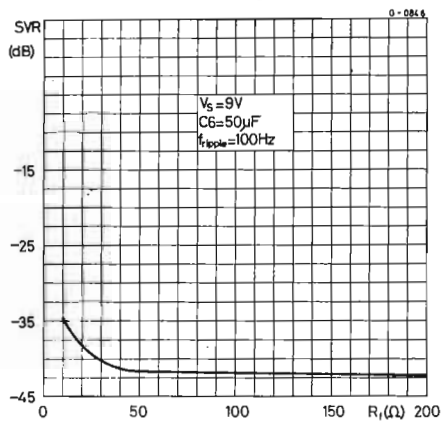


Fig. 12 - Typical supply voltage rejection (fig. 2 circuit)



TBA 820

Fig. 13 - Quiescent output voltage at pin 12

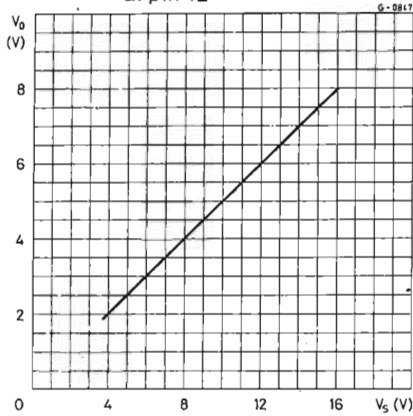


Fig. 14 - Quiescent current

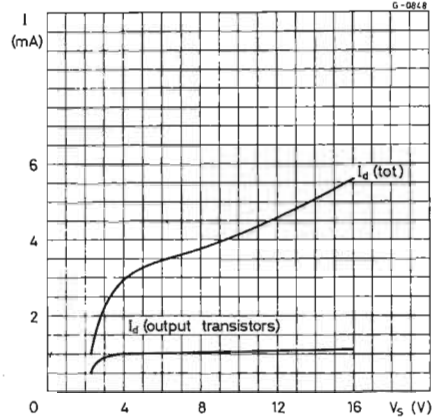
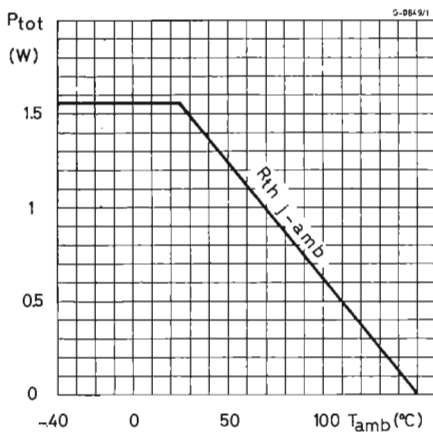


Fig. 15 - Power rating chart



NOTE: Mounting compatibility with TAA 611 provided that P.C. board strips of pins 2 and 13 are disconnected.

TCA 511

LINEAR INTEGRATED CIRCUIT

PRELIMINARY DATA

TV HORIZONTAL AND VERTICAL PROCESSOR

The TCA 511 is a silicon monolithic integrated circuit in a 16-lead dual in-line plastic package. It incorporates the following functions: high stability horizontal oscillator, horizontal APC circuit with high noise immunity and large pull-in range, high stability vertical oscillator and sawtooth generator.

It is intended for driving TV horizontal and vertical transistorized output stages.

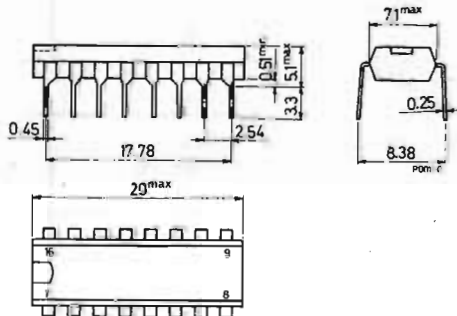
ABSOLUTE MAXIMUM RATINGS

V_s	Vertical section supply voltage (between pins 3 and 13)	15 V
V_s	Horizontal section supply voltage (between pins 4 and 13)	15 V
V_7, V_{12}, V_{15}	Pin 7, 12, 15 voltage (collector to ground)	15 V
V_i	Vertical sync. input voltage (between pins 2 and 13 - see note)	-5 V
V_i	Horizontal sync. input voltage (between pins 6 and 13 - see note)	-5 V
I_B	DC current (from pin 8)	30 mA
I_{12}, I_{14}, I_{15}	Peak current (into pins 12, 14 and 15)	50 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 60^\circ\text{C}$	500 mW
T_{stg}	Storage temperature	-55 to 125 °C
→ T_{op}	Operating temperature	0 to 60 °C

NOTE: The positive input voltage at pin 2 and pin 6 must not be greater than the voltage at pin 3 and pin 4 respectively.

MECHANICAL DATA

Dimensions in mm



TCA 511

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25^{\circ}\text{C}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
-----------	-----------------	------	------	------	------	------

VERTICAL SECTION

I_3	Quiescent current	$V_s = 12\text{ V}$ $f = 50\text{ Hz}$		4	mA	2
V_3^*	Supply voltage			9	V	
V_1	Peak to peak oscillator sawtooth voltage	$V_s = 12\text{ V}$ $f = 50\text{ Hz}$		2.6	V	
V_2	Peak sync. input voltage	$V_s = 12\text{ V}$ $f = 50\text{ Hz}$		3	V	3
V_{14}	Low level output voltage	$V_s = 12\text{ V}$		1.5	V	
V_{15}	Low level output voltage	$I_{15} = 15\text{ mA}$		0.5	V	
R_2	Parallel input resistance at pin 2	$V_s = 12\text{ V}$ $V_2 = 3\text{ V}$		50	k Ω	—
t^{**}	Output pulse width at pin 15	$V_s = 12\text{ V}$ $f = 50\text{ Hz}$ $R_{10} = 15\text{ k}\Omega$		0.75	ms	2
Δf	Locking range	$V_s = 12\text{ V}$ $f = 50\text{ Hz}$		-17	%	
$\frac{\Delta f}{\Delta T_{amb}}$	Frequency/temperature coefficient	$V_s = 12\text{ V}$ $T_{amb} = 20\text{ to }70^{\circ}\text{C}$		-0.015	$\frac{\text{Hz}}{^{\circ}\text{C}}$	

HORIZONTAL SECTION

I_4	Quiescent current	$V_s = 12\text{ V}$ $f = 15625\text{ Hz}$ $R_{11-13} = 0$		19	mA	2
V_4^*	Supply voltage			9	V	
V_6	Peak sinc. input voltage	$V_s = 12\text{ V}$ $f = 15625\text{ Hz}$		3	V	

TCA 511

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
→ V_8	Regulated output voltage	$V_s = 12\text{ V}$ $f = 15625\text{ Hz}$	7.5		V	2
V_{10}	Peak to peak oscillator sawtooth voltage		3.3		V	
V_{12}	Low level output voltage	$V_s = 12\text{ V}$ $I_{12} = 15\text{ mA}$		0.45	V	3
R_6	Parallel input resistance at pin 6	$V_s = 12\text{ V}$ $V_6 = 3\text{ V}$	50		k Ω	—
t^{***}	Output pulse width at pin 12	$V_s = 12\text{ V}$ $f = 15625\text{ Hz}$ a) $R_{11-13} = 0$ b) $R_{11-13} = \infty$	13	35	μs μs	2
t_d	Leading edge of output pulse to leading edge of sync. pulse phasing	$V_s = 12\text{ V}$ $f = 15625\text{ Hz}$	4		μs	
Δf	Pull-in range		± 1.3		kHz	
Δf	Hold-in range		± 1.4		kHz	
$\frac{\Delta f}{\Delta V_9}$	Oscillator control sensitivity		$V_s = 12\text{ V}$	10		
$\frac{\Delta f}{\Delta t_d}$	APC loop gain	2			$\frac{\text{kHz}}{\mu\text{s}}$	
$\frac{\Delta f}{\Delta V_s}$	Oscillator frequency drift	$V_s = 9\text{ to }14\text{ V}$	+ 0.7		$\frac{\%}{\text{V}}$	
$\frac{\Delta f}{\Delta T_{\text{amb}}}$	Frequency/temperature coefficient	$V_s = 12\text{ V}$ $T_{\text{amb}} = 20\text{ to }70\text{ }^\circ\text{C}$	+ 5		$\frac{\text{Hz}}{^\circ\text{C}}$	

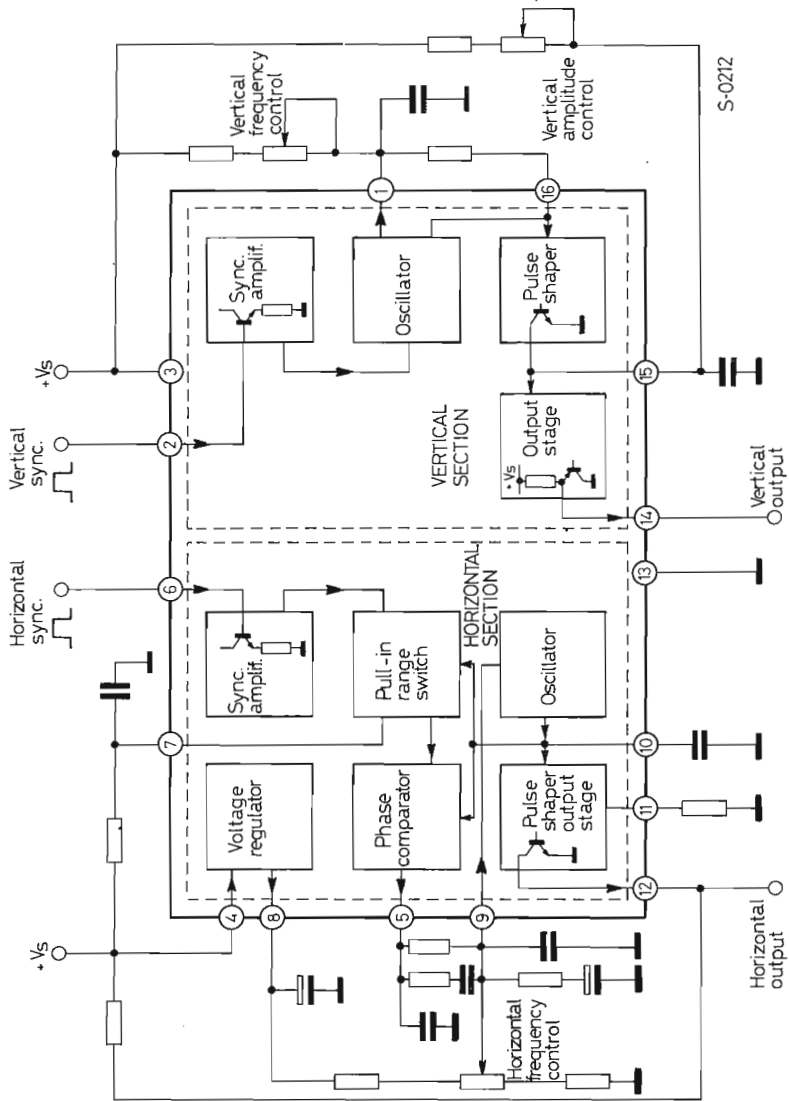
NOTES: * Minimum supply voltage for correct operation of the device.

** The output pulse width can be adjusted by means of the external resistance connected between pins 1 and 6.

*** The output pulse width can be adjusted by means of the external resistance or by a voltage $\leq 5.3\text{ V}$, connected between pin 11 and pin 13.

TCA 511

Fig. 1 - Functional block diagram



TCA 511

Fig. 2 - Test circuit

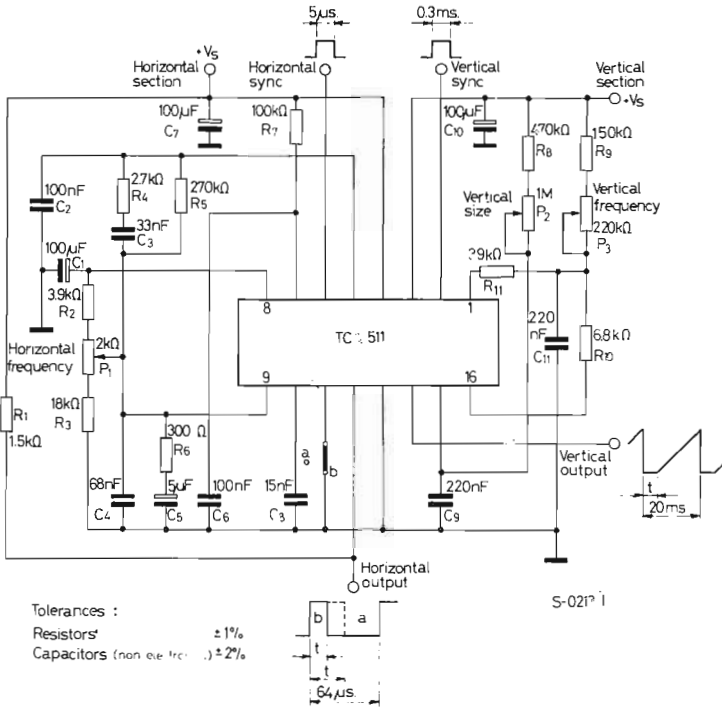
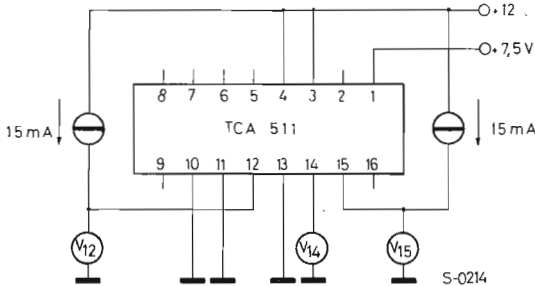
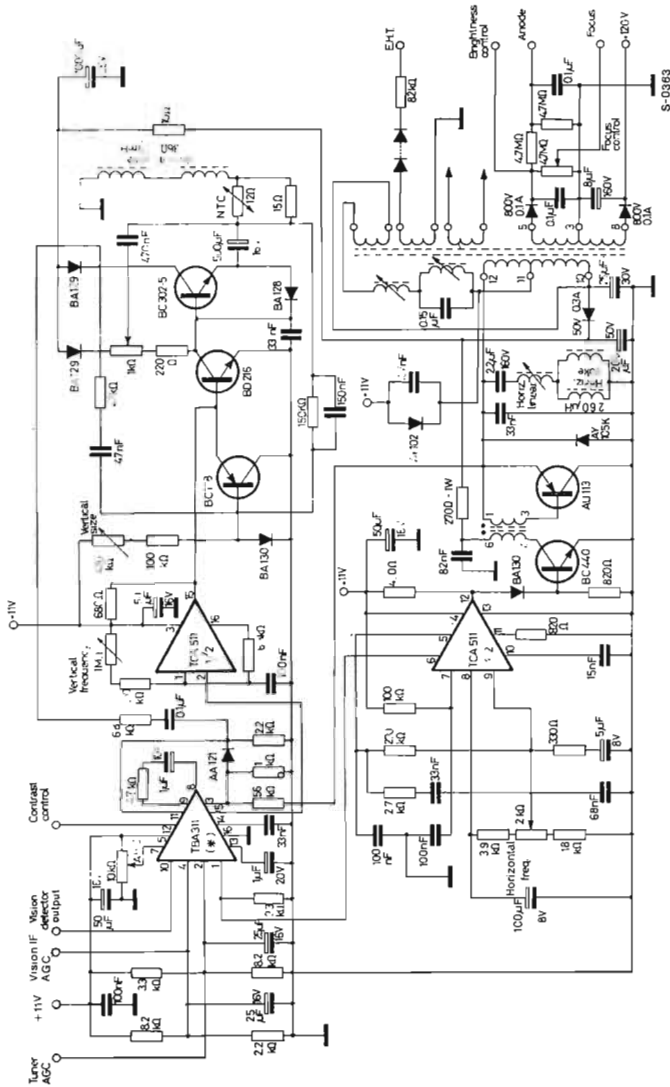


Fig. 3 - V_{12} , V_{14} and V_{15} test circuit



TCA 511

Fig. 4 - Typical application circuit for 12"110° TV set



(*) The jungle circuit TBA 311 performs the following functions:
 video preamplifier, IF AGC, PNP and NPN tuner AGC, sync. separator, noise gate.
 It is particularly suitable for driving the TCA 511 sync. inputs.

APPLICATION INFORMATION

Power Supply

The circuit can work with stabilized supply voltage having a value from 9 to 15 V. A dropping resistor and a filter capacitor may be used to obtain the supply from higher voltages; however, the voltage on pins 3 and 4 must never exceed the maximum permitted voltage.

Synchronization

Pins 2 and 6 can be DC driven if the reference level of the synchronization pulses is less than 1 V. With reference levels greater than this value, a coupling capacitor must be inserted in series with the input, and pins 2 and 6 must be connected to ground via a resistor.

Vertical Oscillator

The capacitor connected to pin 1 must be selected with regard to the frequency tolerance, to the thermal stability and to the capacitor's ageing.

The width of the output pulse, to be chosen according to the needs of the output stages, is defined by the resistor connected between pin 1 and pin 16.

Vertical Output

The vertical output is taken from pin 14, which is a buffered output of the sawtooth voltage generated at pin 15.

The output current from pin 14 is defined by an internal resistor in the integrated circuit. If a greater current is needed, a resistor may be connected between pin 14 and pin 3.

The oscillator output pulse is available at pin 15 if the capacitor C9 is not connected. This configuration is used for driving output stages in which the sawtooth is generated by Miller effect.

Horizontal Oscillator

The capacitor connected between pin 10 and ground must be selected with regard to the frequency tolerance, to the thermal stability and to the capacitor's ageing.

In multistandard receivers, the oscillation frequency may be changed by switching the value of the capacitor connected to pin 10.

TCA 511

APPLICATION INFORMATION (continued)

Phase Comparator

The phase comparator's output consists of current pulses acting on the oscillator control voltage.

The external components C2, C3, C4, C5, R4, R5 and R6 (fig. 2) define the circuit performance with respect to the pull-in range, the hold-in range and the frequency variations that occur on switching-on and switching-off.

Moreover the pull-in range depends on the absolute value of the voltage divider R2, P1 and R3.

A coincidence detector is connected to pin 7; this modifies the pull-in range and the noise immunity, depending on whether the system is synchronised or is searching for synchronization. The time constant applied to pin 7 avoids uncertainty during the switch from one state to the other.

Horizontal Output

The collector of the output transistor is connected to pin 12; its load resistor, externally connected between pin 12 and pin 4, defines the amplitude of the output current pulse.

The width of the output pulse can be varied between 13 and 35 μs by means of the resistor connected between pin 11 and ground, or else by means of a voltage ≤ 5.3 V applied between pin 11 and ground. This control acts upon the trailing edge of the pulse, hence the phase advance of the leading edge stays constant with respect to the synchronism.

LINEAR INTEGRATED CIRCUITS

TCA 600 TCA 610

PRELIMINARY DATA

MOTOR SPEED REGULATORS

The TCA 600 and TCA 610 are integrated circuits in Jedec TO-39 metal case. They are designed for use as speed regulators for DC motors of record players, tape recorders and cassettes.

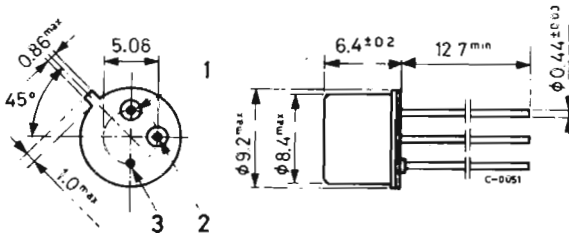
The TCA 600 is particularly suitable for battery operated portable equipments, and the TCA 610 for car-battery and mains operations.

ABSOLUTE MAXIMUM RATINGS

		TCA 600	TCA 610
V_s	Supply voltage	14 V	20 V
P_{tot}	Total power dissipation at $T_{amb} = 55^\circ\text{C}$ at $T_{case} = 75^\circ\text{C}$		0.55 W 3 W
T_{stg}	Storage temperature	-55 to 150 °C	
T_j	Junction temperature	150 °C	

MECHANICAL DATA

Dimensions in mm



TO-39

TCA 600

TCA 610

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	Typ.	25	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	Typ.	175	°C/W

ELECTRICAL CHARACTERISTICS

($T_{amb} = 25\text{°C}$ and $R_s = \infty$ unless otherwise indicated)

	Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
→	V_{ref} Reference voltage (between pins 2 and 3)	$V_s = 5.5\text{ V}$ $I_m = 70\text{ mA}$ $R_T = 0$		2.6		V	1
→	I_{d3} Quiescent current (at pin 3)	$V_{1-3} = 5.5\text{ V}$ $I_2 = 0$ $R_T = 0$		2.6		mA	—
→	V_m Output voltage (for TCA 600 only)	$V_s = 5.5\text{ V}$ $I_m = 70\text{ mA}$ $R_T = 91\ \Omega$		3.6	3.9	V	1
→	V_m Output voltage (for TCA 610 only)	$V_s = 9\text{ V}$ $I_m = 70\text{ mA}$ $R_T = 270\ \Omega$		5.6		V	1
	V_{1-2} Dropout voltage	$\Delta V_m/V_m = -1\%$ $I_m = 70\text{ mA}$ $R_T = 91\ \Omega$		1.2		V	1
→	I_2 Limiting output current (at pin 2)	$V_{1-3} = 5.5\text{ V}$ $V_{2-3} = 0$		400		mA	—
→	$k = \Delta I_2/\Delta I_3$	$V_s = 5.5\text{ V}$ $I_2 = -70\text{ mA}$ $\Delta I_2 = \pm 10\text{ mA}$ $R_T = 0$		8.5		—	1
	$\frac{\Delta V_m}{V_m}/\Delta V_s$ Line regulation (for TCA 600 only)	$V_s = 5.5\text{ V to }12\text{ V}$ $I_m = 70\text{ mA}$ $R_T = 91\ \Omega$		0.1		%/V	1

TCA 600 TCA 610

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
$\frac{\Delta V_m}{V_m} / \Delta V_s$ Line regulation (for TCA 610 only)	$V_s = 10 \text{ V to } 16 \text{ V}$ $I_m = 70 \text{ mA}$ $R_T = 270 \Omega$		0.1		%/V	1
$\frac{\Delta V_m}{V_m} / \Delta I_m$ Load regulation	$V_s = 5.5 \text{ V}$ $I_m = 40 \text{ to } 100 \text{ mA}$ $R_T = 0$		0.005		%/mA	1
$\frac{\Delta V_{ref}}{V_{ref}} / \Delta T_{amb}$ Temperature coefficient	$V_{1-3} = 5.5 \text{ V}$ $I_2 = -70 \text{ mA}$ $T_{amb} = -20 \text{ to } 70 \text{ }^\circ\text{C}$		0.01		%/°C	—

Fig. 1 - Test circuit.

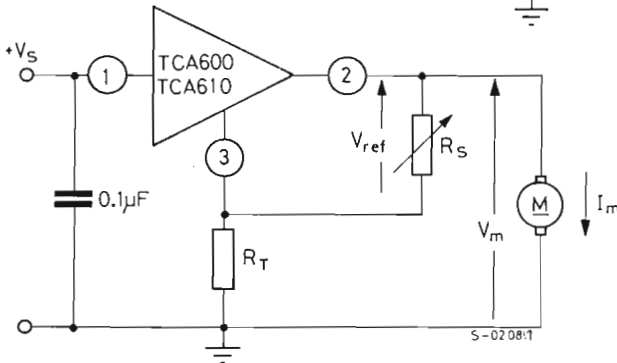
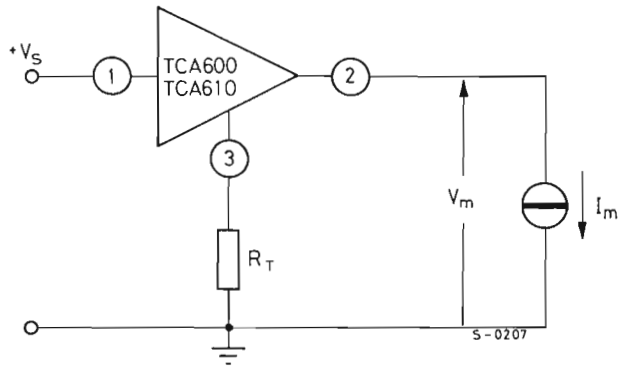


Fig. 2 - Typical application circuit.

TCA 600 TCA 610

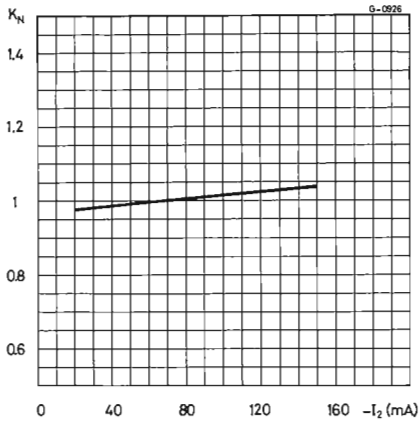


Fig. 3 - Normalized k versus I_2 .

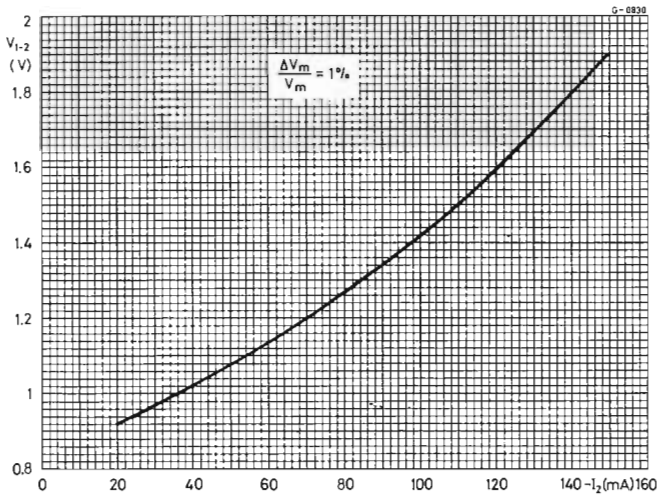
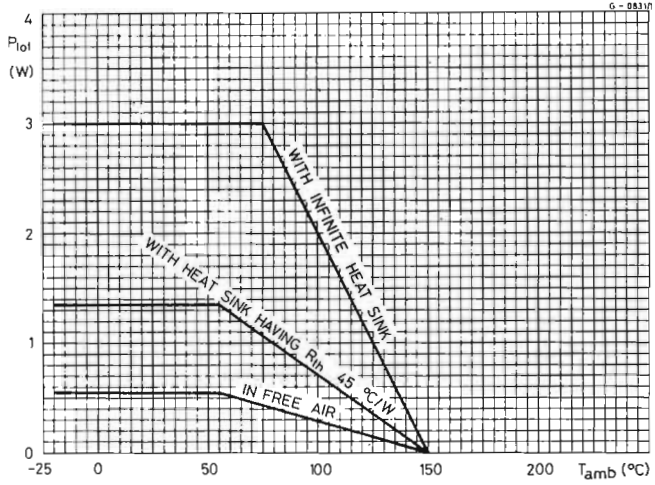


Fig. 4 - Dropout voltage versus output current.

Fig. 5 - Maximum allowable power dissipation versus ambient temperature.



APPLICATION INFORMATION

The regulator supplies the motor in such a way as to keep its speed constant, independent of supply voltage, applied torque and ambient temperature variations. The basic equation for the motor is:

$$V_m = E_0 + R_m I_m = a_1 n + a_2 c$$

- Where:
- V_m = supply voltage applied to the motor
 - E_0 = back electromotive force
 - n = motor speed (r.p.m)
 - R_m = internal resistance (of the motor)
 - I_m = current absorbed (by the motor)
 - a_1 and a_2 = constants
 - c = drive torque

TCA 600 TCA 610

A voltage supply with the following characteristics

$$E = E_0 \quad E = \text{electromotive force}$$

$$R_o = -R_m \quad R_o = \text{output resistance}$$

gives performance required.

This means that a variation in current absorbed by the motor, due to a variation in torque applied, causes a proportional variation in regulator output voltage.

In fig. 6 is shown the minimum allowable E_0 versus R_T .

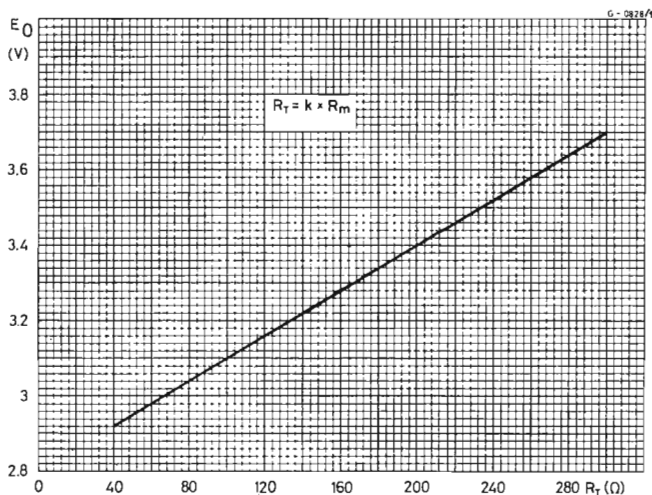


Fig. 6 - Minimum E_0 allowable versus R_T .

The TCA 600 and TCA 610 give a reference constant voltage V_{ref} (between pins 2 and 3) independent of variations of V_s , I_2 and ambient temperature.

They also give:

$$I_3 = I_{d3} + I_2/k$$

Where:

- I_3 = total current at pin 3
- I_{d3} = quiescent current at pin 3 ($I_2 = 0$)
- I_2 = current at pin 2
- k = constant.

The output voltage V_m , applied to the motor has the following value:

$$V_m = \underbrace{V_{ref} + R_T \left[\frac{V_{ref}}{R_s} \left(1 + \frac{1}{k} \right) + I_{d3} \right]}_{\text{Term 1}} + \underbrace{\frac{I_m}{k} R_T}_{\text{Term 2}}$$

Term 1 equals E_0 and fixes the motor speed by means of the variable resistor R_s ;

Term 2 $\frac{I_m}{k} \cdot R_T$ equals the term $R_m \cdot I_m$ and, therefore, compensates variations of torque applied.

Complete compensation is achieved when:

$$R_T = k R_m$$

If $R_{T \max} > k R_{m \min}$ instability may occur.

LINEAR INTEGRATED CIRCUITS

TCA 900 TCA 910

PRELIMINARY DATA

MOTOR SPEED REGULATORS

The TCA 900 and TCA 910 are linear integrated circuits in Jedec TO-126 plastic package. They are designed for use as speed regulators for DC motors of record players, tape recorders and cassettes.

The TCA 900 is particularly suitable for battery operated portable equipments, and the TCA 910 for car-battery and mains operations.

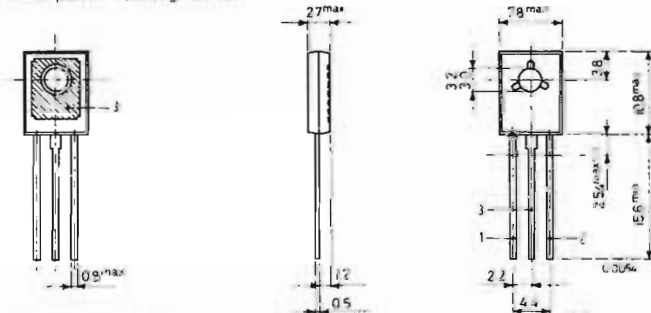
ABSOLUTE MAXIMUM RATINGS

		TCA 900	TCA 910
V_s	Supply voltage	14 V	20 V
P_{tot}	Total power dissipation at $T_{amb} = 70^\circ\text{C}$ at $T_{case} = 100^\circ\text{C}$		0.8 W 5 W
T_{stg}	Storage temperature	-55 to 150°C	
T_j	Junction temperature	150 $^\circ\text{C}$	

MECHANICAL DATA

Dimensions in mm

Pin 3 connected to metal part of mounting surface



TO-126 (SOT-32)

(1) Within this region the cross-section of the leads is uncontrolled

TCA 900

TCA 910

THERMAL DATA

$R_{th\ j-case}$	Thermal resistance junction-case	Typ.	10	$^{\circ}C/W$
$R_{th\ j-amb}$	Thermal resistance junction-ambient	Typ.	100	$^{\circ}C/W$

ELECTRICAL CHARACTERISTICS ($T_{amb} = 25^{\circ}C$ and $R_s = \infty$ unless otherwise specified)

	Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
→	V_{ref} Reference voltage (between pins 2 and 3)	$V_s = 5.5\ V$ $I_m = 70\ mA$ $R_T = 0$		2.6		V	1
→	I_{d3} Quiescent current (at pin 3)	$V_{1,3} = 5.5\ V$ $I_2 = 0$ $R_T = 0$		2.6		mA	—
→	V_m Output voltage (for TCA 900 only)	$V_s = 5.5\ V$ $I_m = 70\ mA$ $R_T = 91\ \Omega$		3.6	3.9	V	1
	V_m Output voltage (for TCA 910 only)	$V_s = 9\ V$ $I_m = 70\ mA$ $R_T = 270\ \Omega$		5.6	6.3	V	1
	$V_{1,2}$ Dropout voltage	$\Delta V_m/V_m = -1\%$ $I_m = 70\ mA$ $R_T = 91\ \Omega$		1.2		V	1
	I_2 Limiting output current (at pin 2)	$V_{1,3} = 5.5\ V$ $V_{2,3} = 0$		400		mA	—
	$k = \Delta I_2/\Delta I_3$	$V_s = 5.5\ V$ $I_2 = -70\ mA$ $\Delta I_2 = \pm 10\ mA$ $R_T = 0$		8.5		—	1
	$\frac{\Delta V_m}{V_m}/\Delta V_s$ Line regulation (for TCA 900 only)	$V_s = 5.5\ V$ to $12\ V$ $I_m = 70\ mA$ $R_T = 91\ \Omega$		0.1		%/V	1

TCA 900 TCA 910

ELECTRICAL CHARACTERISTICS (continued)

Parameter	Test conditions	Min.	Typ.	Max.	Unit	Fig.
$\frac{\Delta V_m}{V_m} / \Delta V_s$ Line regulation (for TCA 910 only)	$V_s = 10 \text{ V to } 16 \text{ V}$ $I_m = 70 \text{ mA}$ $R_T = 270 \Omega$		0.1		%/V	1
$\frac{\Delta V_m}{V_m} / \Delta I_m$ Load regulation	$V_s = 5.5 \text{ V}$ $I_m = 40 \text{ to } 100 \text{ mA}$ $R_T = 0$		0.005		%/mA	1
$\frac{\Delta V_{ref}}{V_{ref}} / \Delta T_{amb}$ Temperature coefficient	$V_{1,3} = 5.5 \text{ V}$ $I_2 = -70 \text{ mA}$ $T_{amb} = -20 \text{ to } 70 \text{ }^\circ\text{C}$		0.01		%/°C	—

Fig. 1 - Test circuit.

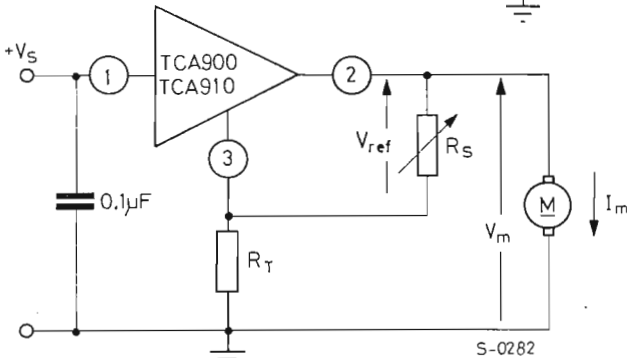
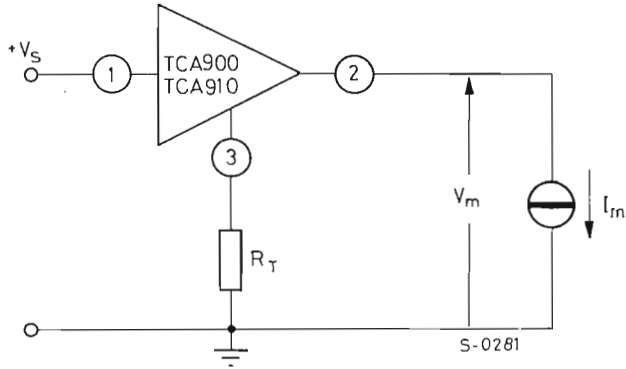


Fig. 2 - Typical application circuit.

TCA 900 TCA 910

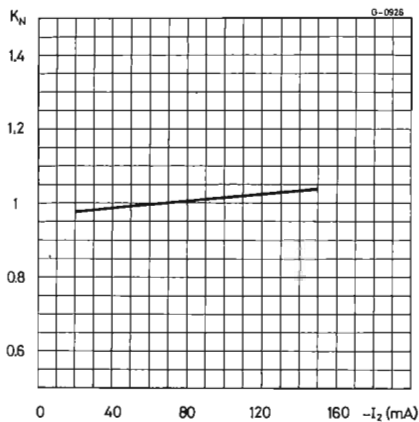


Fig. 3 - Normalized k versus I_2 .

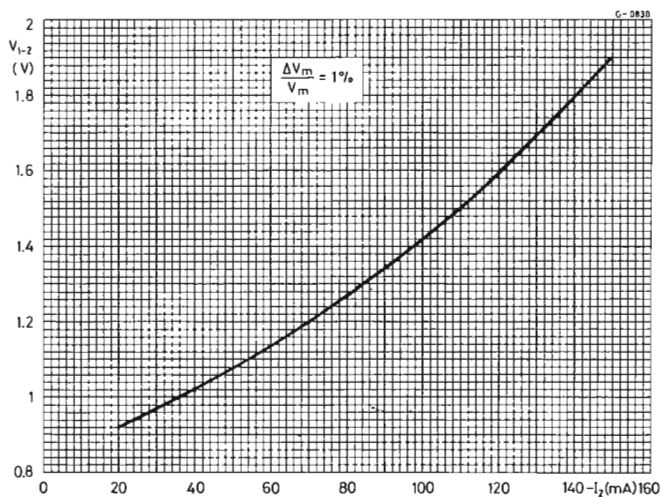
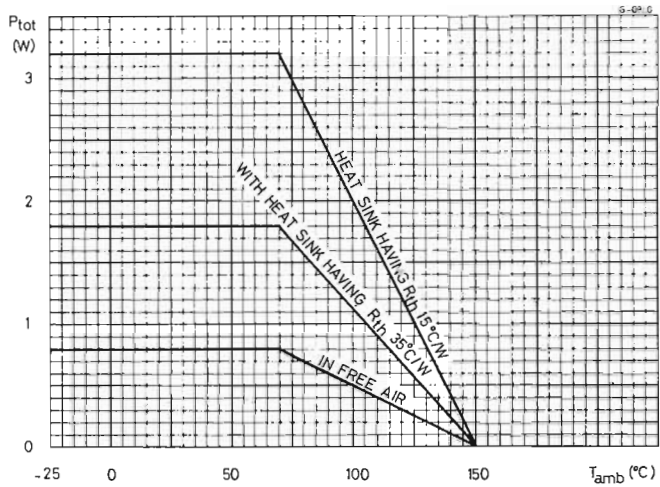


Fig. 4 - Dropout voltage versus output current.

TCA 900 TCA 910

Fig. 5 - Maximum allowable power dissipation versus ambient temperature.



APPLICATION INFORMATION

The regulator supplies the motor in such a way as to keep its speed constant, independent of supply voltage, applied torque and ambient temperature variations. The basic equation for the motor is:

$$V_m = E_0 + R_m I_m = a_1 n + a_2 c$$

- Where:
- V_m = supply voltage applied to the motor
 - E_0 = back electromotive force
 - n = motor speed (r.p.m)
 - R_m = internal resistance (of the motor)
 - I_m = current absorbed (by the motor)
 - a_1 and a_2 = constants
 - c = drive torque

TCA 900 TCA 910

A voltage supply with the following characteristics

$$E = E_0 \quad E = \text{electromotive force}$$

$$R_o = -R_m \quad R_o = \text{output resistance}$$

gives performance required.

This means that a variation in current absorbed by the motor, due to a variation in torque applied, causes a proportional variation in regulator output voltage.

In fig. 6 is shown the minimum allowable E_0 versus R_T .

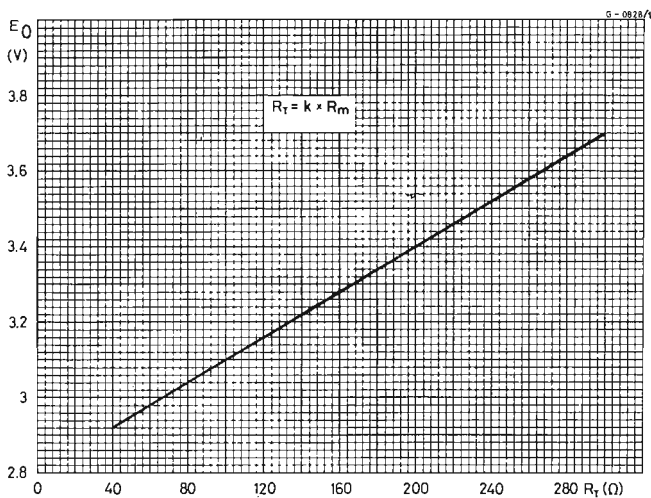


Fig. 6 - Minimum E_0 allowable versus R_T .

TCA 900 TCA 910

The TCA 900 and TCA 910 give a reference constant voltage V_{ref} (between pins 2 and 3) independent of variations of V_s , I_2 and ambient temperature.

They also give:

$$I_3 = I_{d3} + I_2/k$$

Where: I_3 = total current at pin 3
 I_{d3} = quiescent current at pin 3 ($I_2 = 0$)
 I_2 = current at pin 2
 k = constant.

The output voltage V_m , applied to the motor has the following value:

$$V_m = V_{ref} + R_T \underbrace{\left[\frac{V_{ref}}{R_s} \left(1 + \frac{1}{k} \right) + I_{d3} \right]}_{\text{Term 1}} + \underbrace{\frac{I_m}{k} R_T}_{\text{Term 2}}$$

Term 1 equals E_0 and fixes the motor speed by means of the variable resistor R_s ;

Term 2 $\frac{I_m}{k} \cdot R_T$ equals the term $R_m \cdot I_m$ and, therefore, compensates variations of torque applied.

Complete compensation is achieved when:

$$R_T = k R_m$$

If $R_{T_{max}} > k R_{m_{min}}$ instability may occur.

LINEAR INTEGRATED CIRCUIT

TDA 1200

PRELIMINARY DATA

FM-IF RADIO SYSTEM

- HIGH LIMITING SENSITIVITY
- HIGH AMR
- HIGH RECOVERED AUDIO
- GOOD CAPTURE RATIO
- LOW DISTORTION
- MUTING CAPABILITY

The TDA 1200 is a silicon monolithic integrated circuit in a 16-lead dual in-line plastic package. It provides a complete subsystem for amplification of FM signals.

The functions incorporated are:

- FM amplification and detection
- interchannel controlled muting
- AFC and delayed AGC for FM tuner
- switching of stereo decoder
- driving of a field strength meter

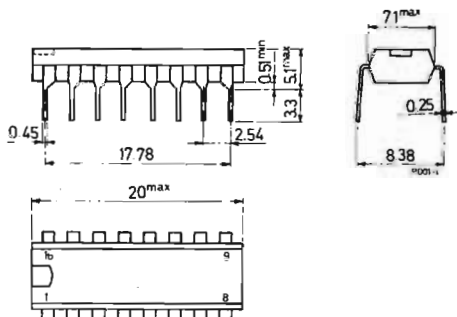
The TDA 1200 can be used for FM-IF amplifier application in HI-FI, car-radios and communication receivers.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	16 V
I_o	Output current (from pin 15)	2 mA
P_{tot}	Total power dissipation at $T_{amb} \leq 70^\circ\text{C}$	500 mW
T_{stg}	Storage temperature	-55 to 150 °C
T_{op}	Operating temperature	-25 to 70 °C

MECHANICAL DATA

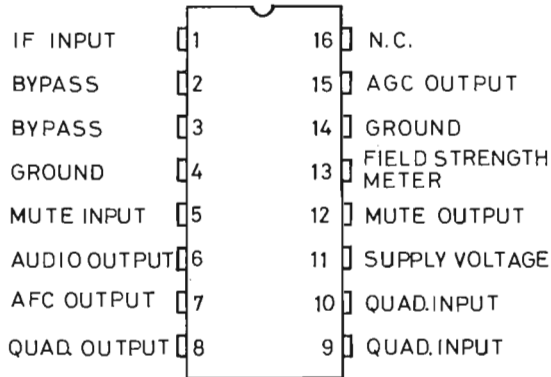
Dimensions in mm



TDA 1200

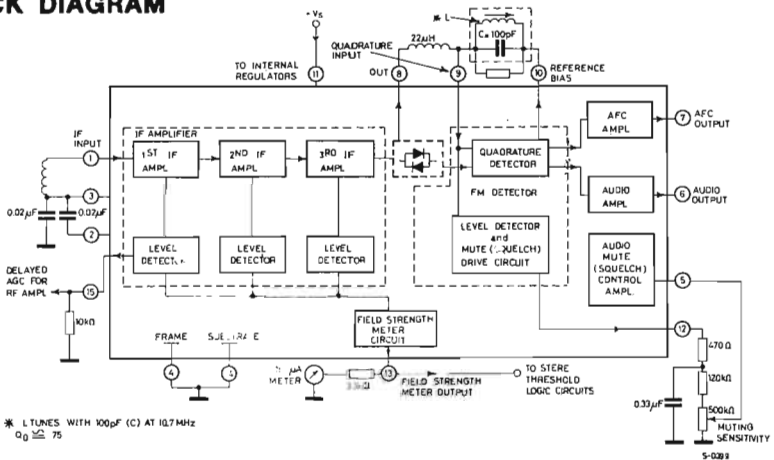
CONNECTION DIAGRAM

(top view)



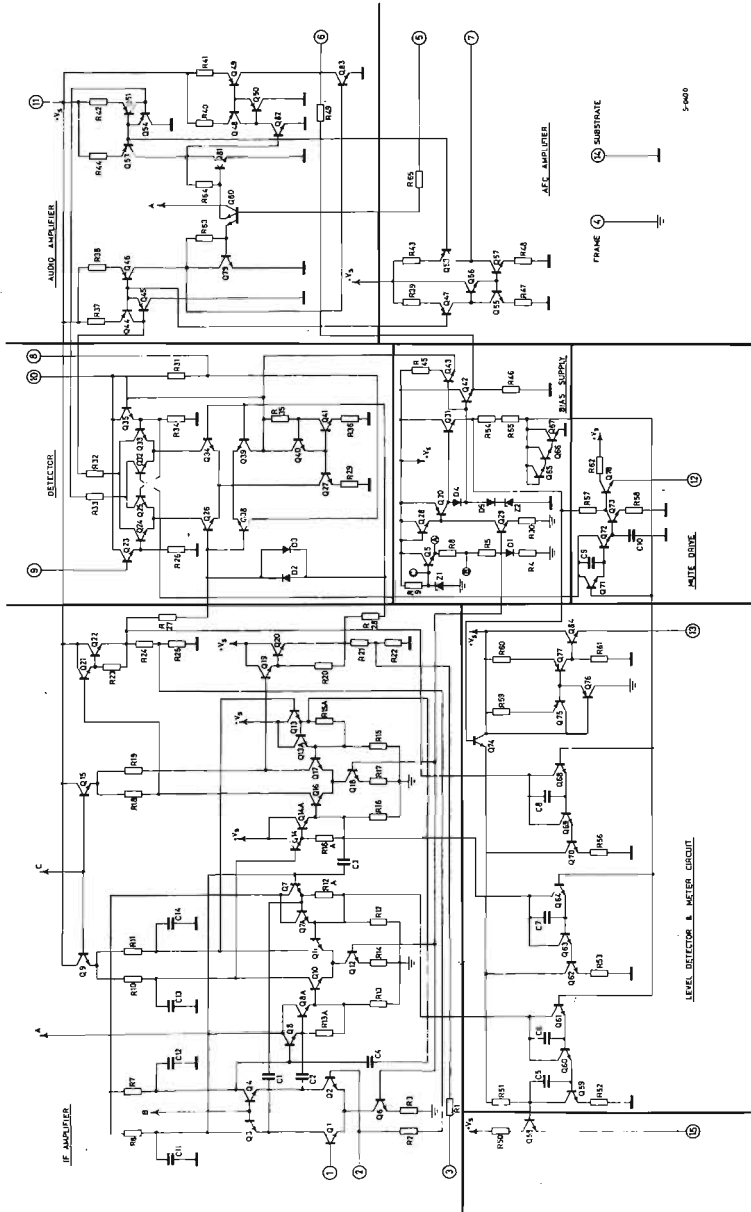
S-0398

BLOCK DIAGRAM



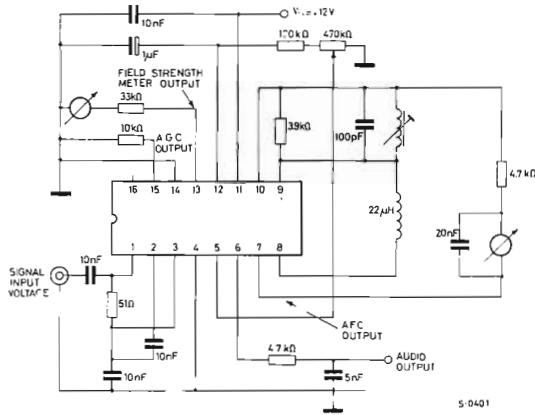
SCHEMATIC DIAGRAM

TDA 1200



TDA 1200

TEST CIRCUIT



THERMAL DATA

$R_{th\ j-amb}$ Thermal resistance junction-ambient	Typ. 160 °C/W
---	---------------

ELECTRICAL CHARACTERISTICS

(Refer to the test circuit; $V_s = 12\text{ V}$, $T_{amb} = 25\text{ °C}$)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
-----------	-----------------	------	------	------	------

STATIC (DC) CHARACTERISTICS

I_s	Supply current		23		mA
V_1	Voltage at the IF amplifier input		1.9		V
V_2, V_3	Voltage at the input bypassing		1.9		V
V_6	Voltage at the audio output		5.6		V
V_{10}	Reference bias voltage		5.6		V

TDA 1200

ELECTRICAL CHARACTERISTICS (continued)

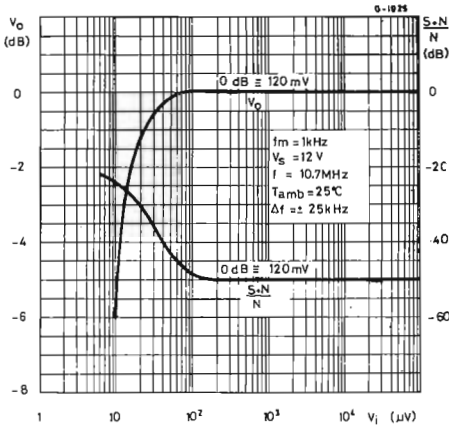
Parameter	Test conditions	Min.	Typ.	Max.	Unit
-----------	-----------------	------	------	------	------

DYNAMIC CHARACTERISTICS

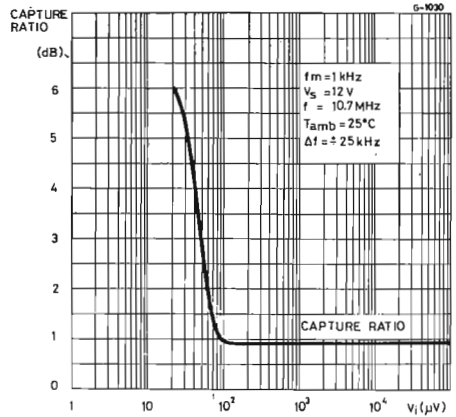
$V_{i(\text{threshold})}$	Input limiting voltage (-3 dB) at pin 1	$f = 10.7 \text{ MHz}$ $f_m = 1 \text{ kHz}$ $\Delta f = \pm 25 \text{ kHz}$		12		μV
V_o	Recovered audio voltage (pin 6)	$V_i \geq 50 \mu\text{V}$ $f = 10.7 \text{ MHz}$ $f_m = 1 \text{ kHz}$ $\Delta f = \pm 25 \text{ kHz}$		140		mV
d	Distortion	$V_i \geq 1 \text{ mV}$ $f = 10.7 \text{ MHz}$		0.5		%
$\frac{S+N}{N}$	Signal and noise to noise ratio	$f_m = 1 \text{ kHz}$ $\Delta f = \pm 75 \text{ kHz}$		60		dB
AMR	Amplitude modulation rejection	$V_i \geq 1 \text{ mV}$ $f = 10.7 \text{ MHz}$ $f_m = 1 \text{ kHz}$ $\Delta f = \pm 25 \text{ kHz}$ $m = 0.3$		40		dB
V_i	Input voltage for delayed AGC action (pin 1)			10		mV
$\frac{\Delta V_{15}}{\Delta V_i}$	AGC control slope	$V_i \geq 10 \text{ mV}$ $f = 10.7 \text{ MHz}$		40		dB
$\frac{\Delta I_7}{\delta f}$	AFC control slope			1		$\frac{\mu\text{A}}{\text{kHz}}$
$\frac{\Delta V_{13}}{\Delta V_i}$	Field strength meter output slope			42		dB
V_{13}	Field strength meter output sensitivity	$V_i = 1 \text{ mV}$ $f = 10.7 \text{ MHz}$		1.7		V

TDA 1200

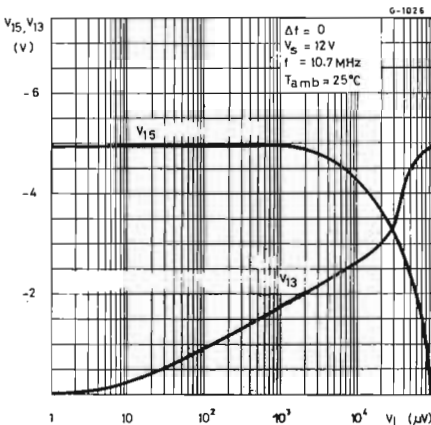
Typical recovered audio output and signal to noise ratio versus input voltage



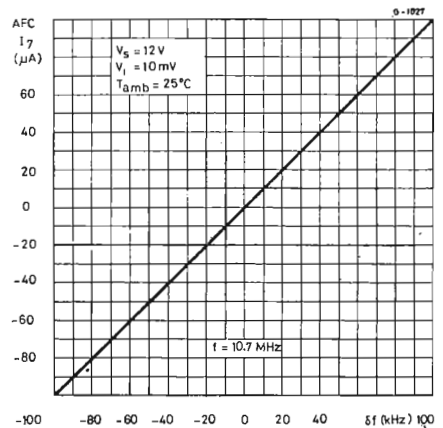
Typical capture ratio versus input voltage



Typical AGC (V_{15}) and field strength meter output (V_{13}) versus input signal

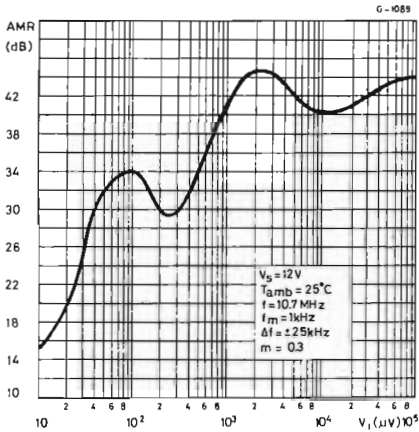


Typical AFC output current versus change-in tuning frequency

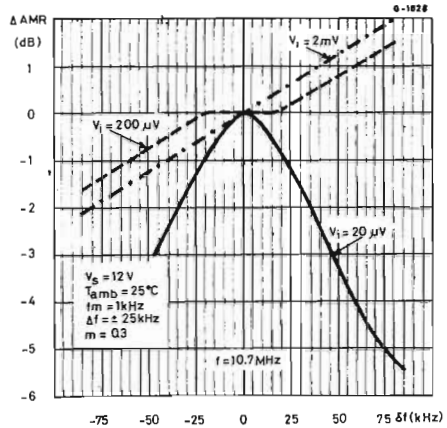


TDA 1200

Typical amplitude modulation rejection versus input signal

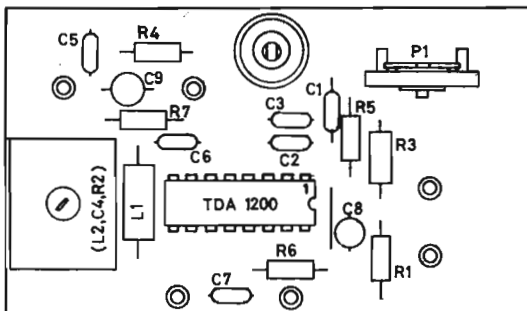


Typical AMR (relative to the value of $f = 10.7$ MHz) versus change-in tuning frequency



APPLICATIONS

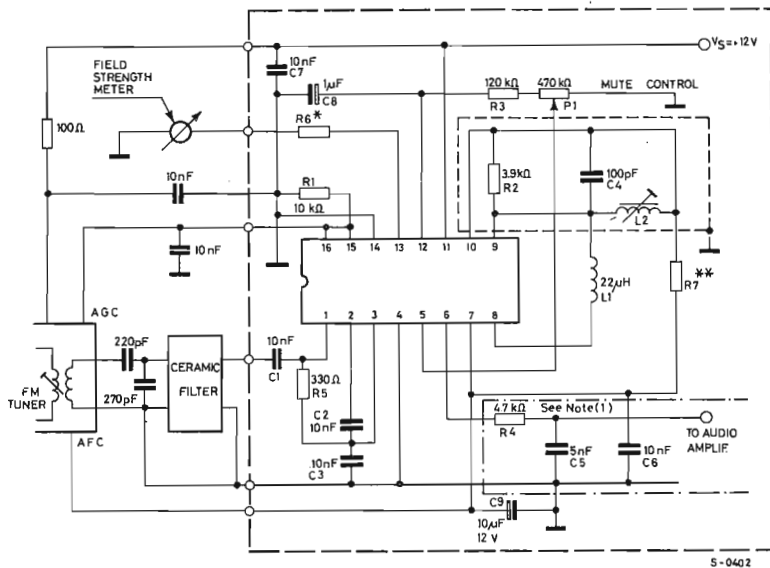
PC board and component layout of the circuit on next page (1:1 scale).



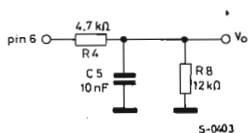
CS-0024

TDA 1200

Typical application circuit



NOTES: (1) When V_s is less than 12 V, a resistor $R_8 = 12\text{ k}\Omega$ must be connected between audio output and ground, and the integrator capacitor C_5 must be changed to 10 nF, as follows:



- * Dependent on field strength meter sensitivity.
- ** Dependent on the tuner's AFC circuit.

LINEAR INTEGRATED CIRCUIT

SAJ 210

7-STAGE FREQUENCY DIVIDER FOR ELECTRONIC ORGANS

- HIGH CROSSTALK IMMUNITY - TYP. 70 dB
- OUTPUT SHORT CIRCUIT PROTECTION

The SAJ 210 is a monolithic integrated circuit in a 14-lead quad in-line or dual in-line plastic package. It has been created by means of the standard bipolar technique and especially developed for use as frequency divider for electronic organs. Seven flip-flops connected in 5 groups are housed on one silicon chip. The input and the output of each flip-flop is externally accessible.

ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	14	V
V_i	Input voltage	$V_i = V_s$	
I_o^*	Output current	5	mA
P_{tot}	Power dissipation at $T_{amb} \leq 70^\circ\text{C}$	0.5	W
T_{stg}	Storage temperature	-55 to 125	$^\circ\text{C}$
T_{op}	Operating temperature	0 to 70	$^\circ\text{C}$

* With reference to Fig. 5, the current can be greater than 5 mA, but for $t < 0.1$ ms.

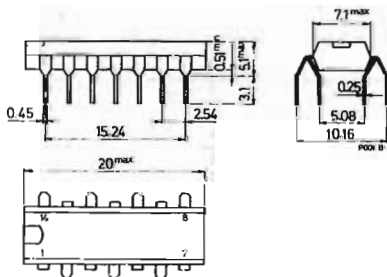
ORDERING NUMBERS:

SAJ 210 AX2 (for 14-lead quad in-line plastic package)

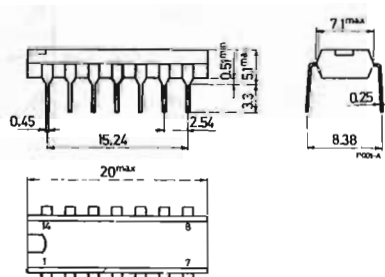
SAJ 210 AX7 (for 14-lead dual in-line plastic package)

MECHANICAL DATA

Dimensions in mm



SAJ 210 AX2

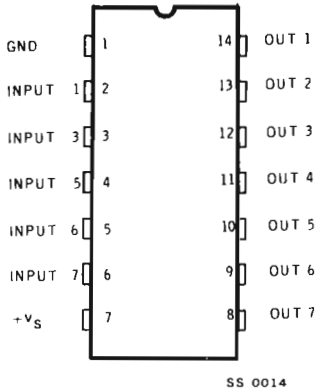


SAJ 210 AX7

SAJ 210

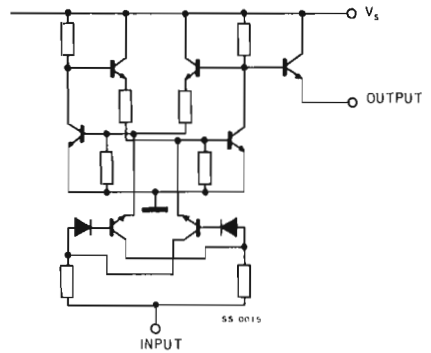
CONNECTION DIAGRAM

(top view)



SCHEMATIC DIAGRAM

(each flip-flop)



ELECTRICAL CHARACTERISTICS

($T_{amb} = 25^{\circ}\text{C}$, $V_s = 9\text{ V}$ unless otherwise specified)

Parameter	Test conditions	Min.	Typ.	Max.	Unit
-----------	-----------------	------	------	------	------

DATA INPUT

V_{iL}	Input low level	$V_s = 8$ to 14 V	0	1.5	V
V_{iH}^*	Input high level	$V_s = 8$ to 14 V	6		V
I_{iH}	Input high level current	$V_i = 8\text{ V}$	1	3	mA

DATA OUTPUT

V_{oL}	Output low level	$R_L = 3\text{ k}\Omega$		0.1	V
V	Output voltage impressed	Low level!		6	V
V_{oH}	Output high level	$R_L = 3\text{ k}\Omega$ $V_s = 12\text{ V}$	$R_L = 3\text{ k}\Omega$	7 9.5	V V
t_r	Rise time	$V_i = 8\text{ V}$	$C_L = 10\text{ pF}$	0.1	μs
t_f	Fall time	$R_L = 3\text{ k}\Omega$	$C_L = 10\text{ pF}$	0.2	μs
I_d	Total current drain	$R_L = 3\text{ k}\Omega$ All flip-flops at high level All flip-flops at low level		35 16	mA mA

SAJ 210

ELECTRICAL CHARACTERISTICS (continued)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
V_o	Output swing	$R_L = 3 \text{ k}\Omega$		7.6		V
**	Cross talk immunity level	$R_L = 3 \text{ k}\Omega$ $C_L = 10 \text{ pF}$		70		dB
R_o	Dynamic output resistance	V_o impressed = 0 to 2 V at high level at low level	1	160	6	Ω M Ω
	Ripple on output voltage at 2 f out	$V_i = 8 \text{ V}$		5		mV _{pp}

* Input high level is never reached if the input pulse is lower than 3.5 V.

** Two independent dividers are triggered

Divider A: triggering frequency 20 kHz

Divider B: triggering frequency 2240 Hz

V_o B 1120 Hz

Cross talk level = $20 \log \frac{V_o \text{ B } 1120 \text{ Hz}}{V_o \text{ B } 10 \text{ kHz}}$

Fig. 1 - Typical input current vs input voltage

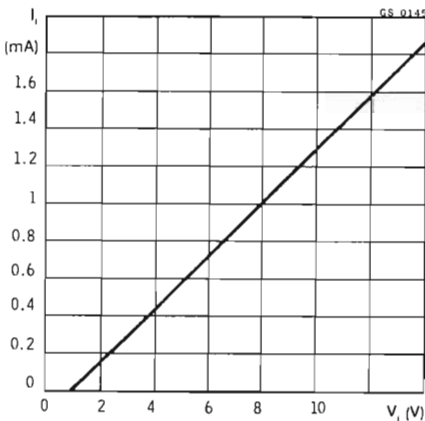
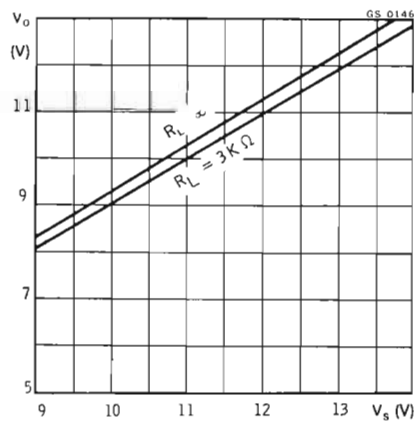


Fig. 2 - Typical output level vs supply voltage



SAJ 210

Fig. 3 - Typical input voltage for triggering

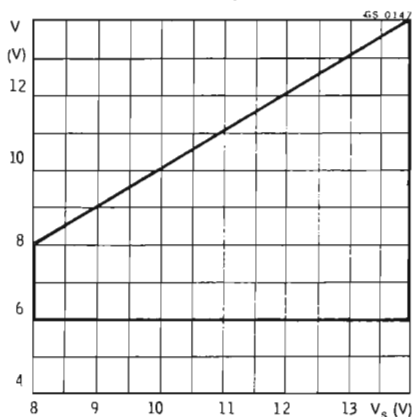


Fig. 4 - Power rating chart

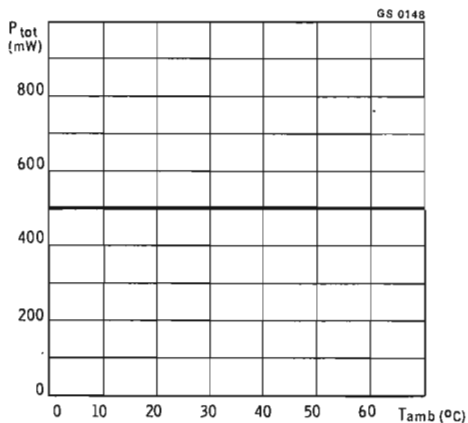
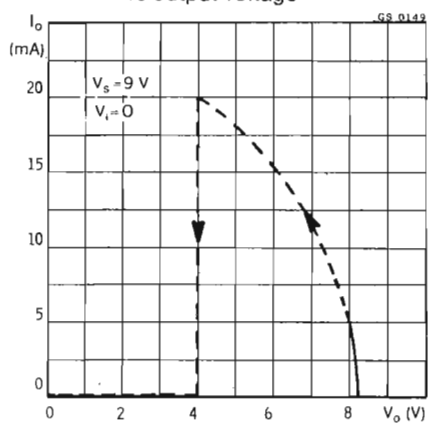
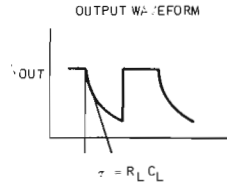
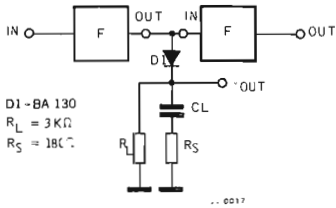
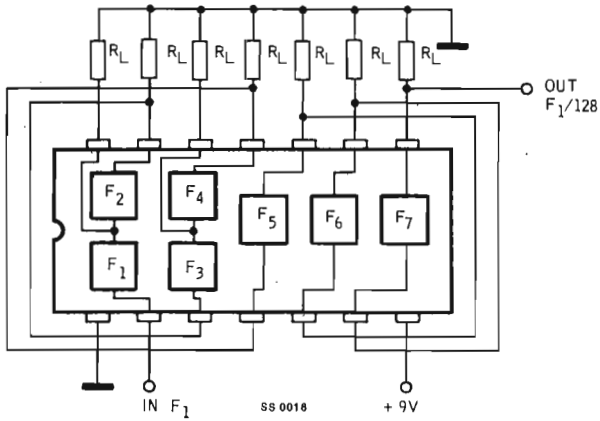


Fig. 5 - Typical output current vs output voltage



SAJ 210

TYPICAL APPLICATIONS



Off. Graf. Morell - Osnago (Como)
Printed in Italy

Information furnished by SGS-ATES is believed to be accurate and reliable. However, no responsibility is assumed by SGS-ATES for the consequences of its use nor for an infringement of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of SGS-ATES.

the 1990s, the number of people in the world who are undernourished has increased from 600 million to 800 million. The number of people who are malnourished has increased from 1.2 billion to 1.5 billion. The number of people who are obese has increased from 100 million to 300 million.

There are a number of reasons for this. One is that the world population has increased from 5 billion to 6 billion. Another is that the world has become more urbanized. A third is that the world has become more affluent. A fourth is that the world has become more industrialized. A fifth is that the world has become more developed.

There are a number of reasons for this. One is that the world population has increased from 5 billion to 6 billion. Another is that the world has become more urbanized. A third is that the world has become more affluent. A fourth is that the world has become more industrialized. A fifth is that the world has become more developed.

There are a number of reasons for this. One is that the world population has increased from 5 billion to 6 billion. Another is that the world has become more urbanized. A third is that the world has become more affluent. A fourth is that the world has become more industrialized. A fifth is that the world has become more developed.

There are a number of reasons for this. One is that the world population has increased from 5 billion to 6 billion. Another is that the world has become more urbanized. A third is that the world has become more affluent. A fourth is that the world has become more industrialized. A fifth is that the world has become more developed.

There are a number of reasons for this. One is that the world population has increased from 5 billion to 6 billion. Another is that the world has become more urbanized. A third is that the world has become more affluent. A fourth is that the world has become more industrialized. A fifth is that the world has become more developed.

There are a number of reasons for this. One is that the world population has increased from 5 billion to 6 billion. Another is that the world has become more urbanized. A third is that the world has become more affluent. A fourth is that the world has become more industrialized. A fifth is that the world has become more developed.

There are a number of reasons for this. One is that the world population has increased from 5 billion to 6 billion. Another is that the world has become more urbanized. A third is that the world has become more affluent. A fourth is that the world has become more industrialized. A fifth is that the world has become more developed.